

Comparing standard and revised MPN  
treatment in 10G Ethernet spreadsheet  
models, benchmarking to 10GBASE-SR

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# MPN treatment background

- During the MMF ad hoc calls there was some discussion on whether Mode Partition Noise (MPN) penalty is enhanced by ISI, and how it should be treated in the 10GE spreadsheet model.
  - [Mode partition noise handling in spreadsheet model](#) \*  
pepeljugoski\_01\_0112\_NG100GOPTX\_MMFAHoc.pdf
- A revised 10GE spreadsheet model, with an explicit ISI term added to modify MPN penalty, was posted to the MMF ad hoc archive.
  - [10GEPBud3 1 16a 25G with MPN changes pepeljugoski for web](#) \*
- This presentation compares the standard 10GE spreadsheet and revised spreadsheet for legacy 10GBASE-SR link modeling, and reports recent work by David Cunningham on MPN treatment, submitted to Fibre Channel

\*Both available at: <http://www.ieee802.org/3/100GNGOPTX/public/mmfadhoc/meetings>

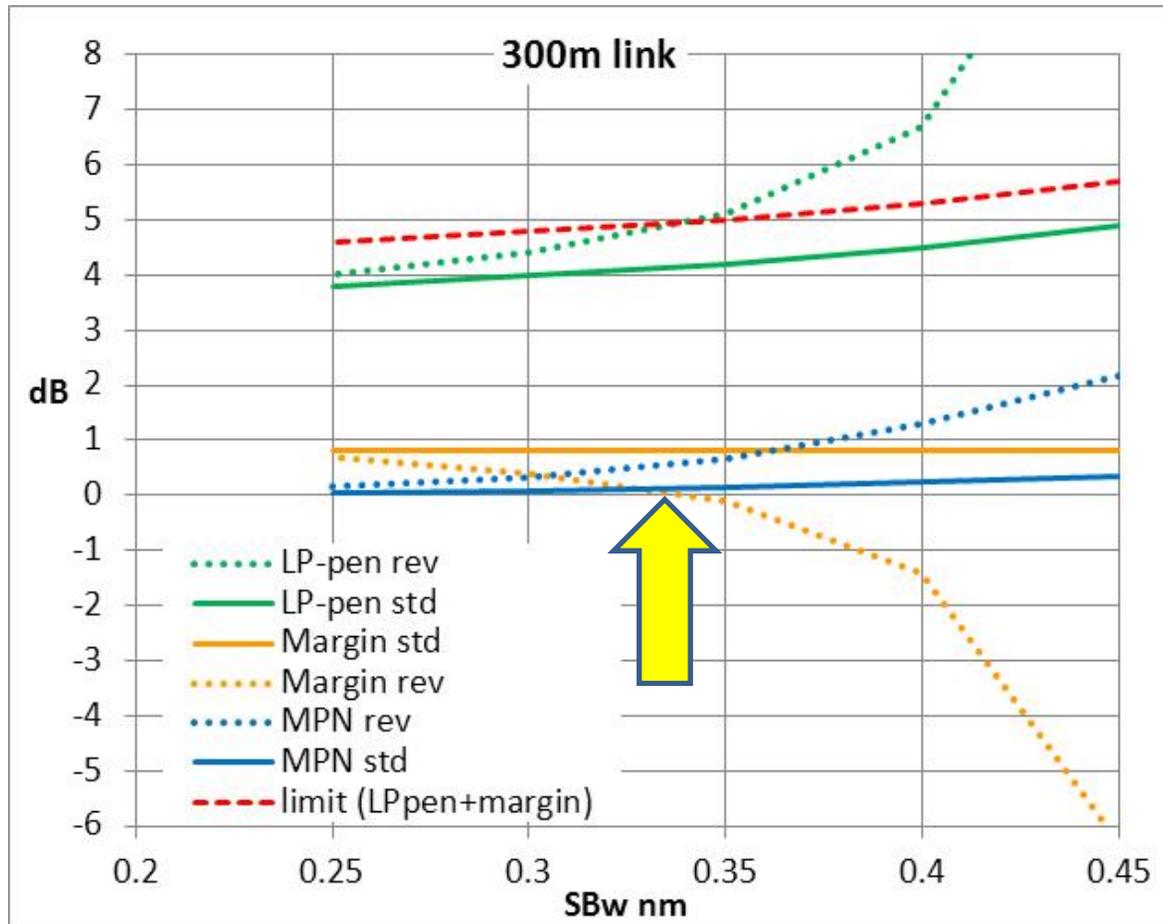
# Set up

- Spreadsheet input parameters were set to identical values for each spreadsheet version (as illustrated in the eye test below)
- LP-pen, margin and MPN penalty were compared for 300m and 280m OM3 link lengths, for various source spectral widths.
  - For each spectral width, Tx OMA was set to the minimum allowed by the triple trade off table in cl.52, 802.3, assuming worst case TDP

Spreadsheet by Del Hanson, David Cunningham, Piers Dawe, David Dolfi Agilent Technologies				Rev. 3.2/3	This file	10GEPBud3_1_16a.xls	of	17-Oct-01
<b>Basics</b>	Input= <b>Bold</b>	Ts(20-80) <b>35</b> ps	Case: 850nm serial <b>newMMF</b>	Attenuation= <b>3.5</b> dB/km	Model/format rev <b>3.1.16a</b>	of	31-Oct-01	
	Q= <b>7.04</b>	Ts(10-90) <b>53</b> ps	Target reach <b>0.28</b> km	Fiber at <b>850</b> nm	NomSens OMA <b>-11.10</b> dBm	Margin <b>-0.89</b> dB at		
	Base Rate= <b>10313</b> MBd	RIN(OMA) <b>-130</b> dB/Hz	and L_start= <b>0.2</b> km	C_att= <b>1.00</b>	Receive Refl Rx <b>-12</b> dB	Answer! <b>0.28</b> km		
<b>Transmitter</b>		RIN at MinER <b>-139.6</b> dB/Hz	graph L_inc= <b>0.008</b> km	Attenuation= <b>3.62</b> dB/km	Rec_BW= <b>8,250</b> MHz	Est Rx BW <b>7500</b> MHz		
Wavelength Uc <b>840</b> nm	RIN_Coef= <b>0.70</b>	Power Budget P= <b>8.30</b> dB		at <b>840</b> nm	c_rx <b>329</b> ns.MHz			
Uw (see notes) <b>0.45</b> nm	Det.Jitter <b>6.0</b> ps inc.	DCD Connections C <b>1.5</b> dB		Disp. min. Uo= <b>1320</b> nm	T_rx(10-90) <b>39.9</b> ps	Test Source ER=		
Tx pwr OMA= <b>-2.80</b> dBm	DCD_DJ= <b>6</b> ps TP3	Pwr.Bud.-Conn.Loss <b>6.8</b> dB		Disp. So= <b>0.11</b> ps/nm^2*km	TP4 Eye <b>19</b> ps	Test Tx <b>6.5</b> dB		
Min. Ext Ratio= <b>3.00</b> dB	Effect. DJ= <b>0.00</b> (U) ex	DCD C1= <b>480</b> ns.MHz		Disp. D1= <b>-117.76</b> ps/(nm.km)	Opening (=Tx eye)	TestERper <b>1.98</b> dBo		
Worst"ave.TxPwr <b>-1.03</b> dBm	MPN k(OMA) <b>0.3</b>	Reflection Noise factor <b>0</b> no units		RMS Baseline wander SD <b>0.025</b> fraction of 1/2 eye				
Ext. ratio penalty <b>4.78</b> dBo	Tx eye height <b>70.7%</b>	Effective Rate <b>10993</b> MBd		(not in use) <b>10</b>		V.E.C.P. <b>3.19</b> dBo		
Tx mask X1= <b>0.3</b> UI	Refl Tx <b>-12</b> dB	Tb_eff= <b>91</b> ps		BWm= <b>2000</b> MHz*km	P_BLW(no ISI) <b>0.07</b> dB			Stressed
X2= <b>0.4</b> UI	ModalNoisePen <b>0.3</b> dB	Effective Rec Eye <b>0.21</b> UI		Eff. BWm= <b>2.0E+03</b> MHz*km	P_BLW <b>0.07</b> dB			Rx sens
Y1= <b>0.25</b>	Tx mask top <b>0.2</b> UI	Pisi P Eye <b>P_DJ P_DJ</b>		Preflection	Pcross	Ptotal <Ptotal		LP Pen <b>OMA</b>

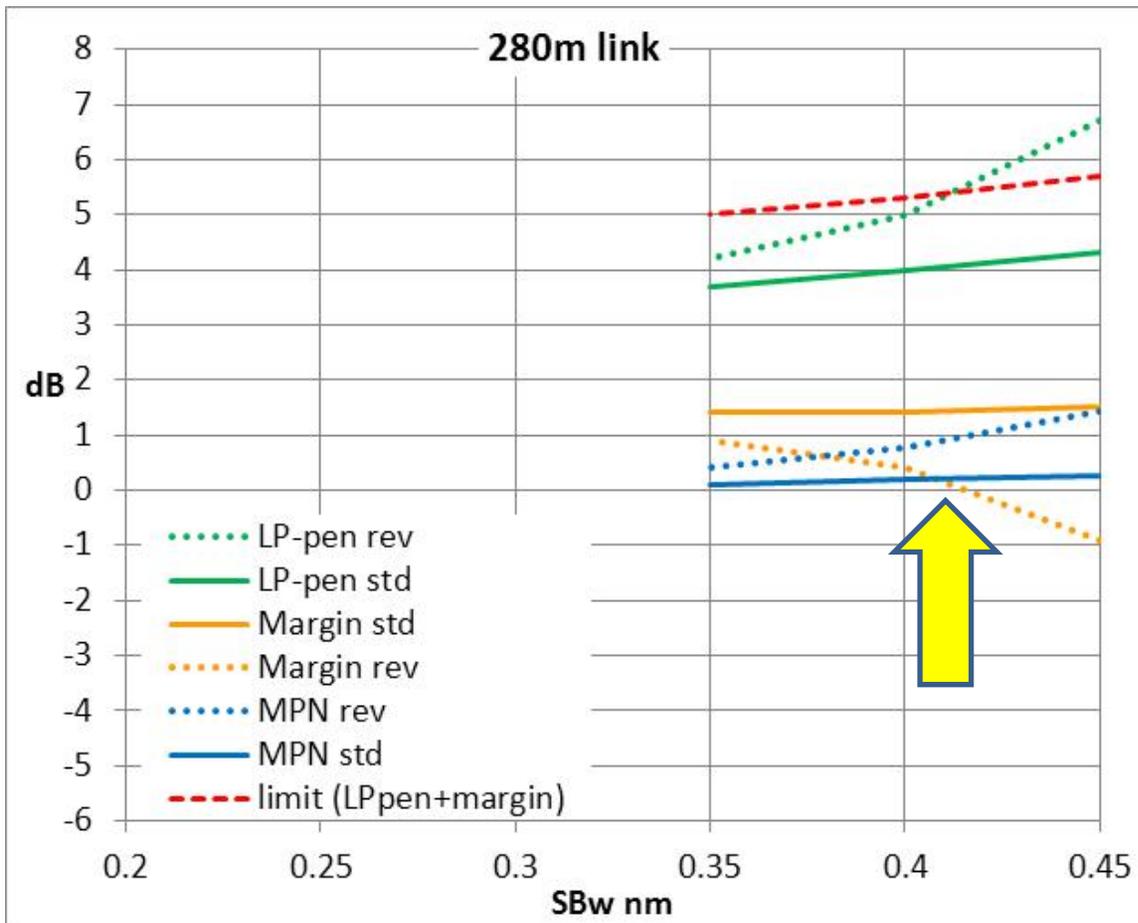
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# 10GBASE-SR, standard vs revised model, 300m OM3



- Revised model shows negative margin for sources with RMS spectral width  $>0.33\text{nm}$ 
  - Quite likely to see this spectral width in practice at room temperature
- Sources with RMS spectral width =  $0.45\text{nm}$  do not have a closed link budget over 300m of worst case fibre.
  - The shortfall is  $>6\text{dB}$ , unlikely to be covered by Tx or Rx margins.

# 10GBASE-SR, standard vs revised model, 280m OM3



- Revised model shows negative margin for sources with RMS spectral width  $>0.41\text{nm}$ 
  - Likely to see this spectral width for some VCSELs at cold temperatures
- Sources with RMS spectral width =  $0.45\text{nm}$  do not have a closed link budget over 280m of worst case fibre.
  - The shortfall is  $>1\text{dB}$ , but could be covered by Tx or Rx margins to spec

# Input to Fiber Channel from David Cunningham - 1

- From Feb 2012 T.11 meeting, contribution number 12-042v0.pdf

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## $\sigma_{mpn}$ & VECP

Eye Diagram A

Time, U.I.

$f_i = \cos(\pi \cdot B \cdot (t_0 - \Delta\tau_i))$

$$\sigma_a = k \cdot \sqrt{\left[ \sum_{i=1}^N a^2 \cdot f_i^2 \cdot \bar{A}_i - \left( \sum_{i=1}^N a \cdot f_i \cdot \bar{A}_i \right)^2 \right]}$$

$$\sigma_a = a \cdot k \cdot \sqrt{\left[ \sum_{i=1}^N f_i^2 \cdot \bar{A}_i - \left( \sum_{i=1}^N f_i \cdot \bar{A}_i \right)^2 \right]}$$

Eye Diagram B

Time, U.I.

$f_i = \cos(\pi \cdot B \cdot (t_0 - \Delta\tau_i))$

$$\sigma_b = k \cdot \sqrt{\left[ \sum_{i=1}^N b^2 \cdot f_i^2 \cdot \bar{A}_i - \left( \sum_{i=1}^N b \cdot f_i \cdot \bar{A}_i \right)^2 \right]}$$

$$\sigma_b = b \cdot k \cdot \sqrt{\left[ \sum_{i=1}^N f_i^2 \cdot \bar{A}_i - \left( \sum_{i=1}^N f_i \cdot \bar{A}_i \right)^2 \right]}$$

$$\therefore \sigma_b = \frac{b}{a} \cdot \sigma_a$$

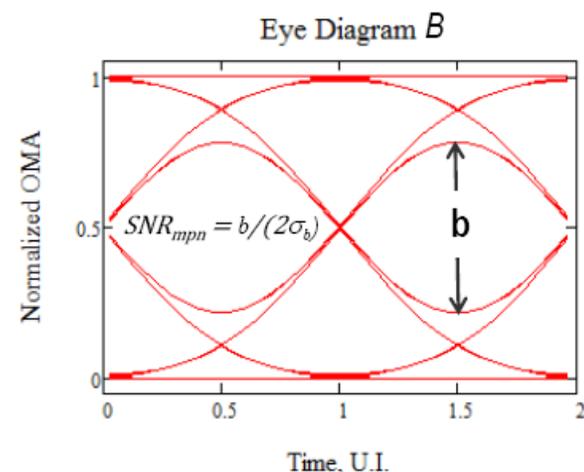
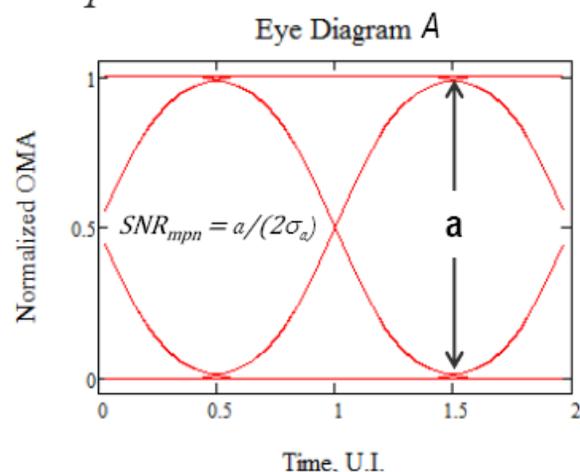
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# Input to Fiber Channel from David Cunningham - 2

- From Feb 2012 T.11 meeting, contribution number 12-042v0.pdf

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## $P_{mpn}$ and VECP for Non-Equalised Links



Since,  $\sigma_b = (b/a)\sigma_a$ , both of the inner eyes, A and B, have the same SNR due to mpn

But signal B has suffered an additional penalty due to ISI of:  $VECP = P_{isi} = 10\log(b/a)$

The link model calculates the loss of signal due to ISI,  $P_{isi}$ , as a separate penalty

The total penalty for signal B, due to the combination of ISI and mpn, is then:  $P_{isi} + P_{mpn}$

This means that the power penalty equation for  $P_{mpn}$  is already normalised with respect to VECP

# Comments and conclusion on MPN treatment

- ISI *should* be accounted for in calculating MPN penalties
  - The standard version of the 10GE spreadsheet model does so; this was demonstrated by David Cunningham (Fibre Channel, Feb 2012 T.11 meeting, contribution 12-042v0.pdf)
- The revised spreadsheet model adds an explicit ISI modifying term (in column Q of the model, MPN penalty calculation)
  - this double counts the effect of ISI on MPN penalty
  - in it's present form, the revised spreadsheet predicts link failures for worst case 300m links, and increased probability of fails for 280m links
- Conclusion:
  - The treatment of MPN penalty in the standard 10GE spreadsheet model already includes the effect of ISI, no modifications are required.

Back up

# Example: 26Gb/s 100m OM4 penalties

