

4D 4-State 3dB/6dB Transparent Trellis Code for 1000BASE-T

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Outline

- Goals
- Current 4D and 24D coding schemes
- The 3dB/6dB transparent 4D 4-state Trellis Code
- Equivalent 16D 2-state Trellis Code interpretation
- Achieving Finite Decoding Delay
- Latency Budgets
- Simulation Results
- Conclusions

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Goals

- **Allow for conventional DFE/DFSE implementations.**
- **Use finite-length decoding**
 - ◆ facilitates use of ISI Cancellers that require dual decoders
- **Based on 4D encoding**
 - ◆ facilitates reuse of PCS principles from 100BASE-T2
- **3dB/6dB Transparency**
 - ◆ Achieve 3dB nominal coding gain with symbol decoding
 - ◆ Achieve 6dB nominal coding gain with sequence estimation.
 - ◆ facilitates use of "good" decisions for fast feedback loops such as timing recovery.

Current 4D and 24D coding schemes

- **4D 8-state Trellis Code**
 - ◆ allows for conventional DFE/DFSE implementations
 - ◆ requires long decoding delays especially when used without Decision Feedback Sequence Estimation
 - ◆ based on 4D code subsets
 - ◆ No Transparency
 - ◆ Meets Goal#1 and Goal#3
- **24D Code**
 - ◆ difficult to do DFE/DFSE implementations for Baseband line codes
 - ◆ allows for finite length (6 BT) decoding
 - ◆ based on 8D code subsets
 - ◆ No Transparency
 - ◆ Meets Goal#2

Transparent 4D 4-state Trellis Code

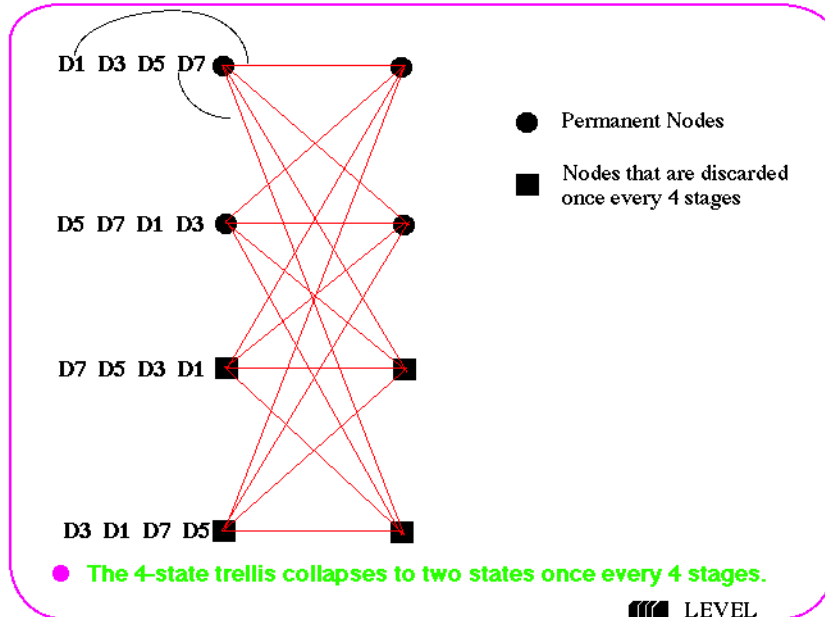
- **Meeting Goals:**
 - ◆ Allows for conventional DFE/DFSE implementations
 - ◆ Achieves 8BT decoding delay (4BT in Block mode) with full 6dB nominal coding gain.
 - ◆ Based on 4D code subsets
 - ◆ Achieves 3dB coding gain with symbol decoding and 6dB coding gain with sequence estimation.
 - ◆ Meets Goal#1, Goal#2, Goal#3 and Goal#4.
- **Reduces complexity of Per-survivor processing by a factor of > 4 over 4D 8-state trellis code.**
 - ◆ reduces power consumption by ~250mW.
- **Provides Error Detection capability for symbol-by-symbol decoded receivers.**

The 4D ODD subsets

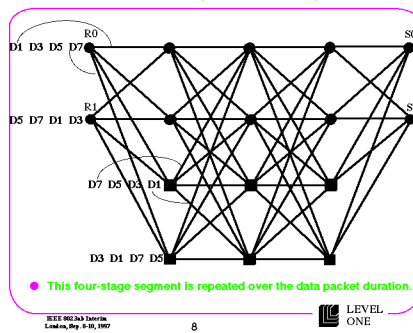
Subset	X-Primary Code (Pair A first)	Y-Primary Code (Pair A first)
D1	XXX \bar{Y}	\bar{Y} YX
D3	XX \bar{Y} X	\bar{Y} X \bar{Y}
D5	X \bar{Y} XX	\bar{Y} X \bar{Y}
D7	\bar{Y} XXX	X \bar{Y} \bar{Y}

- **All four ODD subsets are pair permutations of each other.**
- **Each ODD subset has 24+54 = 78 code points**
- **Each ODD subset has either**
 - ◆ one X point and 3 Y points, or
 - ◆ one Y point and 3 X points
- **Each ODD subset is uniquely determined by its 'unlike' pair.**

The Trellis Diagram - Single Stage



The Trellis Diagram - Four Stages



The Data Encoding Method

- **Encoding is done in blocks of 4 Bytes.**
 - ◆ 7-bits select subset pattern for the 4-stage trellis
 - ◆ 25-bits select a point within the 4 subsets.
- **Designate one point in each subset as Delimiter point and code 25 bits using the other 77 points on 4 subsets.**
 - ◆ No. of points needed = $2^{25} = 33,554,432$
 - ◆ No. of points available = $77^4 = 35,153,041$
- **Ensures 3dB coding gain with symbol-by-symbol decoding since only ODD subsets are used.**

Partitioning an ODD subset

- **Each ODD subset with 78 points can be partitioned as:**
 - ◆ S-ESC Code points (Points with zero or one ESC symbols) = 64 points
 - Points with No ESC symbol = 32 points
 - Points with 1 ESC symbol = 32 points
 - ◆ M-ESC Code points (Points with two or more ESC symbols) = 13 points
 - Points with 2 ESC symbols = 12 points
 - ESC, ESC, ESC, -1, a 3-ESC point = 1 point
 - ◆ Delimiter Code point
 - ESC, ESC, ESC, +1, a 3-ESC point = 1 point
- **Delimiter Code point is not used for data transmission.**

Mapping 25bits to Symbols

- **Case 1: BIT[24] = 0: (Only S-ESC points)**
 - ◆ Code BIT[23:0] using S-ESC code points on the 4 subsets.
- **Case 2a: BIT[24] = 1, AND BIT[23:20] <= 12: (One M-ESC subset)**
 - ◆ Use BIT[23:20] to code one of 13 M-ESC point in M-ESC subset.
 - ◆ Use BIT[19:18] to determine position of M-ESC subset.
 - ◆ Use BIT[17:0] to code S-ESC points in other three S-ESC subsets.
- **Case 2b: BIT[24] = 1, AND BIT[23:20] >= 13: (Two M-ESC subsets)**
 - ◆ Use BIT[23:19] to determine one of 6 combinations of 2 M-ESC subset positions.
 - ◆ Use BIT[18:12] to code one of 128 points from 12X12 2-ESC constellation of the two M-ESC subsets.
 - ◆ Use BIT[11:0] to pick S-ESC points in the other two S-ESC subsets.

Mapping 7bits to Subsets

- Use BIT[31:25] and the previous state bit, PS, to determine subset pattern and New state bit, NS.
- **NS = BIT[31]**
- **PS is set to 0 during Idle and at End of Packet.**

	Index, 2	Index, 1	Index, 0	Trellis State
Subset_0	PS+BIT[26]	BIT[25]	1	BIT[26:25]
Subset_1	BIT[28]+BIT[26]+BIT[25]	BIT[27]+BIT[26]	1	BIT[28:27]
Subset_2	BIT[30]+BIT[28]+BIT[27]	BIT[29]+BIT[28]	1	BIT[30:29]
Subset_3	BIT[30]+BIT[29]	BIT[31]+BIT[30]	1	0 BIT[31]

Note: the recovered BIT[31:25] at the receiver can be read off from the states of the best surviving path.

- **This mapping corresponds exactly to the 4-stage trellis on Page 8.**

Error Detection in Symbol-by-symbol Decoding

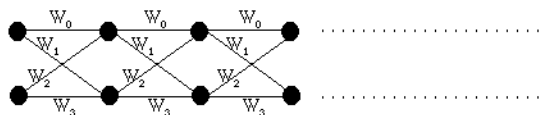
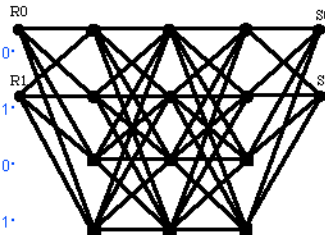
- Assume $SUB\{0:3\}[2:0]$ are the received ODD subsets in symbol-by-symbol decoding for the current block.
- Let PS be the previous state bit ($PS=0$ during idle).
- Then, $BITS[31:25]$ are recovered as follows:

Recovered Bits	MS Bit/Error?	LS Bit
$BIT[26:25]$	$PS+SUB0[2]$	$SUB0[1]$
$BIT[28:27]$	$BIT[26]+BIT[25]+SUB1[2]$	$BIT[26]+SUB1[1]$
$BIT[30:29]$	$BIT[28]+BIT[27]+SUB2[2]$	$BIT[28]+SUB2[1]$
$\{\text{Error? } BIT[31]\}$	$BIT[30]+BIT[29]+SUB3[2]$	$BIT[30]+SUB3[1]$

- The New State Bit, $NS = BIT[31]$.

Equivalent 2-state 16D Trellis code

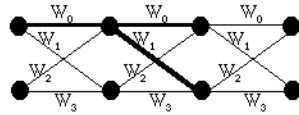
- Define 16D subsets as follows
 - W_0 comprises all paths from R_0 to S_0 .
 - W_1 comprises all paths from R_0 to S_1 .
 - W_2 comprises all paths from R_1 to S_0 .
 - W_3 comprises all paths from R_1 to S_1 .
- MSED between W_i and W_j for $i \neq j$ is 2.
- MSED between points in W_i is 4 for all i .
- The equivalent 2-state 16D trellis code:



The finite decoding delay follows.

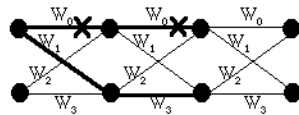
Finite Decoding delay

- **Case 1: Both paths have merged at previous node**



- ◆ Previous 16D symbol has been decoded

- **Case 2: Both paths have not merged at previous node**



The worse path is discarded since it is distance of ≥ 4 away from winning path.

- ◆ Both 16D symbols have been decoded!!

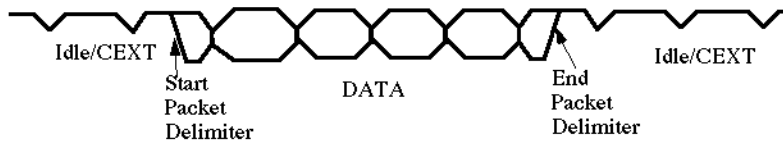
Data Packet Encapsulation Options

- **4-byte Block Trellis Encoding starts at Start Packet Delimiter.**

- ◆ Terminated to state 0 at End Packet Delimiter
- ◆ Idle uses only D1 subset code points
- ◆ Data uses all ODD subset code points



- **4-byte Block Trellis Encoding throughout Idle/Data**



Latency Budgets

- **ISI Canceller Solution (Rockwell presentation from Maui)**
 - ◆ Encoding delay stays the same (?)
 - ◆ Decoding delay for F1 is 8BT.
 - ◆ Decoding delay for F2/F3 is 4BT assuming block processing.
 - ◆ Overall latency stays the same as with 24D code.
- **Conventional DFE Solution**
 - ◆ Encoding delay increases by 4BT.
 - ◆ Decoding delay decreases by 4BT.
 - ◆ Overall latency stays the same with respect to current 4D code.

Simulation Results

● Evaluation with Ideal DFE and AWGN

Noise Level (dBs)	SER@Input (4D Symbols)	SER@Output (Conventional 4D 8-state Trellis Code)	SER@Output (Transparent Code)
-7.0dB	0.093	0.001	0.0013
-8.0dB	0.043	5E-5	8E-5
-9.0dB	0.016	~1E-6	~1E-6

● Evaluation with Full Per-Survivor Processing and AWGN

Noise Level (dBs)	SER@Input (4D Symbols)	SER@Output (Conventional 4D 8-state Trellis Code)	SER@Output (Transparent Code)
-7.0dB	0.093	0.012	0.004
-8.0dB	0.043	8E-4	4E-4
-9.0dB	0.016	1E-5	5E-6

Conclusions

- **4D 4-state trellis code combines the best properties of conventional 4D 8-state trellis code and 24D block code (for passband systems).**
 - ◆ 3dB coding gain with symbol by symbol decoding
 - ◆ 6dB coding gain with sequence estimation
 - ◆ 8BT decoding delay for full 6dB gain
 - ◆ 4BT decoding delay in block mode operation
- **Reduces complexity of per-survivor processing due to**
 - ◆ reduced number of states (4 vs. 8).
 - ◆ automatic pruning of survivor paths to 2 once every 4 stages.
 - ◆ reduces power consumption of implementation.
- **Built-in Error Detection capability for symbol by symbol decoded receivers.**