# Precoder Based 10GBASE-T Architecture Proposal

## IEEE P802.3an March 2004 Plenary

**Scott Powell** 



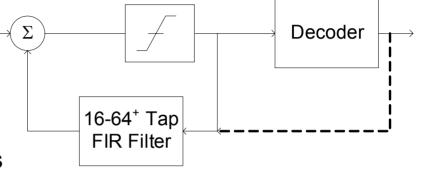
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# **1Gbps Receiver:** the DFSE Loop

## DFE cannot be separated from channel coding

- Catastrophic error propagation
- Zero-delay decisions irreconcilable
   with basic idea of channel coding
- Reduced state, parallel decision
   feedback TCM used in most receivers



## Severely restricts decoder complexity

- Incompatible with block codes

## Critical timing path introduced

- 1Gbps receiver does not scale well to 10Gbps

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# Precoding Avoids DFSE Critical Timing Path

#### Precoding moves DFE to transmitter

- Order of operations doesn't matter in a linear system **Discrete Gaussian** distributed TX RX TX RX Mk(n)Channel Channel a(n)Σ  $\hat{a}(n)$ a(n) $\rightarrow \hat{a}(n) + Mk(n)$ H(z)H(z)H(z) - 1H(z) - 1 Cable+filters+ DFE Precoder transformers+ **Uniformly distributed** ADC/DAC+... between ±M **Spectrally White** 

#### Precoding: no error propagation

- TX DFE operates on ideal, uncorrupted symbols

## Precoding permits more powerful codes

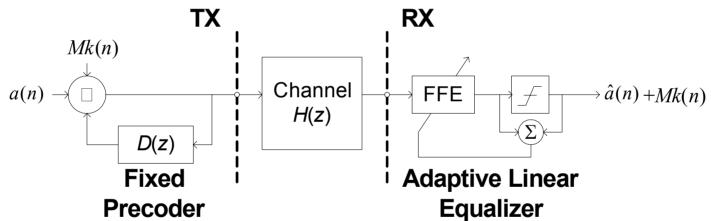
— DFE decoupled from channel coding e v e r y t h i n g<sup>™</sup>



# **Precoder Adaptation Not Necessary**

**Programmable precoding\*** 

- Precoder coefficients chosen at start-up to approximately match channel response
- Adaptive linear RX equalizer removes residual ISI



#### Coefficients are a function of cable length

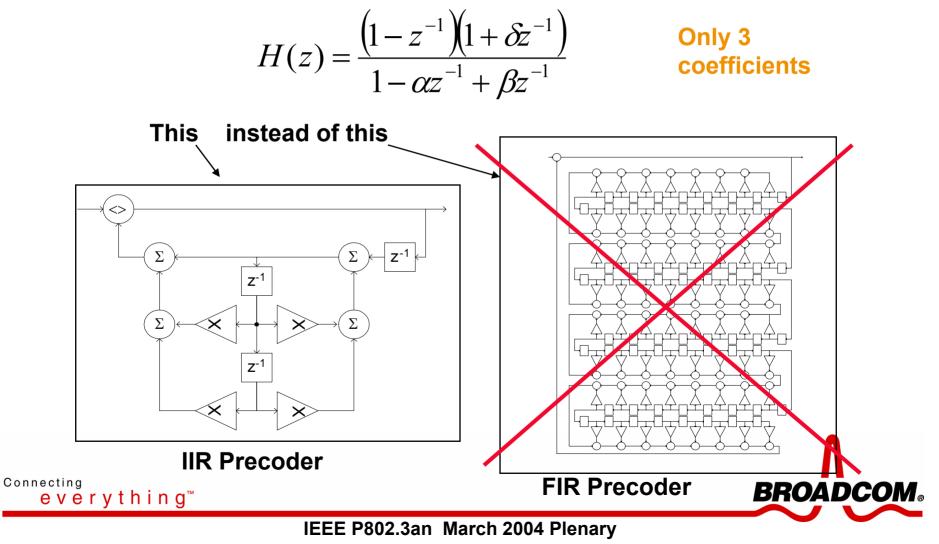
- Pre-store in small look-up table

\* Gerstacker, Fischer, Huber, "A Transmission Scheme for Twisted Pair Lines with Coding, Precoding, and Blind Equalization," Globecom '97, pp 52-56 BROADCO everything

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## Number of Precoder Coefficients Reduced over 10x with IIR Model

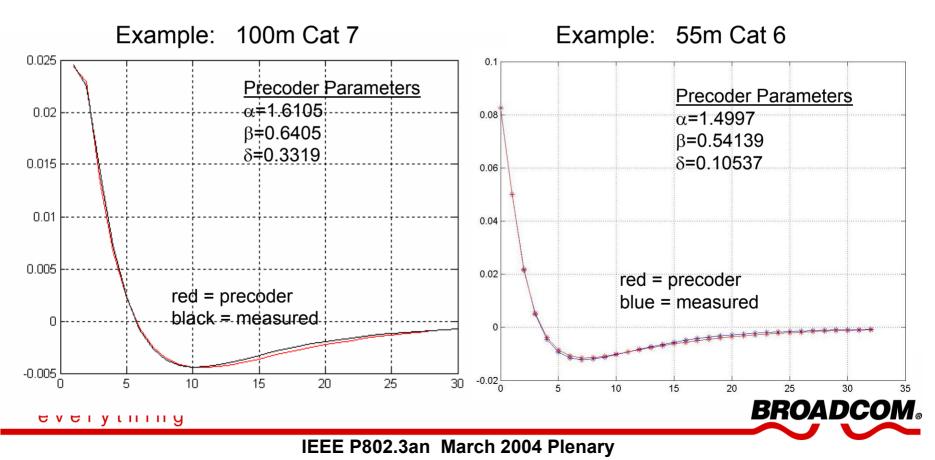
• Overall channel is accurately modeled by 2<sup>nd</sup> order IIR



# **IIR Model a Good Match to UTP**

Parameter values pre-computed to match cable response (compromise precoding)

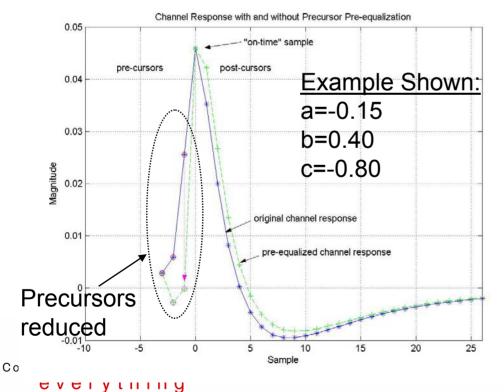
- Tabulated vs length



## Simpler ADC and Start-up with TX Precursor Equalization

Worst case channel response has ≈3 significant precursors

- Merge Precursor pre-equalizer with transmit filter  $T(z) = (1 + az + bz^{2} + cz^{3})z^{-3}$ 



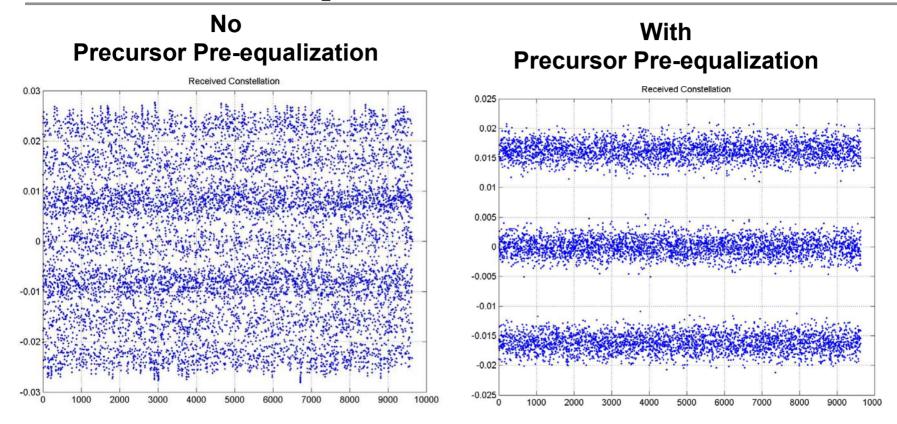
 Reduces dynamic range of signal at the ADC

- No noise enhancement
   small power enhancement
- ✓ Opens eye at t=0 to permit decision directed startup
- Adds small negative precursor to assist timing recovery

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## Precursor Pre-equalized Channel at Start-up (reduced constellation)



 Precursor pre-equalization permits simple decision directed start-up

- Blind start-up not necessary

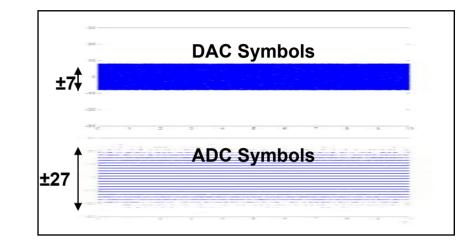
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## DLP Permits Tradeoff Between DAC **Complexity and ADC Complexity**

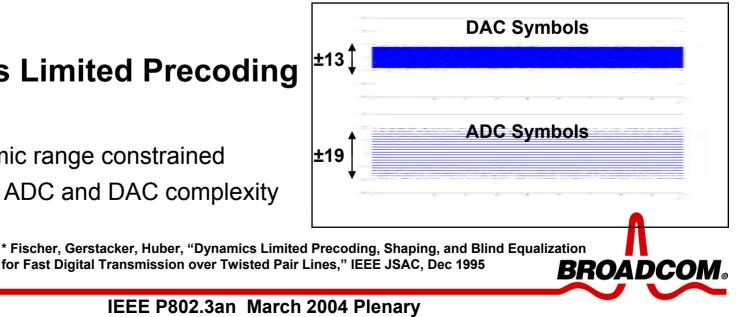
#### **Conventional THP**

- RX dynamic range unconstrained
- ADC more complex than DAC





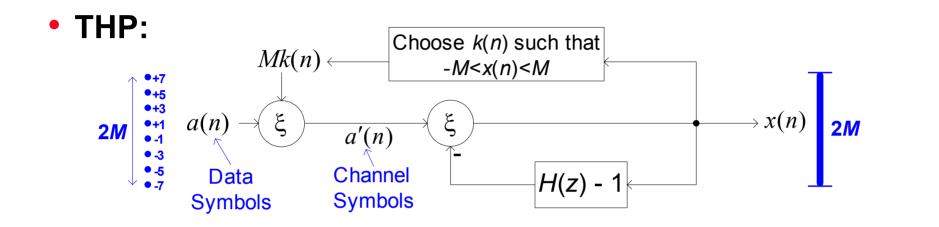
- RX dynamic range constrained
- Balanced ADC and DAC complexity



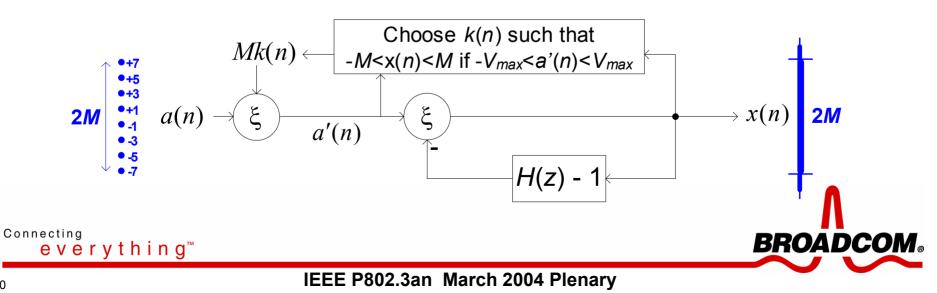
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# **DLP: A Simple Modification to THP**



• DLP:



# **Proposed Precoder Features**

IIR Channel Model

$$H(z) = \frac{(1 - z^{-1})(1 + \delta z^{-1})}{1 - \alpha z^{-1} + \beta z^{-2}}$$

Transmit filter

$$T(z) = (1 + az + bz^{2} + cz^{3})z^{-3}$$

## Stationary parameter values

 At startup, cable length is determined and used to select precoder coefficients from a ROM

## Dynamics limited precoding



# **Precoder Rationale**

## 1) Why precoding vs DFSE ?

- <u>Decouples DFE and decoder</u>
- Permits more powerful coding
- Simplifies timing closure
- No error propagation

## 2) Why IIR precoder ?

- Reduced complexity

# 3) Why compromise precoding ?

- Eliminates need for back channel
- Makes IIR practical

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## 4) Why precursor preequalization ?

- Simplifies start-up sequence
- No noise enhancement
- Reduced ADC complexity

# 5) Why dynamics limited precoding ?

 Reduced dynamic range at ADC input



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# **Precoding vs DFSE: Disadvantages**

## Expanded constellation at receiver

- However: w/o precoding Precoding signal + ISI  $\approx$  signal + expanded levels
- Increased transmit DAC complexity

- M-level DAC  $\rightarrow$  multi-bit DAC

## Power enhancement from Tx pre-equalizer

- Approx equivalent to noise enhancement if moved to Rx

## Not a direct sub/super-set of 1000BT Not a real

- Enough similarities to permit resource sharing

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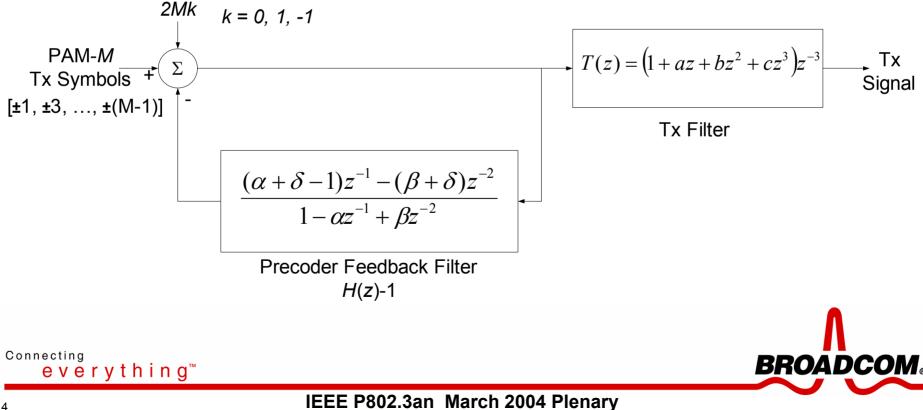
Not a real

disadvantage

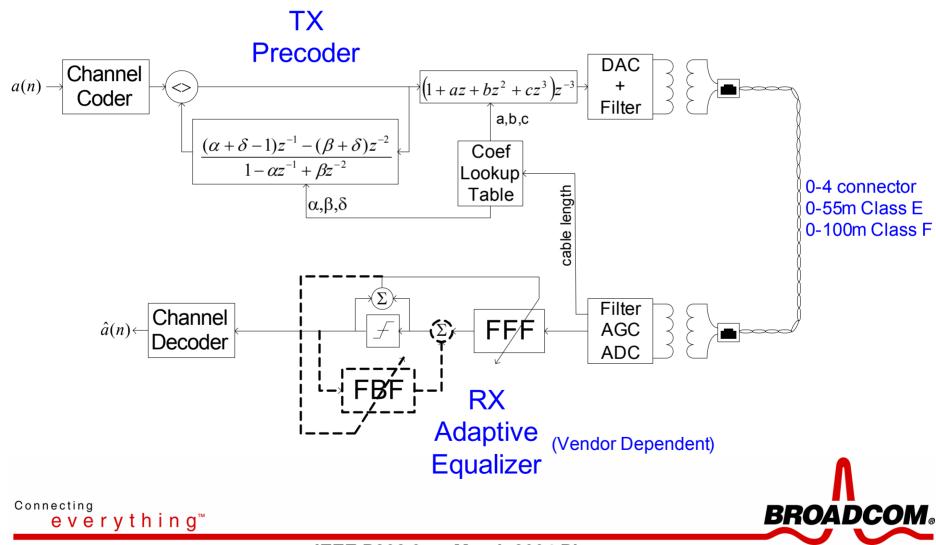
# **Proposed Precoder**

#### Dynamics limited compromise IIR precoder with precursor equalization

-  $\alpha,\beta,\delta,a,b,c$  vs cable length pre-stored in look-up table

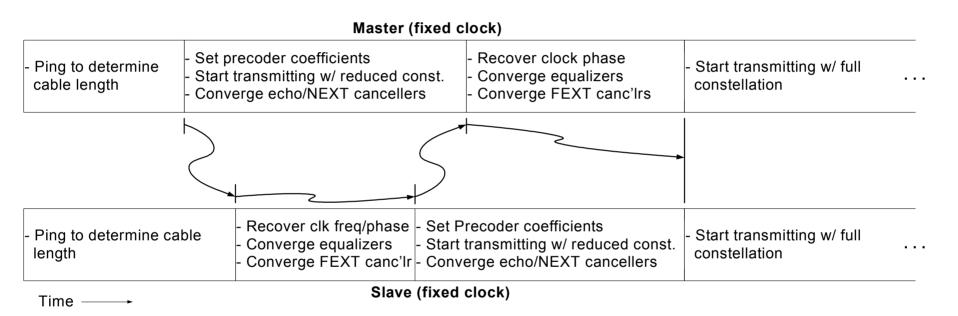


# **Overall System (1 wire)**



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# **Start-up Protocol Similar to 1G**

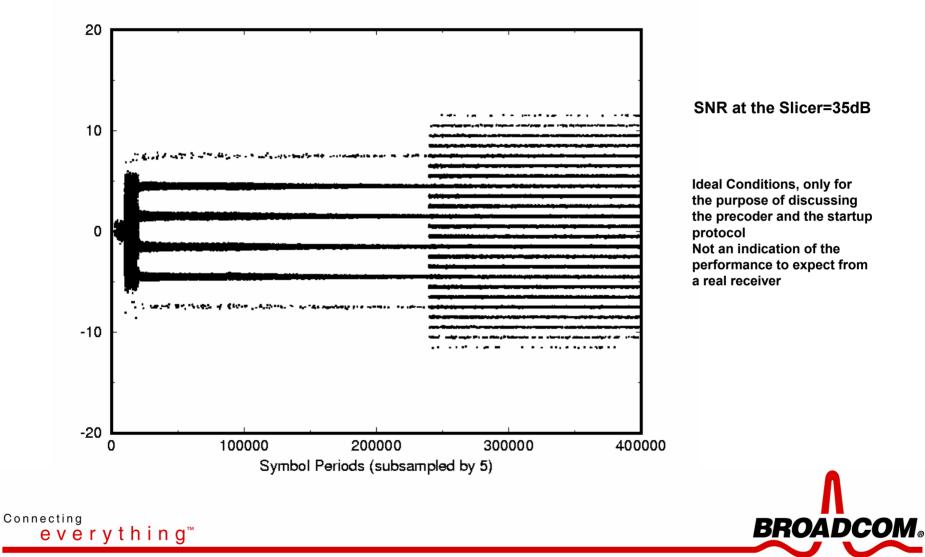






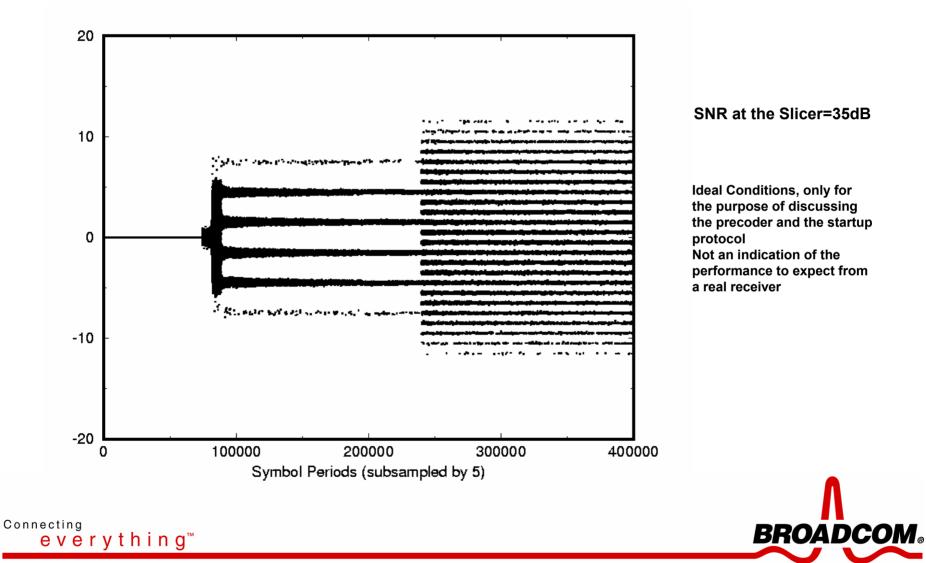
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# Eye Pattern at the Slave (PAM-8)



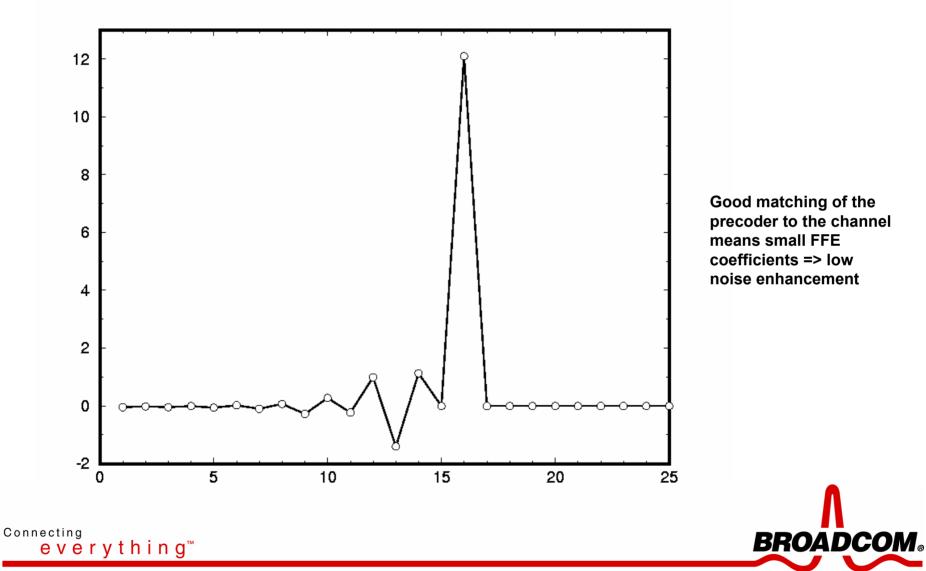
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# Eye Pattern at the Master (PAM-8)



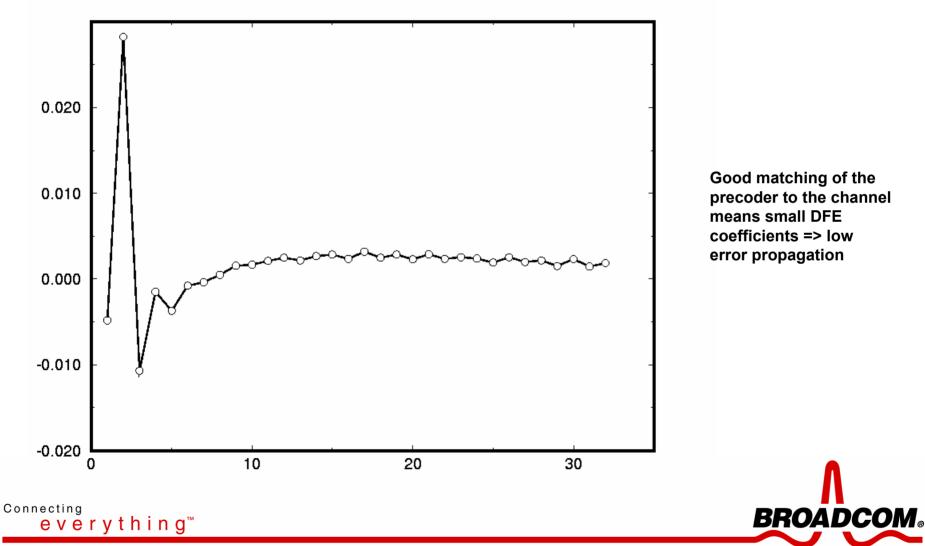
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# **FFE Coefficients**



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# **DFE Coefficients**



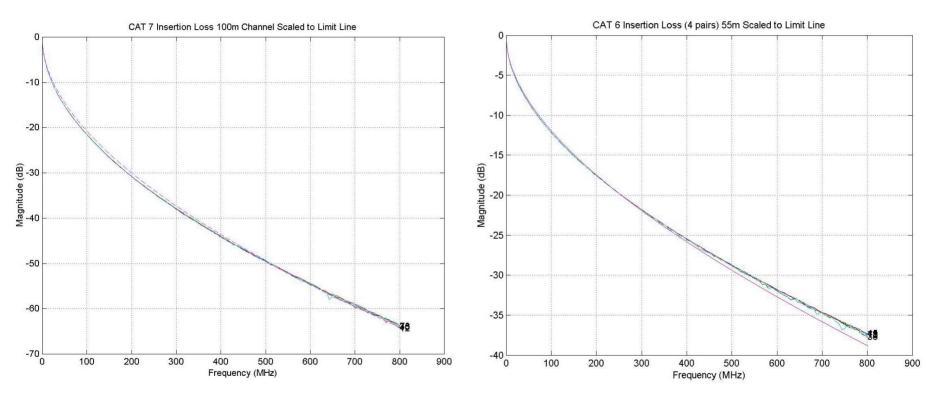
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# **System Simulation Channel Models**

• 100m scaled Cat 7 model

55m scaled Cat 6 model

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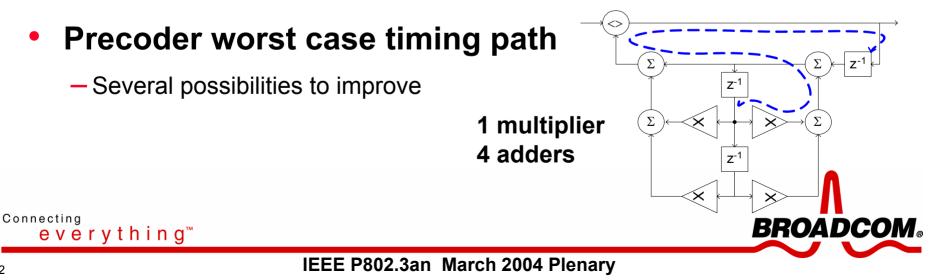


 Measured data linearly scaled (in dB) to touch limit line at maximum rated bandwidth

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## **Open Issues for Further Investigation**

- Cable length (or insertion loss) determination
- Tabulation of precoder filter parameters
  - Evaluation of sensitivity
- Amount of Rx dynamic limiting (DLP)
  - Tx power vs Rx complexity tradeoff





#### Precoding permits more powerful channel codes

- Removes decoding operation from the DFE feedback loop

#### • An IIR precoder is well matched to UTP channels

- Order of magnitude fewer coefficients than FIR
- Well known parallel processing techniques can be applied to the precoder
- Precoder adaptation not necessary with short RX equalizer
  - Coefficients can be set at startup from a LUT based on cable length
- Use of DLP and transmit filtering allows tradeoffs between complexity of the ADC and DAC

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