



Making personal broadband a reality™

Comparison of NRZ, PR-2, and PR-4 signaling

Presented by: Rob Brink

Contributors: Pervez Aziz
Qasim Chaudry
Adam Healey
Greg Sheets

Scope and Purpose

- Operation over electrical backplanes at 10.3125Gb/s is investigated using NRZ, PR2, and PR4 signaling.
- A common equalizer architecture is used in all cases.
- Estimated BER, as well as voltage and timing margin at 1E-12, is reported.

Agenda

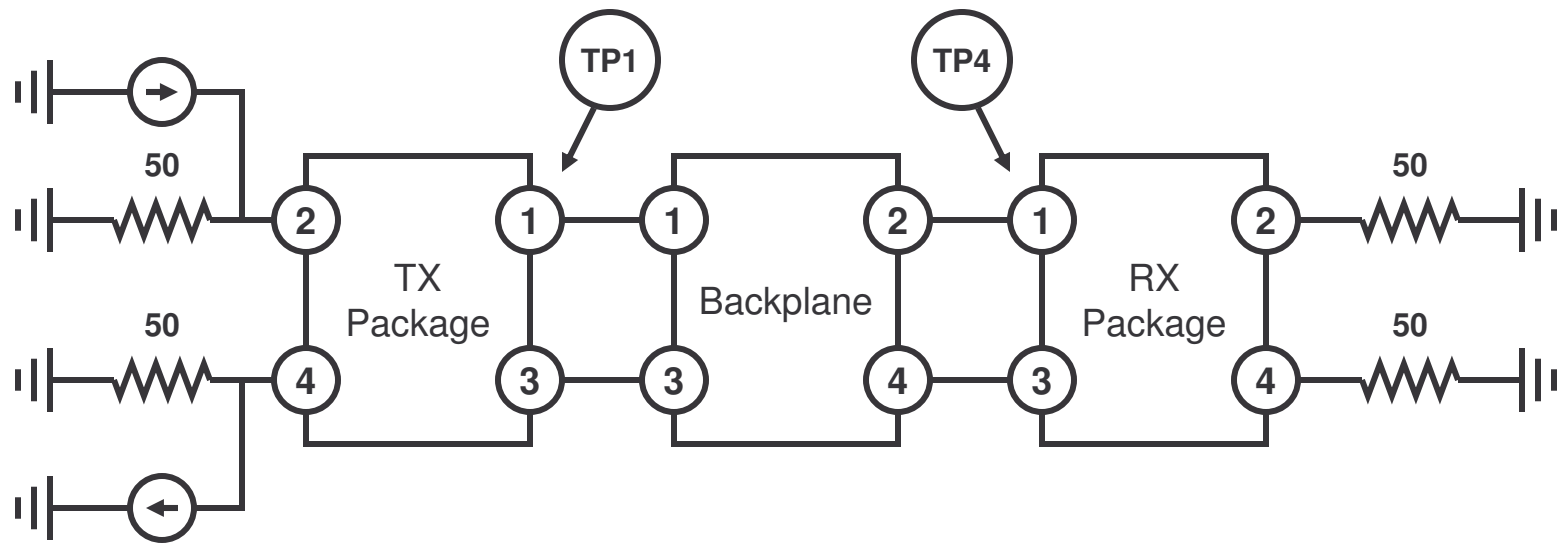
- Simulator Overview
 - Link Model
 - Transmitter Model
 - Receiver Model
 - Equalization Strategy
- Test Cases
- Sample Results
- Results Summary



Making personal broadband a reality™

Simulator Overview

Link Model



TX Package Model	Mellitz "Capacitor-Like" Model
Backplane Model	As described in "Test Cases" section...
RX Package Model	Mellitz "Capacitor-Like" Model

Crosstalk (1/2)

- For each crosstalk aggressor...
 - The response to a PRBS-15 pattern (with an additional trailing “0”) is computed.
 - This response is sampled at baud-spaced intervals at 16 offsets from 0 to $(15/16)T$ in $T/16$ steps.
 - At each offset, the amplitude distribution of sampled response is computed.
 - The aggressor amplitude distribution is defined as the average of the amplitude distributions computed at each sample offset.

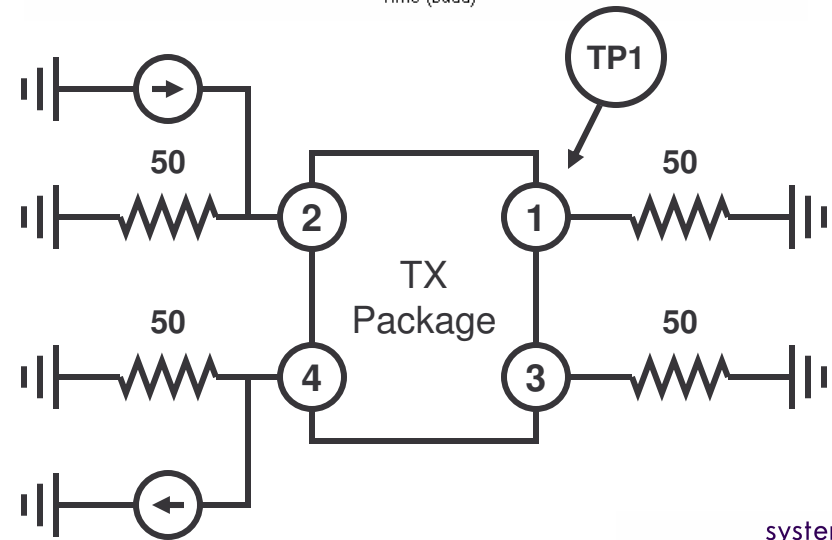
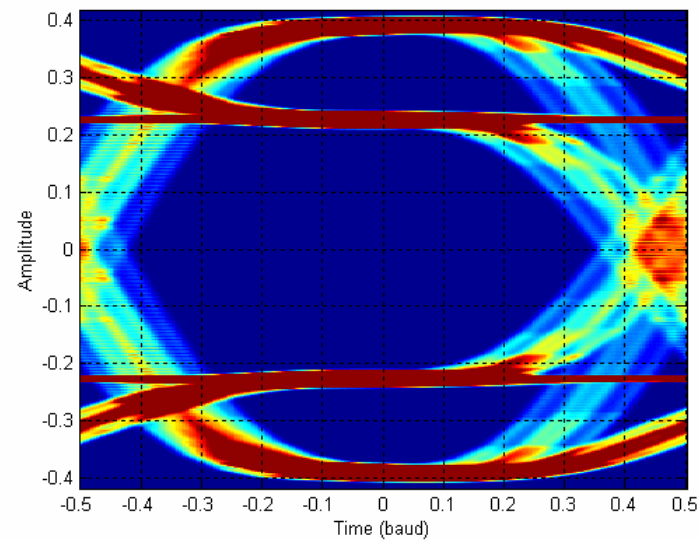
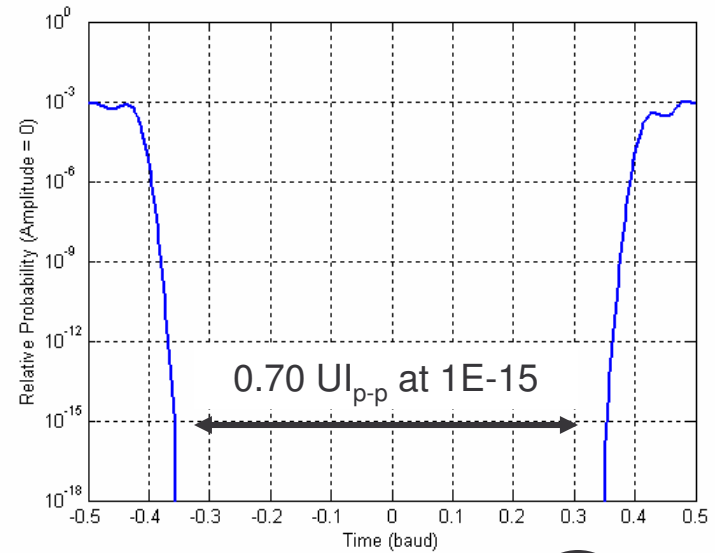
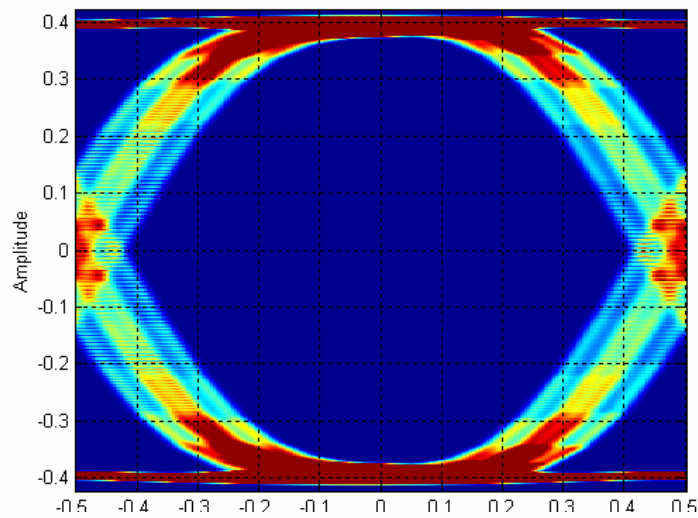
Crosstalk (2/2)

- The overall crosstalk distribution is defined as the convolution of the individual aggressor distributions.
- The effect of crosstalk on the eye is modeled as the convolution of the overall crosstalk distribution and amplitude distribution of the “thru” path at each sample phase.
- This methodology is has been previously described by Moore:
 - http://ieee802.org/3/ap/public/channel_adhoc/moore_c1_0305.pdf
- Computed RMS and peak-peak crosstalk will be reported.

Transmitter Model

- Transmitter differential output voltage fixed at $800\text{mV}_{\text{p-p}}$.
- Transmit filter is Gaussian.
 - Rise Time (20-80%): 24 ps
- Transmitter output jitter
 - Duty Cycle Distortion: $0.05 UI_{\text{p-p}}$ (even-odd)
 - Deterministic Jitter: $0.10 UI_{\text{p-p}}$ (sinusoidal)
 - Random Jitter: $0.15 UI_{\text{p-p}}$ (at $1\text{E-}15$), $9.4\text{m}UI_{\text{rms}}$
- Parameters defined at package model input and do not include package parasitics.
 - The impact of the package model is investigated in the next slide.

Transmitter Output at TP1



Receiver Model

- Receiver modeled as a single pole at 75% fbaud.
 - Noise Bandwidth (B_n): 11.4 GHz $B_n = \int_0^\infty |H(f)|^2 df$
 - Noise Figure: 18 dB
 - $\text{sqrt}(4kTRB_n)$: 1.08 mV_{rms}
- Receiver jitter:
 - Random Jitter: 0.15 UI_{p-p} (at 1E-15), 9.4 mUI_{rms}
- No gain stages have been included in the receive path.

Equalization Strategy

- Transmitter Finite Impulse Response filter.
 - 3 taps, T-spaced with “infinite” tap weight resolution.
- Receiver Decision Feedback Equalizer
 - 5 taps, unconstrained with “infinite” tap weight resolution.
- Sequential adaptation
 - Transmit FIR is adapted first, then the DFE.
- Sample phase chosen to minimize mean-squared error.
 - T/32 resolution



Making personal broadband a reality™

Test Cases

Test Patterns

- Equalizer trained with PN-11 pattern with a trailing “0”.
- Equalizer settings are then frozen.
- Voltage and timing margin is estimated based on PN-15 pattern with a trailing “0”.
 - Thru, NEXT, and FEXT channels share the same output amplitude ($800 \text{ mV}_{\text{ppd}}$) and transmit FIR settings.
- Decision threshold set at the mid-point between nominal signal levels, as determined by the $1\text{E-}4$ contour.
 - Reported margins are twice the minimum distance from the sample phase (or threshold) to the BER contour of interest.

Channels Studied

- All Tyco Electronics channels
 - Test Cases 1 through 7
 - Note that the FEXT channels were included twice due to connector symmetry.
- Recommended subset of Intel channels
 - Test Cases 8, 11, and 12 map to T1, T12, T20
 - Test Cases 14, 17, and 18 map to B1, B12, B20
 - Test Cases 20 and 24 map to M1, and M20
- Other channels not simulated due to a lack of time...



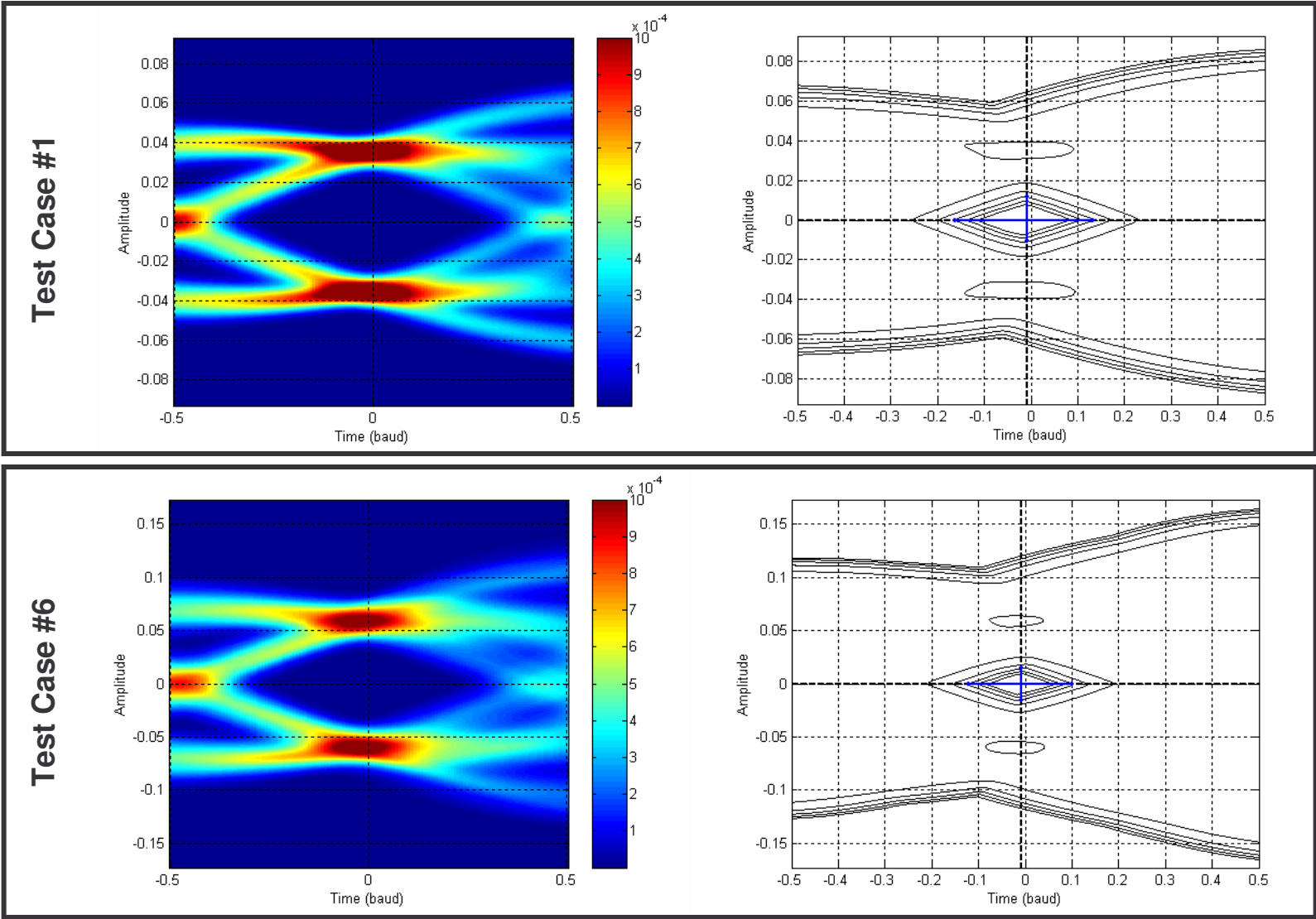
Making personal broadband a reality™

Sample Results

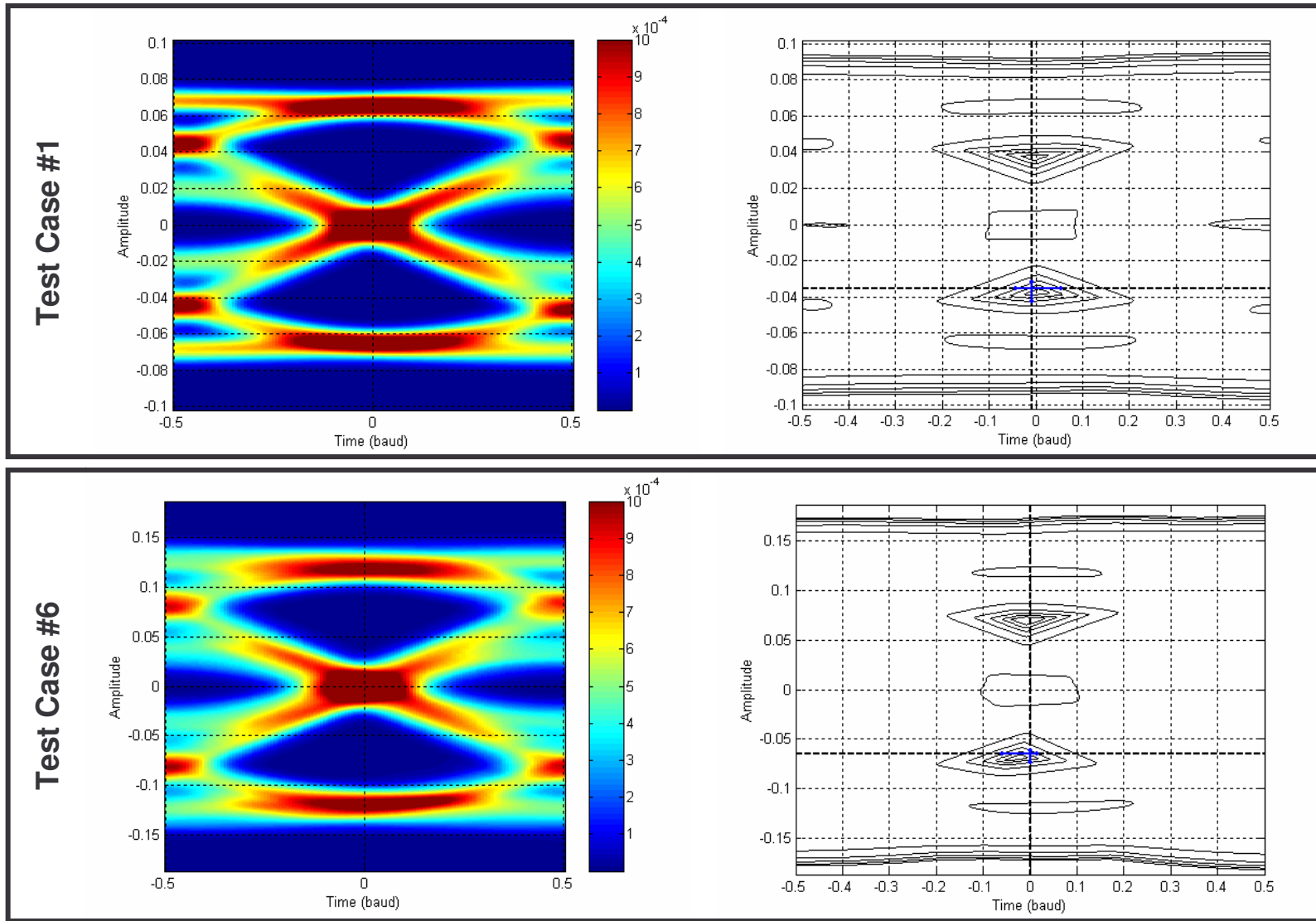
Disclaimer

- The following results are based on PRBS-11 pattern and are included for illustrative purposes.

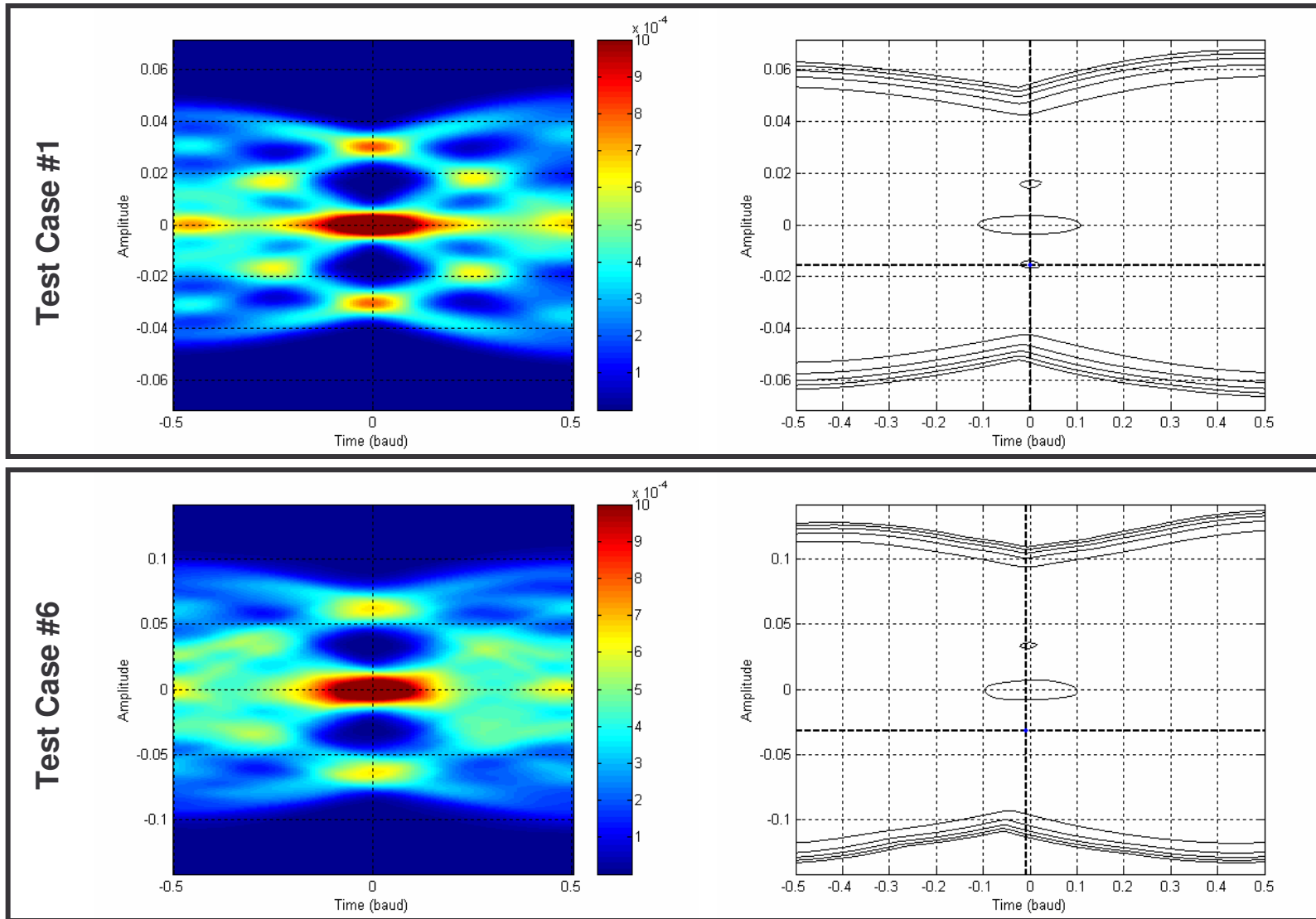
NRZ Sample Results: Eye at Slicer Input



PR2 Sample Results: Eye at Slicer Input



PR4 Sample Results: Eye at Slicer Input

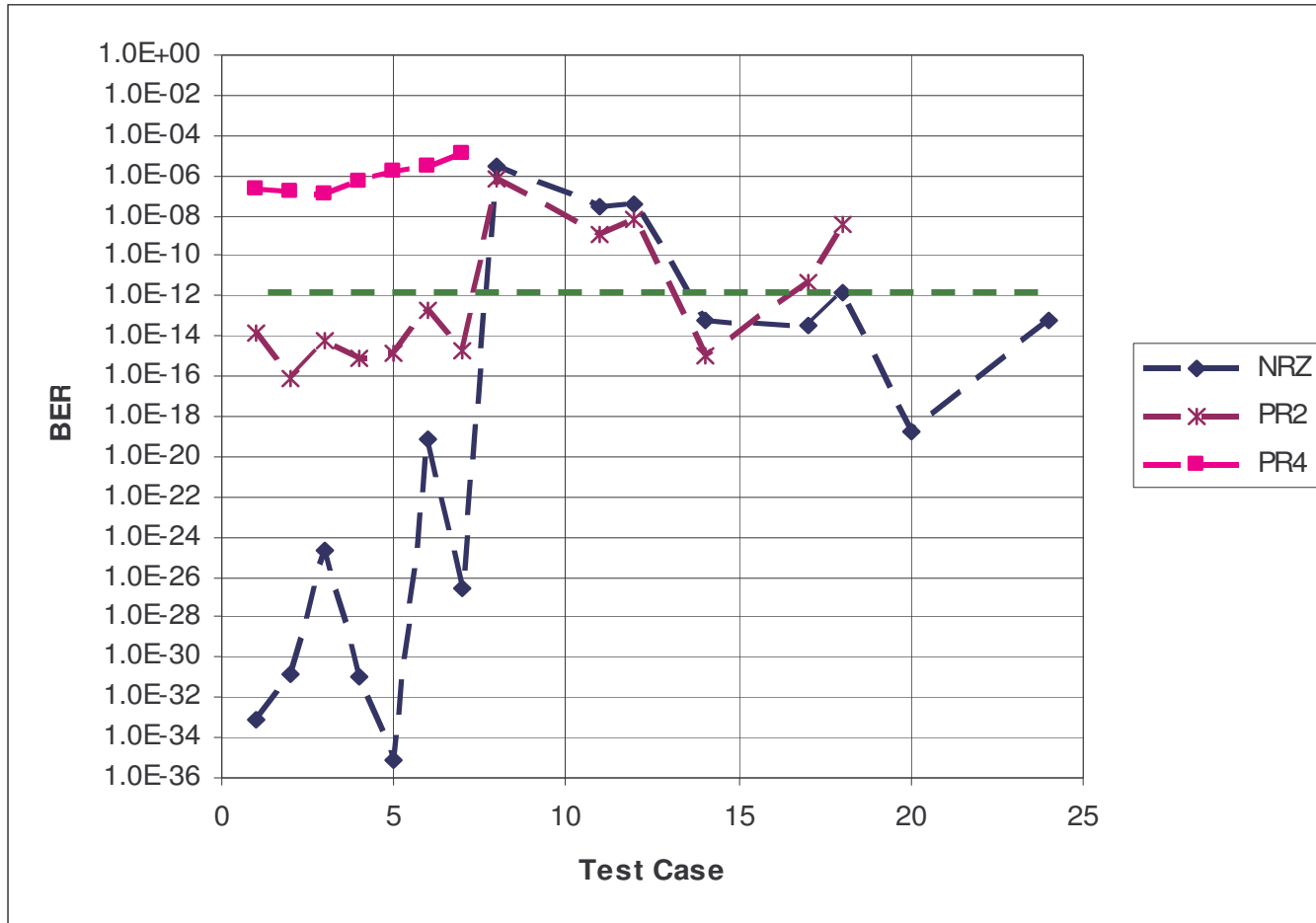




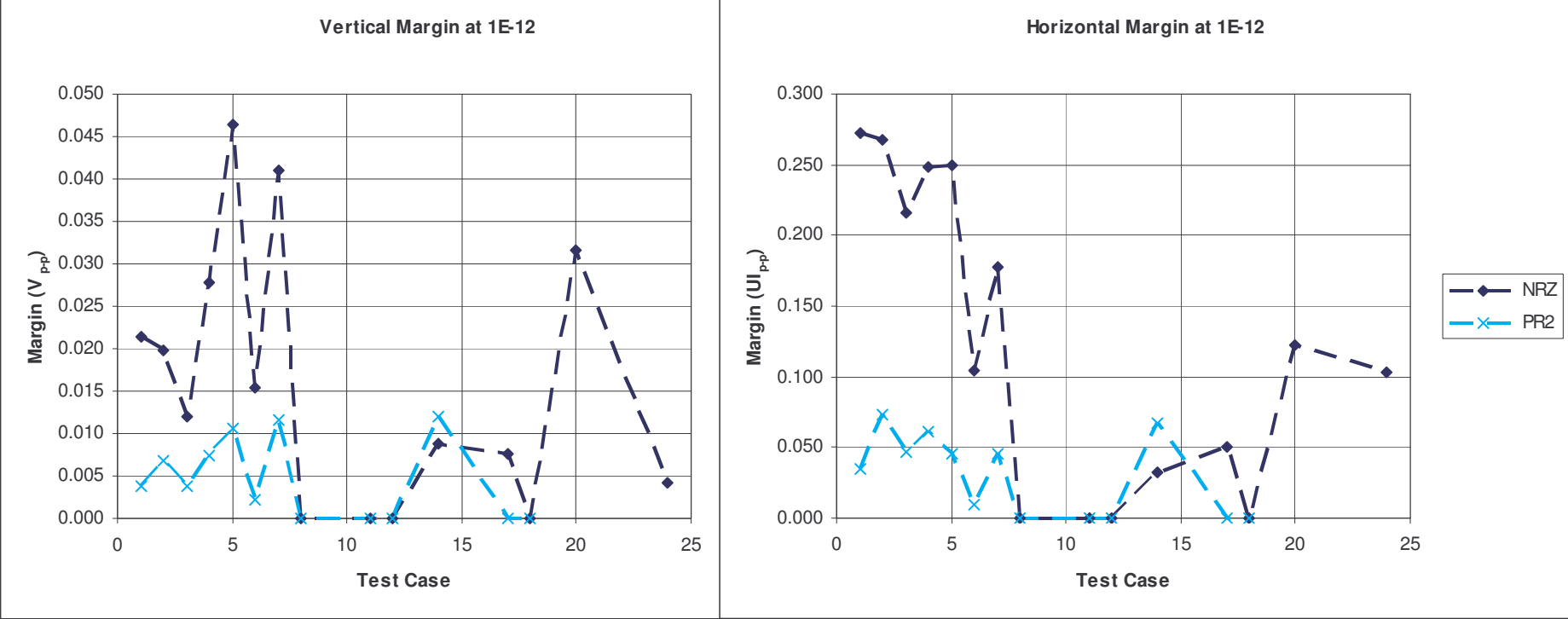
Making personal broadband a reality™

Results Summary

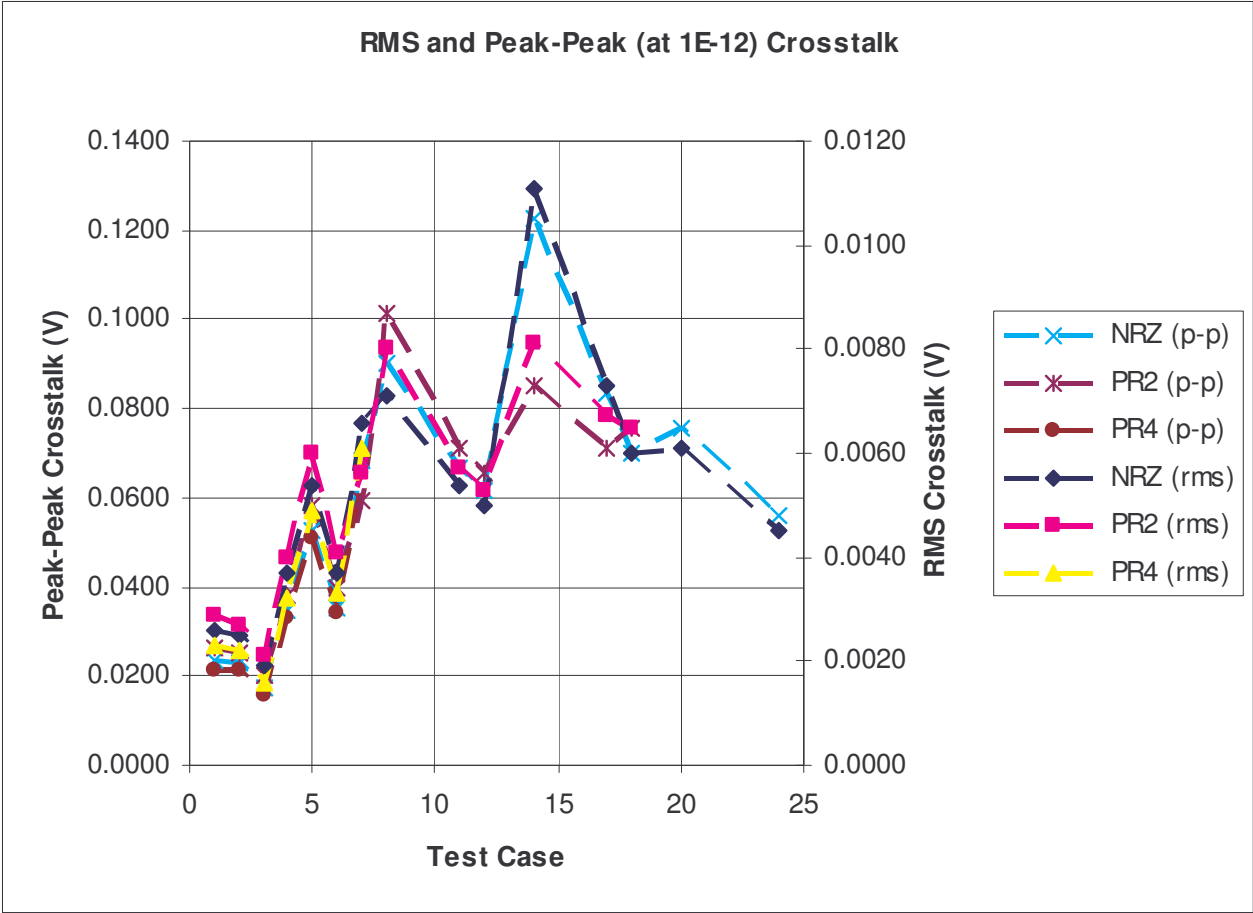
BER Estimates



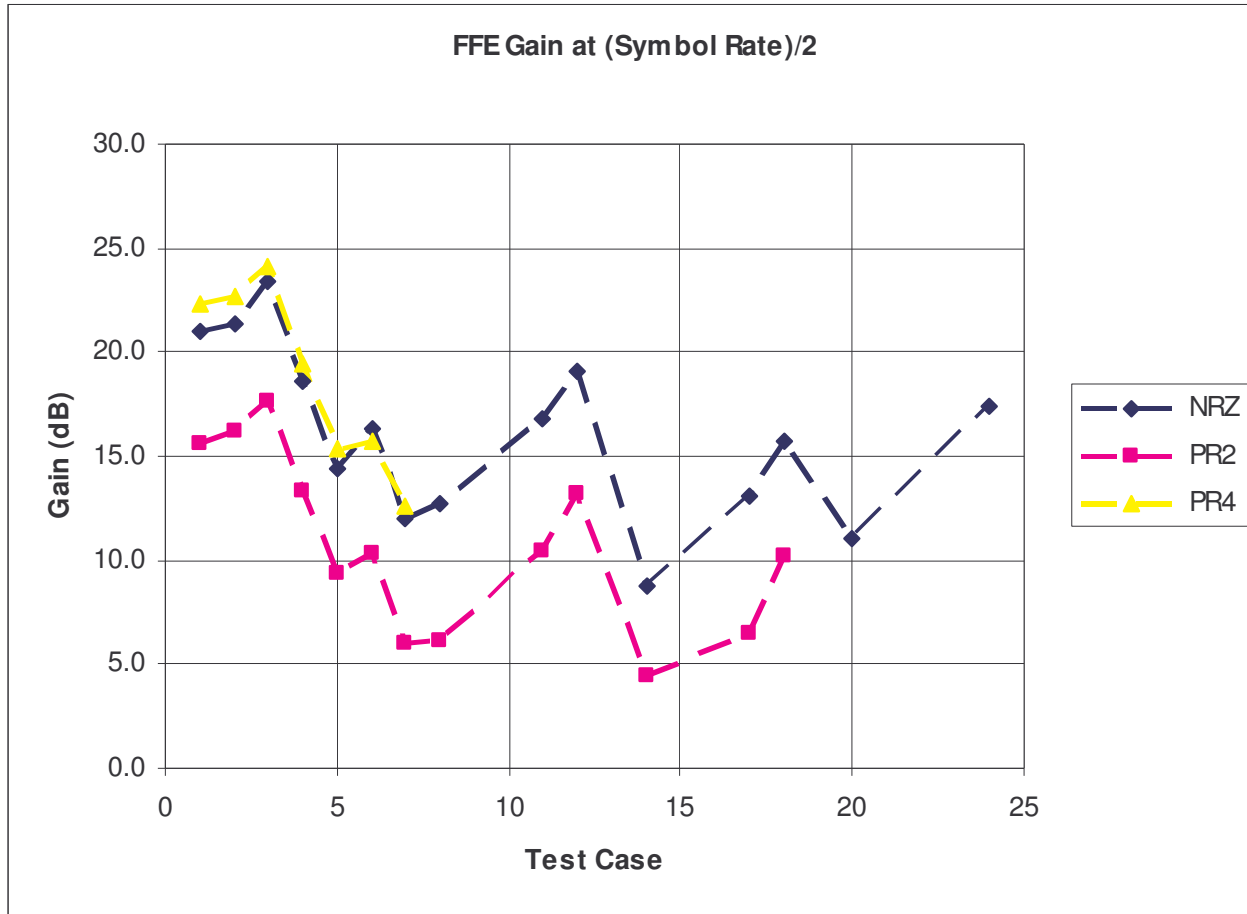
Voltage and Timing Margin at 1E-12



Crosstalk Environment



Required FFE Boost



Conclusions (1/2)

- The target response for PR4 is a poor fit to the channel and therefore higher equalizer complexity is required to achieve acceptable performance.
- NRZ and PR2 both support 1E-12 operation over the Tyco channels.
 - In general, PR2 requires considerably less boost to achieve this objective.
 - In the majority of test cases studied, NRZ offered superior voltage and timing margin.

Conclusions (2/2)

- The Intel “T” channels were not supported by any of the signaling schemes studied with the chosen equalizer architecture.
- NRZ signaling may be feasible for select Intel “B” and “M” channels.
 - Crosstalk is a significant impairment on these channels.



Making personal broadband a reality™

Backup

Relationship Between Crosstalk and Boost

