

Joel Goergen
Beth (Donnay) Kochuparambil
Cisco Systems Inc.
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Dk & Df ALGEBRAIC MODEL v1.01

Let's take a look...

WHAT YOU WILL SEE

LOOK & FEEL

NOTE: The Change Parameter window is a visual basic macro.
If you save the file to your computer, be sure to select the Maco-Enabled file type.

Material/Design Characteristics		Freq	Dk	Df
Simple/Detailed		1.00E+08	3.6	0.0092
Char. Imp. (Ohm)		1.00E+09	3.6	0.0092
Length (inch)	40	2.00E+09	3.5	0.0115
Trace Width (mil)	8	5.00E+09	3.5	0.0115
Cu Thickness (mil)	0.67	1.00E+10	3.4	0.0125
Diel. Thickness (mil)	17	2.00E+10	3.2	0.014

DK fit to Second Order Equation		Df fit to Second Order Equation	
c2	3.64774E-23	c2	-1.1521E-23
c1	-2.02392E-11	c1	4.54439E-13
b	3.59127E77	b	0.009448676

Loss at 6GHz: 20.88 dB
Loss at 12.5GHz: 40.54 dB
Loss at 14GHz: 44.98 dB

CHANGE PARAMETERS

Click to change the input parameters

Quick preview of select frequencies

Fitted equation(s):
 $Dk(f) = c2 * f^2 + c1 * f + b$

Tab allows the user to input material and design specifics

Material Parameters

Frequency (GHz)	Dielectric Constant	Loss Tangent
0.01	49.92196	0.378882
0.05	49.92759	0.89306
0.1	49.93462	1.311603
0.15	49.94166	1.652774
0.2	49.9487	1.953647
0.25	49.95574	2.229015
0.5	49.99096	3.404911
0.75	50.02623	4.414474
1	50.06154	5.341849
1.25	50.09689	6.219524
1.5	50.13229	7.064073
1.75	50.16772	7.885265
2	50.20321	8.6894
2.25	50.23873	9.480792
2.5	50.2742	10.26253
2.75	50.30969	11.0369
3	50.3452	11.80563
3.25	50.3807	12.57005
3.5	50.4162	13.33122
3.75	50.4517	14.08995
4	50.4872	14.84692
4.25	50.5227	15.60264
4.5	50.5582	16.35754
4.75	50.5937	17.11196
5	50.63236	17.86617

Dielectric Constant

Loss Tangent

Frequency (GHz)

UserInput Meg6 ImpvFR4 KR Spec AttenuationPlot

LOOK & FEEL

NOTE: The Change Parameter window is a visual basic macro. If you are having issues opening this window, consider Microsoft button > Excel Options > Trust Center > Trust Center Settings > Macro Settings

The image shows a software interface for entering material parameters. On the left is a table with various material properties. The main window is titled 'Enter Parameters for the Material' and contains a table for Dielectric Constant (real) and Loss Tangent. A large orange circle highlights the table with the text 'Numbers here are for example only, not a specific material'. Below the table is a diagram of a microstrip line on a dielectric substrate between two ground planes, with labels for width (w), length (L), thickness (t), and substrate thickness (b). At the bottom are input fields for b, w, L, and t, along with 'Clear', 'OK', and 'Close' buttons. A callout bubble points to the 'Close' button with the text 'No values will be changed on the sheet if you close'. Another callout bubble points to the input fields with the text 'All boxes must have a value entered'. To the right is a graph titled 'Material Parameters' showing Loss Tangent vs. Frequency (GHz) with data points and trend lines.

Frequency	Dk	Df
100 MHz	3.8	0.006
1 GHz	3.8	0.006
2 GHz	3.7	0.0065
5 GHz	3.7	0.008
10 GHz	3.6	
20 GHz	3.4	0.013

Material Parameters Graph Data (Approximate):

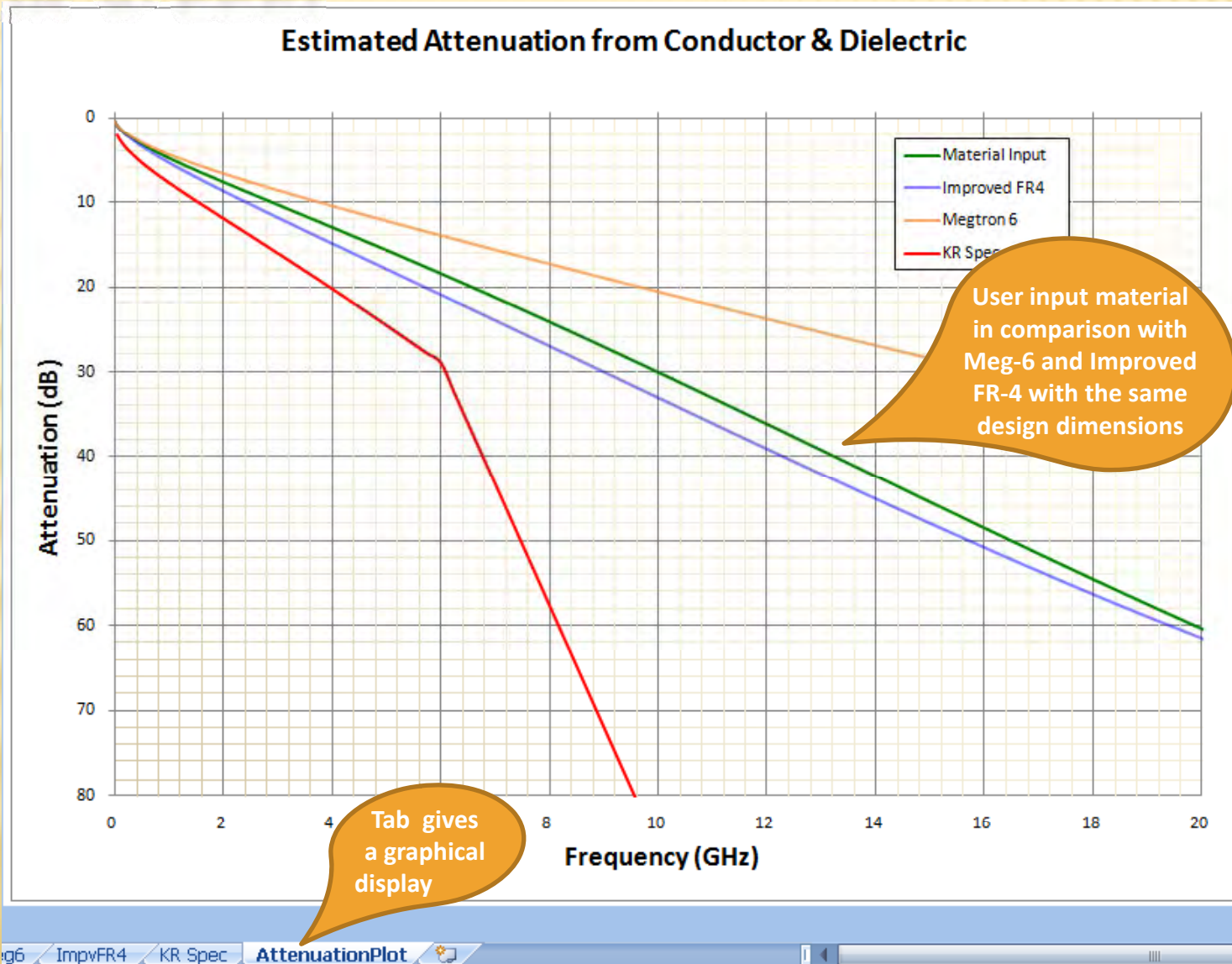
Frequency (GHz)	Loss Tangent
4	0.012
5	0.011
10	0.013
20	0.014

Numbers here are for example only, not a specific material

All boxes must have a value entered

No values will be changed on the sheet if you close

LOOK & FEEL



Behind the Scenes...

EQUATIONS AND REFERENCES OF MODEL

FREQUENCY DEPENDENCE

- ✘ 6 input frequency points for D_k and D_f
- ✘ Fit D_k and D_f to second order equations
 - + Coefficients shown on sheet
 - + Graphical representation shown on sheet
- ✘ Note that frequency dependence fit is only approximated to 20G, therefore, loss approximations should only be considered to 20G
- ✘ Z_0 is calculated with D_k (or ϵ_r) at a given frequency; a similar technique is used in loss calculations

CHARACTERISTIC IMPEDANCE [2, EQU 4-5]

$$Z_0 = \frac{94.15}{\sqrt{\epsilon_r} \left(\frac{\frac{w}{b}}{1-t} + \frac{c'_f}{0.0885\epsilon_r} \right)} \text{ ohms}$$

$$c'_f = \frac{0.0885\epsilon_r}{\pi} \left\{ \frac{2}{1-\frac{t}{b}} \log_e \left(\frac{1}{1-\frac{t}{b}} + 1 \right) - \left(\frac{1}{1-\frac{t}{b}} - 1 \right) \log_e \left(\frac{1}{\left(1-\frac{t}{b}\right)^2} - 1 \right) \right\}$$

ϵ_r = relative dielectric constant (at a given frequency)

b = plate spacing (mil)

w = trace width (mil)

t = trace thickness (mil)

c'_f = fringing capacitance ($\mu\text{f}/\text{cm}$) * assuming semi - infinite plate between two infinite ground planes, but good approximation for $w/(b - t) \geq 0.35$

ATTENUATION IN LOSSY LINES

- ✘ Attenuation per length^[1, EQN 9-54]:

$$\alpha_n = \sqrt{\frac{1}{2} \left[\sqrt{(R_L^2 + \omega^2 L_L^2)(G_L^2 + \omega^2 C_L^2)} - \omega^2 L_L C_L + R_L G_L \right]} \text{ nepers/len gth}$$

- ✘ Using a low-loss approximation^[1, EQN 9-55]:

$$\alpha_n = \frac{1}{2} \left(\frac{R_L}{Z_0} + G_L Z_0 \right) \text{ nepers/len gth}$$

- ✘ But we don't typically discuss in nepers...^[1, EQN 9-57]

$$10^{\frac{\alpha_{dB}}{20}} = e^{\alpha_n} \quad \therefore \alpha_{dB} = 20 \log_{10} e \times \alpha_n$$

for ease of notation : $Y_{n \rightarrow dB} = 20 \log_{10} e$

CONDUCTOR LOSS (per inch)

$$\alpha_{cond} = (Y_{n \rightarrow dB}) \times \frac{R_L}{Z_0}$$

$$\left\{ \begin{array}{l} \alpha_{cond} = \text{attenuation of amplitude due to conductor loss, in dB/length}^{[1, EQN 9-59]} \\ Y_{n \rightarrow dB} = \text{conversion from nepers to dB} \\ R_L = \text{resistance per length of conductor} \\ Z_0 = \text{characteristic impedance} \end{array} \right.$$

✘ Skin effect is accounted for in resistance^{[3, EQN 7]:}

$$R = \frac{l}{w} \sqrt{\pi \mu \rho f} = l \times R_L$$

$$\left\{ \begin{array}{l} l = \text{lenth of trace (inch)} \\ w = \text{width of trace (inch)} \\ \mu = \text{permeability of Cu} \approx 4\pi \times 10^{-7} \times 1 \\ \rho = \text{resistivity of Cu} = \frac{1}{\sigma} = \frac{1}{5.69 \times 10^7} \end{array} \right.$$

✘ Conductor loss per inch as entered in the model:

$$\alpha_{cond} = \frac{1}{2} \times (20 \log_{10} e) \times \left(\sqrt{\frac{\pi (4\pi \times 10^{-7})}{5.69 \times 10^7}} \frac{\sqrt{f_{\text{in Hz}}}}{w_{\text{in mils}}} \right) \times \frac{1}{Z_0}$$

DIELECTRIC LOSS (per inch)

$$\alpha_{diel} = (Y_{n \rightarrow dB}) \times G_L Z_0 \quad \left\{ \begin{array}{l} \alpha_{cond} = \text{attenuation of amplitude due to dielectric loss, in dB/length}^{[1, EQN 9-60]} \\ Y_{n \rightarrow dB} = \text{conversion from nepers to dB} \\ G_L = \text{shunt conductance per length from dielectric} \\ Z_0 = \text{characteristic impedance} \end{array} \right.$$

✘ As developed by Bogatin...

$$\left. \begin{array}{l} G_L = \omega \tan(\delta) C_L \\ Z_0 = \frac{\sqrt{\epsilon_r}}{c C_L} \end{array} \right\} \begin{array}{l} G_L \text{ equation}^{[1, EQN 9-19, EQN 9-60]} \\ Z_0 \text{ equation}^{[1, EQN 9-67]} \text{ is used to cancel the capacitance value,} \\ \text{the } Z_0 \text{ value for a given frequency is NOT used} \\ c = \text{speed of light m/s } \therefore \text{conversion m} \rightarrow \text{in. is needed} \end{array}$$

✘ Dielectric loss per inch as entered in the model:

$$\alpha_{diel} = \frac{1}{2} \times (20 \log_{10} e) \times (2\pi f \times D_f) \times \frac{\sqrt{\epsilon_r}}{299792458 * 39.37}$$

REFERENCES

- [1] E. Bogatin, *Signal Integrity – Simplified*. Pearson Education, Inc., 2004. ISBN 0-13-066946-6
- [2] S. B. Cohn, “Problems in Strop Transmission Lines,” *IRE Trans. Microwave Theory and Techniques*, Vol. MTT-3, March, 1955, pp 119-126.
- [3] H. A. Wheeler, “Formulas for the Skin Effect,” *Proceedings of IRE*, Vol 30, Septempber, 1942, pp 412-424.

