

CAUI-4 Ad hoc

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Agenda

- Patent Policy: The meeting is an official IEEE ad hoc. Please review the patent policy at the following site prior to the meeting.
<http://www.ieee802.org/3/patent.html>
- TBD for CAUI-4 chip-to-chip
 - COM Update [moore_01_073013_cau]
 - Transmitter Eye Mask Measurement
 - Transmit Equalizer
 - Rx Interference

Transmit wave form

Table 83D-1—CAUI-4 transmitter characteristics at 1P0a

| Parameter | Subclause Reference | Value | Units |
|---|---------------------|---------------------------|-------|
| Signaling rate per lane (range) | 83E.3.1.1 | 25.78125 ± 100 ppm | GBd |
| Differential peak-to-peak output voltage (max) (Transmitter disabled) (Transmitter enabled) | 83D.3.1.1 | < 30 1200 | mV |
| Common-mode voltage (max) | 83D.3.1.1 | 1.9 | V |
| Common-mode voltage (min) | 83D.3.1.1 | 0 | V |
| Common-mode AC output voltage (max, RMS) | 83D.3.1.1 | 12 | mV |
| Differential output return loss (min) | 83D.3.1.2 | Equation (83D-2) | dB |
| Common-mode output return loss (min) | 83D.3.1.2 | Equation (83D-3) | dB |
| Transition time (min, 20% to 80%) | 83D.3.1.3 | 8 | ps |
| Output Jitter (max) (Random jitter) (Deterministic jitter) (Total jitter) | 83D.3.1.4 | < 0.15 0.15 0.28 | UI |
| Transmitter eye mask definition X1 | 83D.3.1.5 | 0.14 | UI |
| Transmitter eye mask definition X2 | 83D.3.1.5 | 0.4 | UI |
| Transmitter eye mask definition Y1 | 83D.3.1.5 | 200 | mV |
| Transmitter eye mask definition Y2 | 83D.3.1.5 | 600 | mV |
| Minimum de-emphasis ² (Post Cursor) (Pre Cursor) | 83D.3.1.6 | < TBD TBD | dB |

²Independent of optimal setting used for transmitter jitter and output waveform measurements

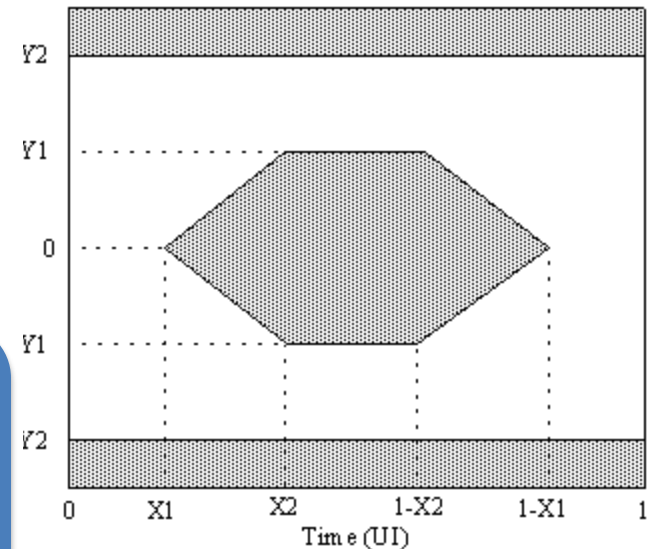


Figure 83D-7—Transmitter eye mask

Measurement of Transmit Jitter

Transmitter output jitter

- The components of the transmitter output jitter is defined in this subclause: total jitter (TJ), deterministic jitter (DJ), and random jitter (RJ).
- Transmitter output is measured at TP0a as defined in . All co-propagating and counter propagating CAUI-4 lanes are active during transmitter output jitter testing. Counter propagating lanes have a target differential peak-to-peak amplitude of 800 mV and transition time of 8 ps. All counter-propagating signals shall be asynchronous to the co-propagating signals using pattern 5 (with or without FEC encoding) pattern 3 or a valid 100GBASE-R signal. For the case of pattern 3, with at least 31 UI delay between the PRBS31 patterns on one lane and any other lane.
- The effect of a single-pole high-pass filter with 3 dB frequency of 10 MHz is applied to the jitter. The test pattern for jitter measurements is PRBS31. The voltage threshold for the measurement of BER or crossing times is the mid-point (0 V) of the AC-coupled differential signal.

Total jitter

- The total jitter (TJ) of a signal is defined as the range (the difference between the lowest and highest values) of sampling times around the signal transitions for which the BER at these sampling times is greater than or equal to 10^{-15} .
- Total jitter for a CAUI-4 chip-to-chip transmitter is than or equal to 0.28 UI and is measured with optimal transmit equalizer setting.

Deterministic and random jitter

- The random jitter (RJ) of a signal is defined to be the difference between the TJ and DJ. DJ is derived from measured jitter distribution as follows
 - Measure the jitter J_n which is defined to be the interval that includes all but 10^{-n} of the jitter distribution. If measured by plotting BER vs. decision time, it is the time interval between the two points with BER of $10^{-n}/4$. Measure two values: J_9 and J_5
 - For each J_n determine the associated Q_n from the inverse normal cumulative probability distribution adjusted for an assumed transition density of 0.5. Q_9 is 5.998 and Q_5 is 4.265
 - Calculate the effective DJ as shown in
- Deterministic jitter is less than or equal to 0.15 UI. Random jitter is less than or equal to 0.15 UI.

Measurement Methodology for Eye Mask – Leverage 83E

Transmitter output waveform

- The eye mask shown in is defined at a BER of 10^{-15} , using the methodology described in TBD. Transmitter equalizer may be adjusted for optimum mask results.

TBD section:

- Transmit output waveform in CAUI-4 chip-to-chip are measured using a fourth-order Bessel-Thomson low-pass filter response with 33 GHz 3 dB bandwidth. X1 is measured using the methodology described in 83D.3.1.4 where $X1 = T/2$. Y2 is measured using the methodology described in 83D.3.1.1. Y1 is measured using the following procedure using a PRBS9 waveform:
 - Capture the PRBS9 pattern using a clock recovery unit with 3 dB tracking bandwidth of 10 MHz and maximum peaking of 0.1 dB and a minimum sampling rate of 3 samples per bit. Collect sufficient samples equivalent to at least 4 million bits to allow for construction of a normalized cumulative distribution function (CDF) to a probability of 10^{-6} without extrapolation.
 - Use the differential signal from step 1 to construct the CDF of the signal amplitude at X2, midpoint (0.5) and 1-X2, for both logic 1 (CDF1) and logic 0 (CDF0), as a distance from the center of the eye. Calculate the eye height for each point (EH6) as the difference in amplitude between CDF1 and CDF0 with a value of 10^{-6} . CDF0 and CDF1 are calculated as the cumulative sum of histograms of the amplitude at the top and bottom of the eye normalized by the total number of sampled bits. For a well balanced number of ones and zeros the maximum value for CDF0 and CDF1 will be 0.5.
 - Apply the Dual-Dirac and tail fitting techniques to CDF1 and CDF0 to estimate the noise at X2, 0.5 and 1-X2 points. Calculate the best linear fit in Q-scale over the range of probabilities of 10^{-4} and 10^{-6} of CDF1 and CDF0 to yield relative noise one (RN1) and relative noise zero (RN0). The eye height for plotting against the eye mask at X2, 0.5, and 1-X2 is then given by
 - $EH15 = EH6 - 3.19 \times (RN0 + RN)$
- where
 - *EH15 peak* is the eye height extrapolated to 10^{-15} probability
 - *EH6* is the eye height at 10^{-6} probability
 - *RN1* is the RMS value of the noise estimated from CDF1
 - *RN0* is the RMS value of the noise estimated from CDF0

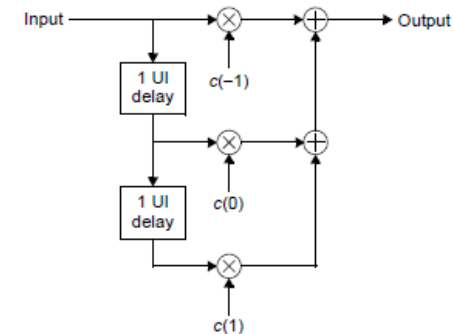
Transmit equalizer

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| Minimum de-emphasis ² (Post Cursor) (Pre Cursor) | 83D.3.1.6 | < TBD TBD | dB |

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- **Transmitter de-emphasis range**
- The CAUI-4 chip-to-chip transmitter includes programmable equalization to compensate for the frequency-dependent loss of the channel and to facilitate data recovery at the receiver. The functional model for the transmit equalizer is the three tap transversal filter shown in TBD. The minimum pre cursor equalization ($c(-1)$) is TBD. The minimum post cursor equalization ($c(1)$) is TBD.
- The transmitter output equalization is characterized using the procedure described in TBD.



Transmit equalization

- ghiasi_010713: post cursor of 6dB in 0.5dB steps and pre is nice to have with 3dB in 0.5dB steps
- OIF MR

14.3.1.6 Transmitter output waveform requirements

The transmitter shall include an equalizer defined as:

$$H(Z) = C_{-1} + C_0 z^{-1} + C_1 z^{-2} \quad (14-10)$$

14.3.1.6.1 Summary of requirements

The normalized amplitudes of the coefficients of the transmitter equalizer (computed per 14.3.1.6.2) shall meet the requirements in [Table 14-9](#).

Table 14-9. Coefficient range and step size

| Coefficient | Normalized Amplitude | | Normalized Step Size (%) |
|-------------|----------------------|---------|--------------------------|
| | Min (%) | Max (%) | |
| C_{-1} | -20 | 0 | 1.25 to 5 |
| C_1 | -25 | 0 | 1.25 to 5 |
| C_0 | 40 | 100 | 1.25 to 5 |

Transmit equalizer characterization (In CL93)

- The transmitter output waveform is characterized using the procedure described in 85.8.3.3. The parameters of the linear pulse fit and equalizer are summarized in Table 93–5.

Table 93–5—Linear fit pulse and equalizer parameters

| Description | Symbol | Value | Units |
|-------------------------|--------|-------|-------|
| Linear fit pulse length | N_p | 8 | UI |
| Linear fit pulse delay | D_p | 2 | UI |
| Equalizer length | N_w | 8 | UI |
| Equalizer delay | D_w | 2 | UI |

Steady-state voltage and linear fit pulse peak

The steady-state voltage v_f is defined to be the sum of the linear fit pulse $p(k)$ divided by M (refer to 85.8.3.3 step 3). The steady-state voltage shall be greater than or equal to 0.4 V and less than or equal to 0.6 V after the transmit equalizer coefficients have been set to the values used for transmit wave form evaluation. The peak value of $p(k)$ shall be greater than $0.8 \times v_f$ after the transmit equalizer coefficients have been set to the values used for transmit wave form evaluation .

Linear fit error

For any configuration of the transmit equalizer, the RMS value of the error between the linear fit and the measured waveform, $e(k)$, normalized to the peak value of the linear fit pulse, $p(k)$, shall be less than or equal to 0.037.