

# Burst Markers in EPoC

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# Introduction

- Burst markers are used to indicate the start and end of each burst in EPoC burst mode
  - The same marker delimits the start and the end of a burst
- A burst marker is a sequence of known modulated symbols (in the frequency domain).
- To indicate the profile of a burst, a different marker can be used for each profile.
  - E.g., assuming the use of a ternary alphabet for the markers

Profile	Marker (8 symbols)
0	{+1, 0,-1,-1, 0, 0,+1,+1}
1	{ 0,-1, 0, 0,+1,+1, 0, 0}

# Upstream Transmission

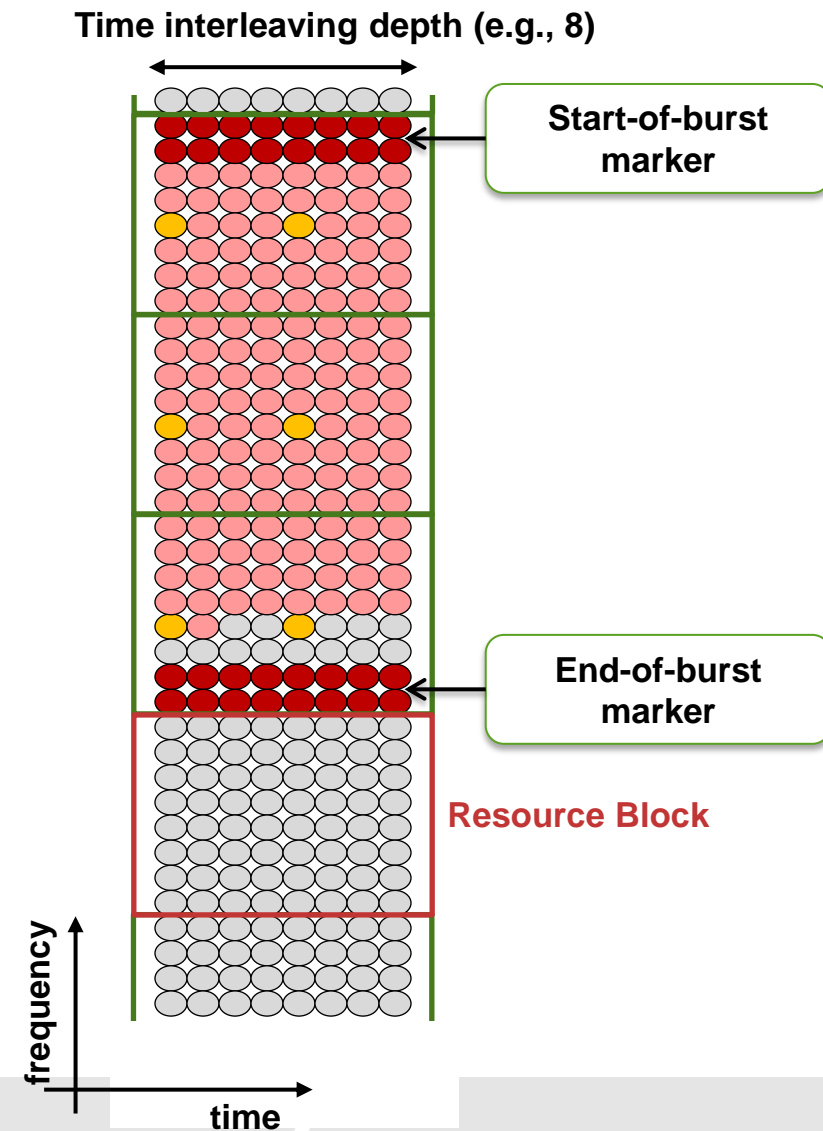
# Upstream Burst Markers

Features specific to Upstream transmission:

- Each CNU performs **pre-equalization**
  - Channel is pre-compensated at transmitter's side (to some extent)
- Each burst is interleaved in Time and Frequency **independently**
  - Different bursts correspond to independent grants
- Time/Frequency resources are divided into **Resource Blocks (RB)**
  - A particular grant spans an integer number of RBs
- Association of pilot tones to a specific burst is **unknown until the burst start/end have been detected**
- Transmissions from different CNUs do not follow back-to-back
  - Need for markers:
    - 1) At the start of each burst
    - 2) At the end of each burst

# Upstream Receiver Processing

- The output of the FFT is buffered in blocks equal to the time interleaving depth
- Each burst occupies an integer number of RBs
- The marker for the specific burst profile is placed **at the start of the first RB and at the end of the last occupied RB**. In this example:
  - ● is a marker symbol (e.g.,  $\{-1, 0, +1\}$ )
  - ● is a data symbol (e.g., 1024QAM)
  - ● is a pilot symbol (e.g., QPSK)
  - ○ is an unused resource element
- **Markers and pilots are independent and are located at predictable locations**

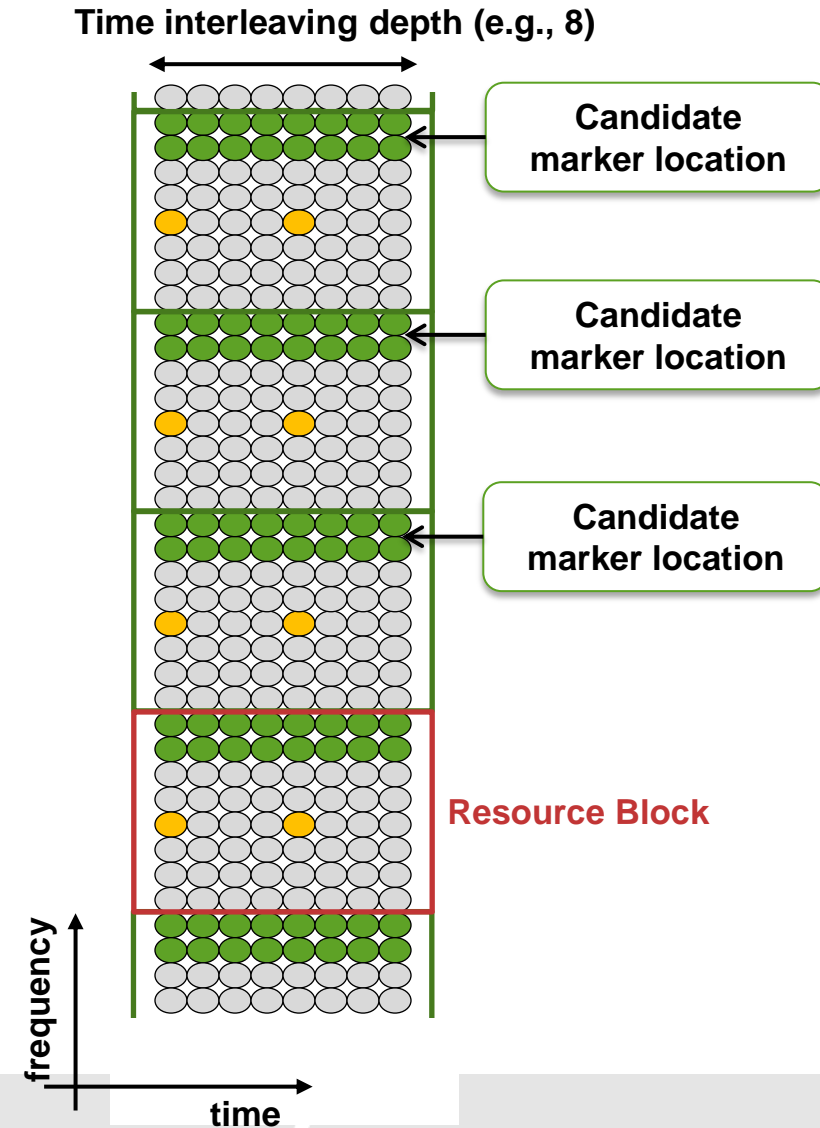


# Upstream Receiver Processing

- Marker detection **follows FFT and buffering**

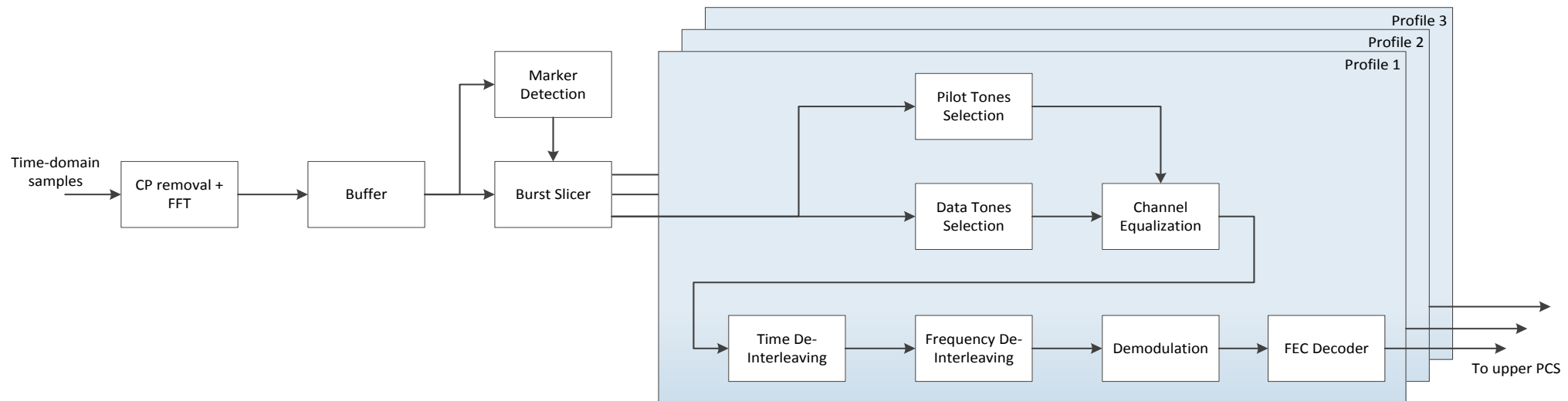


- Marker searcher scans all **candidate marker locations**
- Marker detection needs to be **resilient to residual channel distortion** (not pre-equalized)
  - Phase may change in time (CNU LO instability)
  - Channel response may change in frequency (e.g., RF front-end response varies with temperature)



# Overall Upstream Receiver Processing

- Marker search is performed right **after FFT (possibly jointly with buffering of incoming samples)**
- Channel estimation and equalization, Frequency and Time de-interleaving are performed **separately for each burst, after burst slicing has occurred**



# Downstream Transmission

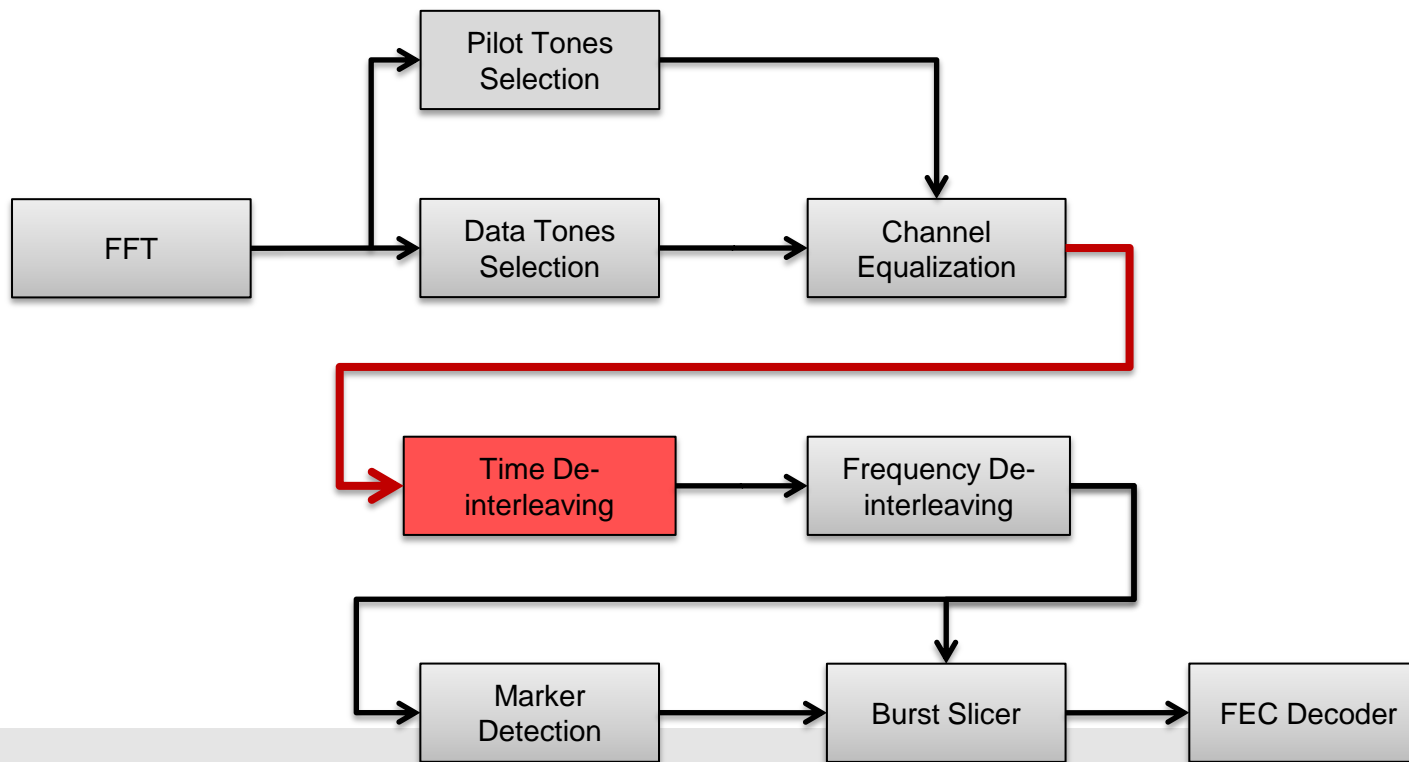


# Downstream Burst Markers

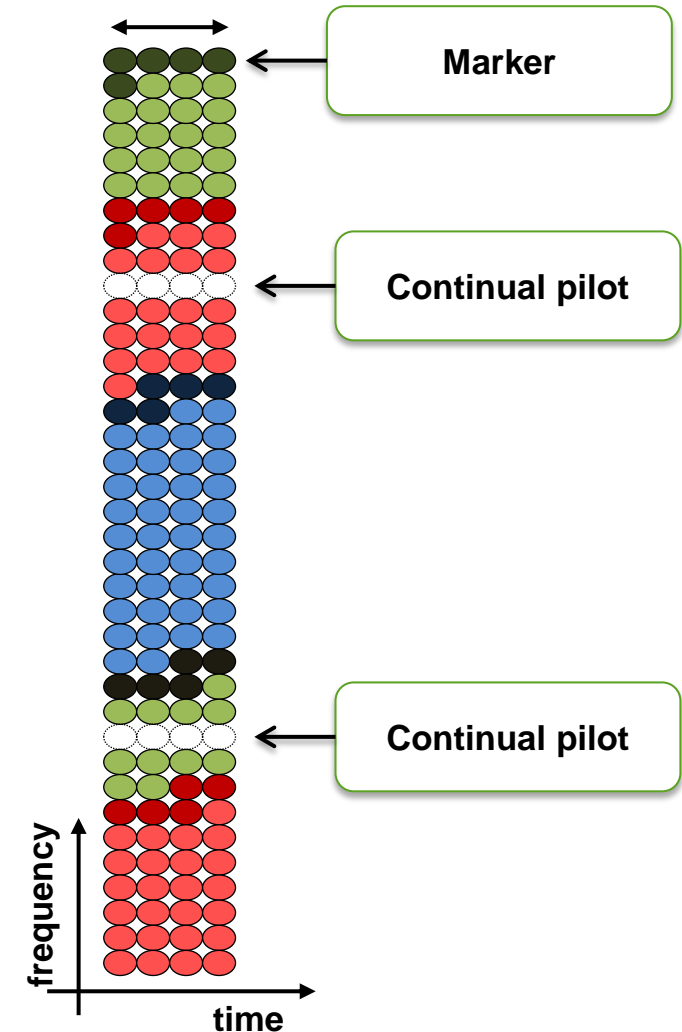
- **In the DS case, transmission is continuous.**
  - One profile follows the previous back-to-back (no notion of Resource Block)
  - Only a **single start marker** for each profile is strictly needed
- **EPoC DS is a system with a **single transmitter****
  - Pilot positions are **known to CNU**s
  - The channel response is estimated **over the full bandwidth** by each CNU **using all available pilots** in order to achieve the largest possible processing gain
  - This implies that **channel equalization is performed before marker detection**
- **Marker detection comes **after frequency and time de-interleaving** (see following slides)**
  - Time and frequency interleaving are **common for all profiles**

# Downstream Receiver Processing

- Time and Frequency de-interleaving is a **reordering** operation applied to modulated symbols (in this figure frequency interleaving is neglected for simplicity)
- Channel estimation and equalization, Frequency and Time de-interleaving are performed after the FFT and are **burst-agnostic**
- Marker search is performed last, **before demodulation and FEC decoding**



Time interleaving depth (e.g., 4)



# Marker Sequence Design

# Proposed burst marker scheme

- In the proposed scheme, a burst marker is a sequence from a ternary alphabet  $\{-1, 0, +1\}$
- In **particular**, a burst marker is formed by combining a fixed pattern of **non-zero symbols**  $P=+1, -1$ , and null (silent) **symbols**,  $N=0$ , interlaced with each other.
- Locations of  $P$  are mutually orthogonal in all burst marker sequences.
- Example of 4 burst marker sequence to represent 4 profiles:

Burst marker sequence #1 =  $P, P, N, N, N, N, N, N$

Burst marker sequence #2 =  $N, N, P, P, N, N, N, N$

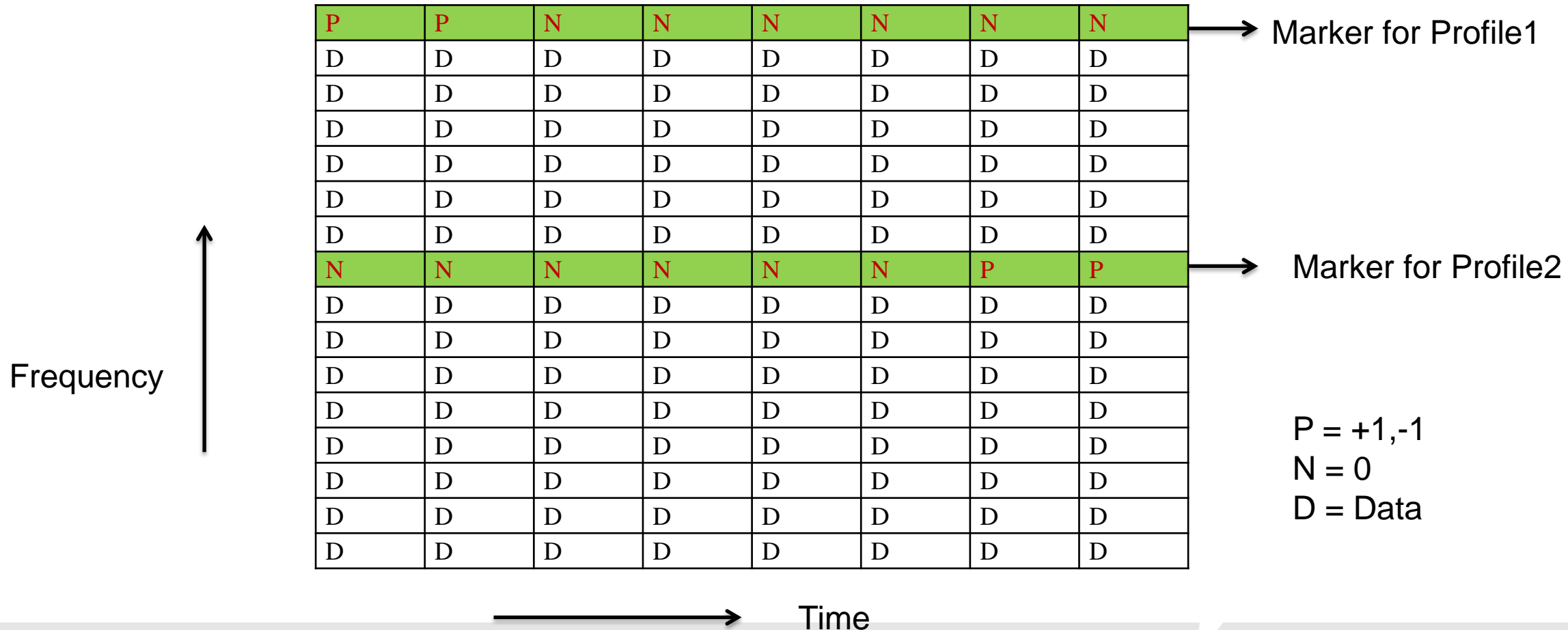
Burst marker sequence #3 =  $N, N, N, N, P, P, N, N$

Burst marker sequence #4 =  $N, N, N, N, N, N, P, P$

# Proposed burst marker scheme example

## Horizontal marker placement

- Minimizes distortion due to channel response changes (not pre-equalized)



# Burst marker detector

- In case of 'x' dB signal to noise ratio, the power of a non-zero symbol is 'x' dB higher than the power of a null symbol.
- This difference in power levels between the non-zero symbols and the null symbols is exploited by the receiver for detection of the correct burst marker sequence in a non-coherent fashion (energy-based), therefore inherently robust to phase distortion due to, e.g., residual carrier frequency offset.

# Burst marker detector

- For the case of 4 profiles, the detector computes 4 power ratios, one for each hypothesis.
- The numerator of the power ratio is the sum of powers of two symbols at the locations of the pilots.
- The denominator of the power ratio is the sum of powers of the 4 symbols, at the location of the nulls.
- This non-coherent detector is compared to a correlation-based detection method.

# Performance Target

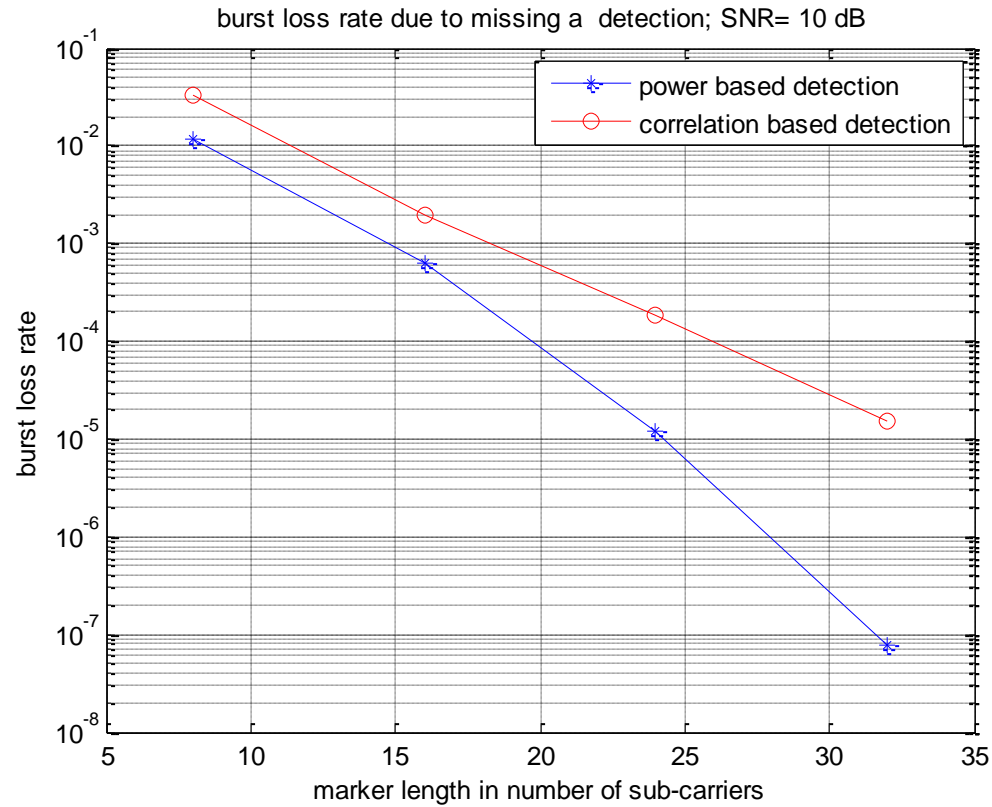
- Packet loss in EPoC will be due to two factors:
  1. Packet loss due to  $1E-8$  BER requirement
  2. Packet loss due to burst detection error.
- Two sources of burst detection error.
  1. Missed detection of a burst.
  2. False detection of burst. This causes dropping of the existing burst.
- Burst loss rate should include both the false detection and missing detection.
- EPoC target packet loss rate is  $1E-5$  for US and  $1E-6$  for DS.
- Cable labs have similar requirements.
- Since a burst has multiple packets, the burst loss rate should be much smaller than packet loss rate.



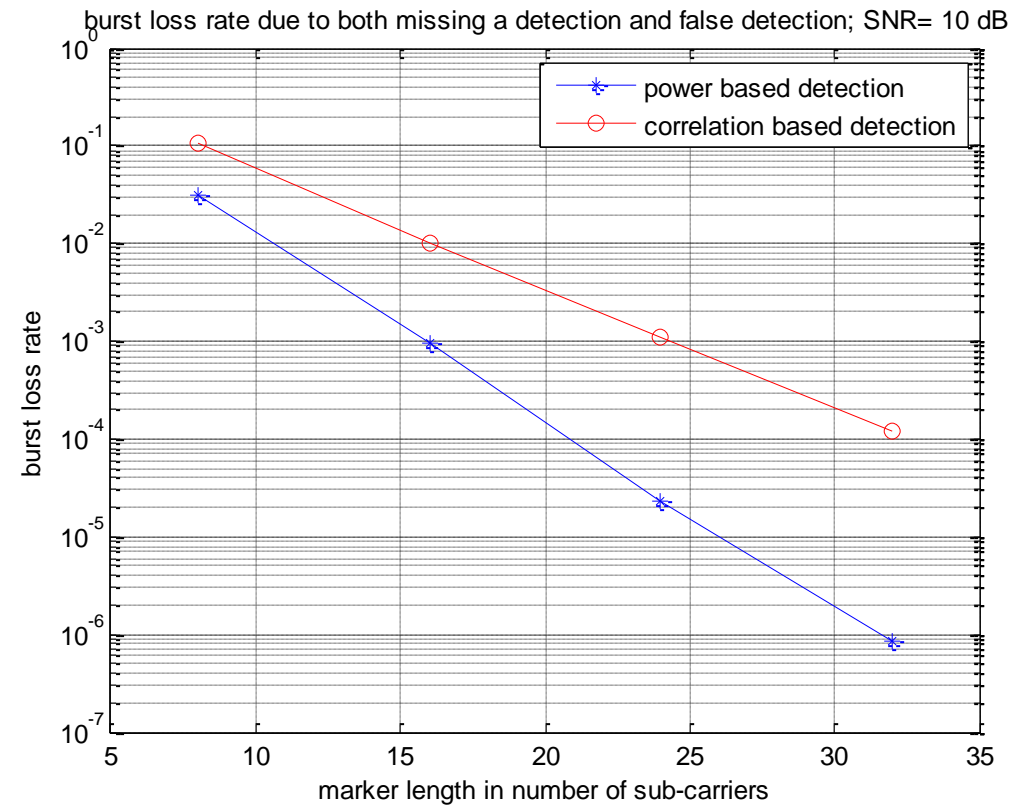
# Simulation Results:

- Upstream transmission
- AWGN
- DOCSIS Multipath profile (Valid 97% of the time):  
delay = [0.5 1 1.5 2 3 4.5]\*1e-6;  
pwr\_db = [-16 -22 -29 -35 -42 -51];
- Residual carrier frequency offset = 25 Hz
- Data modulation order: 1024QAM
- In the plots a missing data point indicates 100% detection.

# Multipath, SNR=10dB

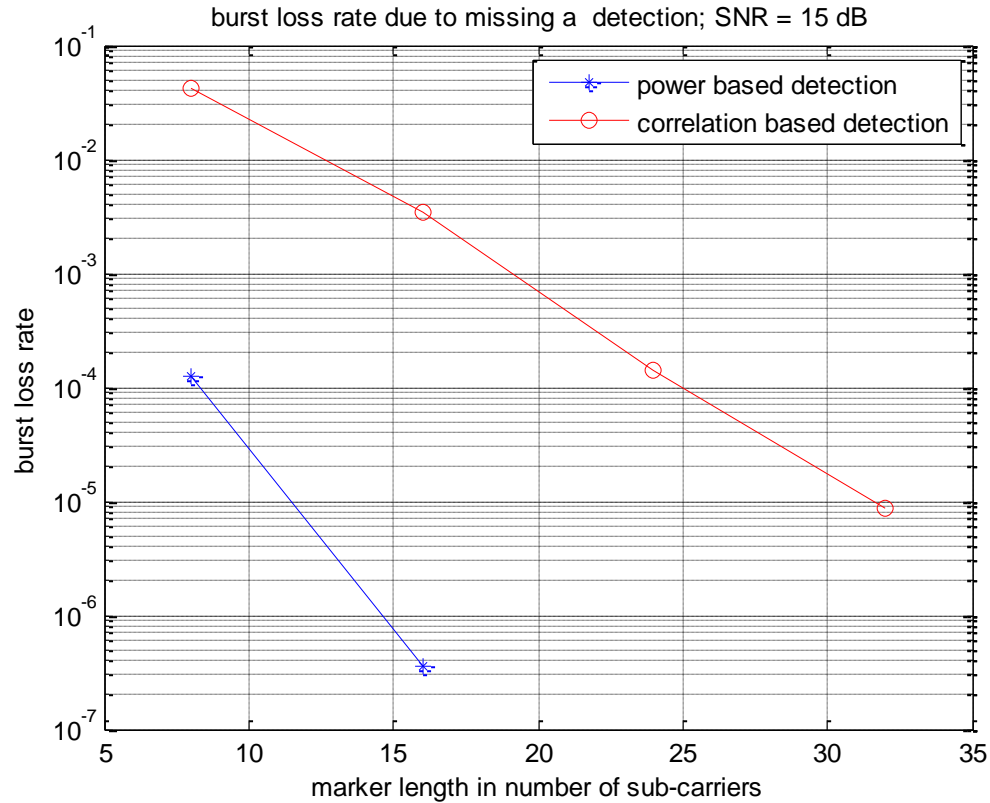


Plot 1A

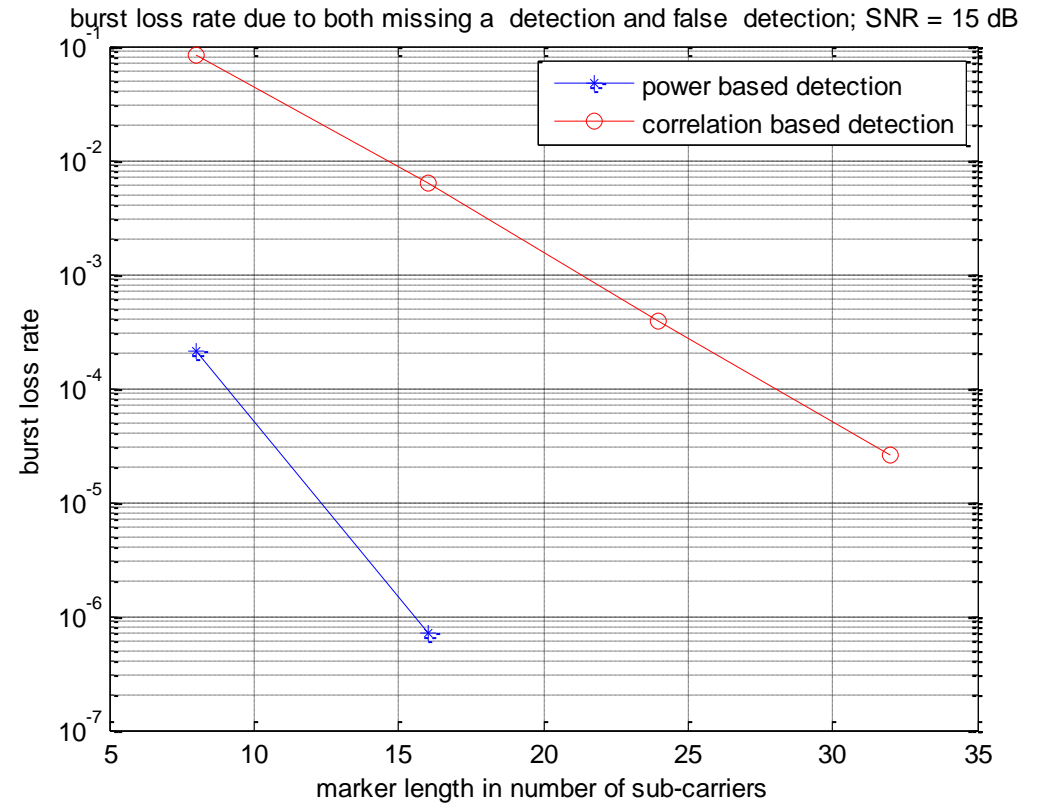


Plot 1B

# Multipath, SNR=15dB

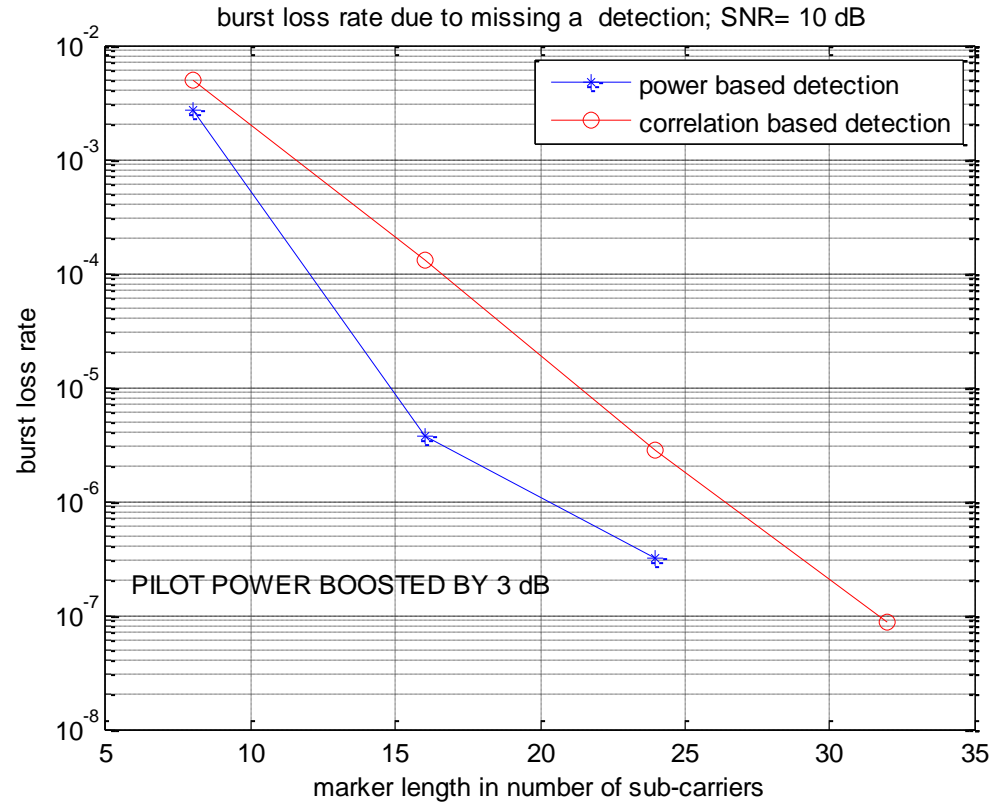


Plot 2A

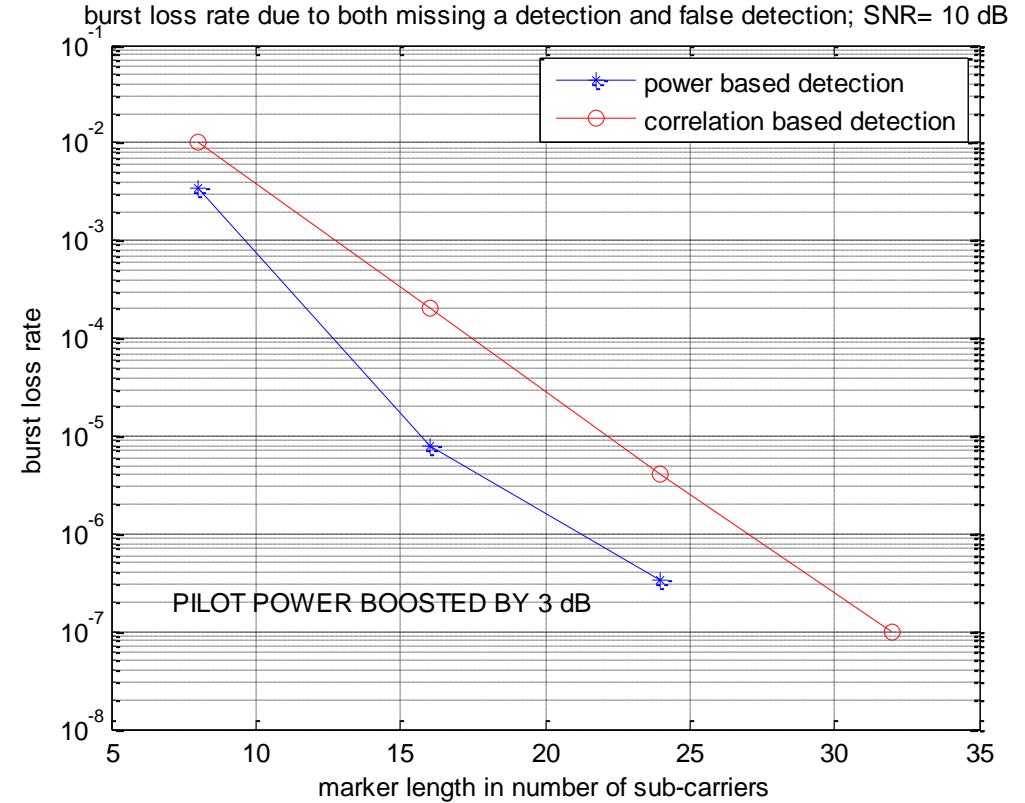


Plot 2B

# Multipath, SNR=10dB, 3dB power boost

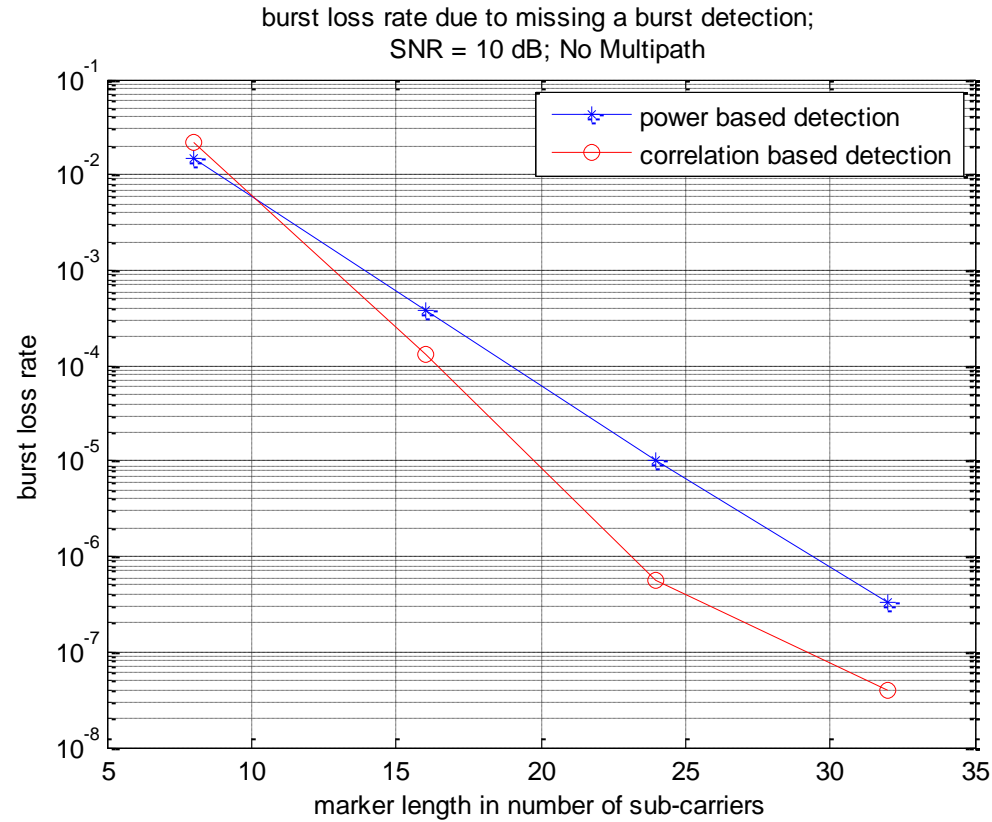


Plot 3A

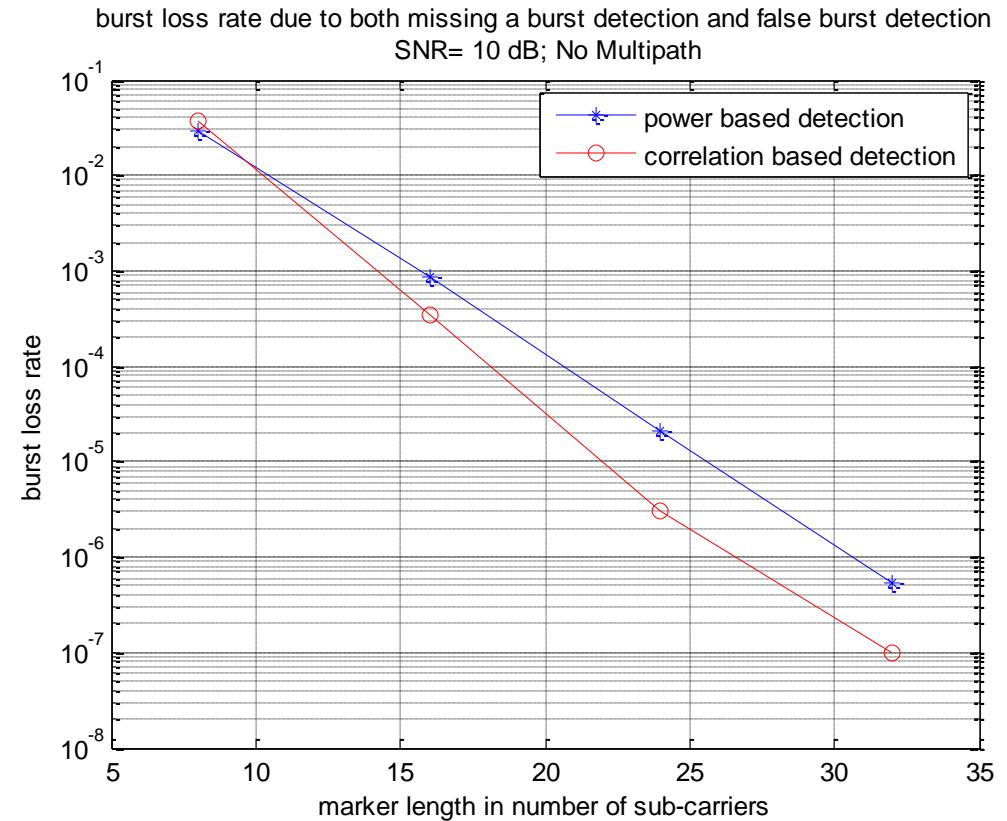


Plot 3B

# No Multipath, SNR=10dB

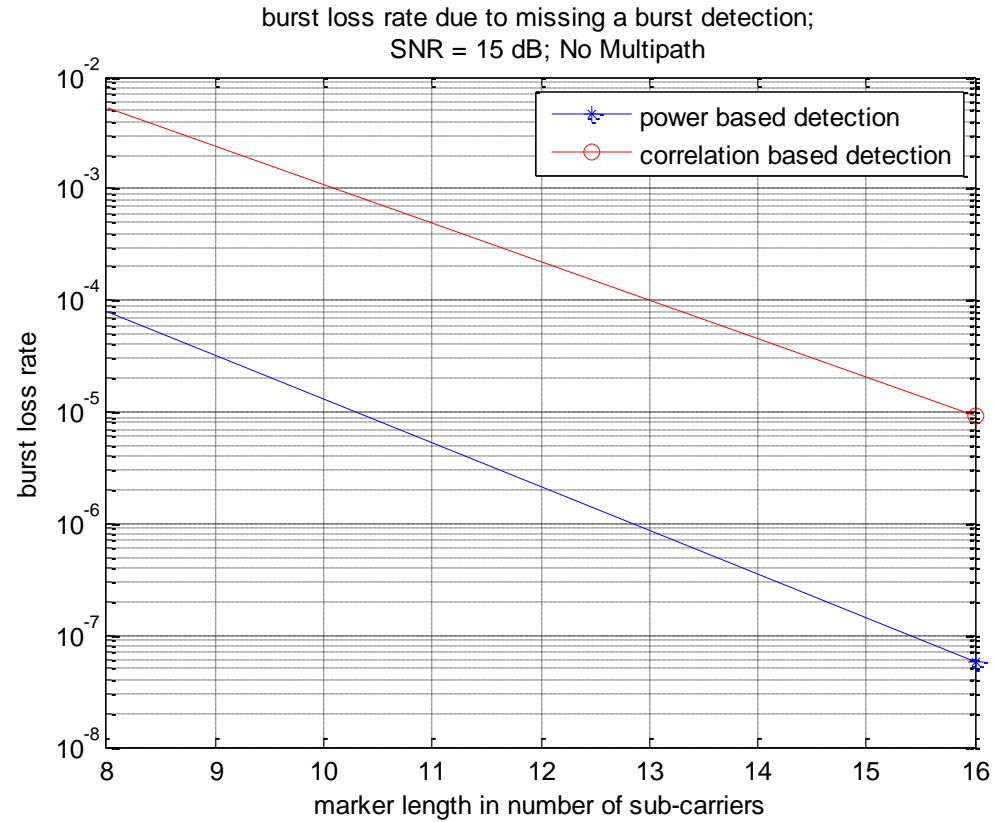


Plot 4A

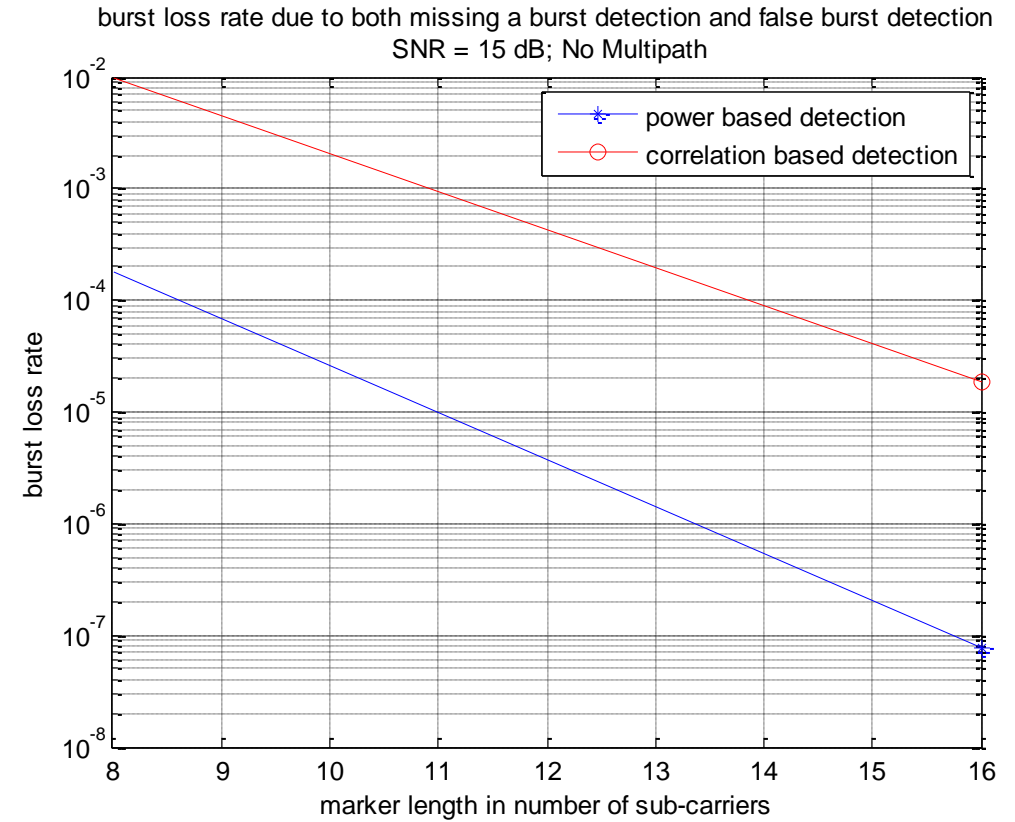


Plot 4B

# No Multipath, SNR=15dB



Plot 5A



Plot 5B

# Conclusions

- The energy based detection scheme is superior to the correlation based scheme in the following cases:
  - If pre-equalization is not performed (multipath), it is superior in all simulated cases.
  - If pre-equalization is performed (no multipath), it is superior for higher SNR values ( $> 10$  dB).
- The correlation based detection scheme is superior to the energy based scheme in following cases:
  - If pre-equalization is performed (no multipath) and for low SNR values ( $10$  dB).
- Boosting the power of marker symbols by 3 dB, improves the performance by an order of magnitude.

# Conclusions

- More simulations are needed with inclusion of pre-equalization. Which will eliminate the multipath effect to a large extent, but not completely.
- More research is needed to explore the possibility of combining both the schemes to improve performance.
- The following parameters need to be specified
  - Targeted total burst loss rate ( due to both missed detection and false detection), for both Upstream and Downstream.
  - Lowest supported QAM modulation order for data (Upstream and Downstream). This sets the minimum SNR for evaluating the burst marker detection schemes.



Thank You