# TRANSMIT EQUALIZER STEP SIZE SPECIFICATIONS (COMMENTS \#62, \#63, \#74, \#10249) 

Adee Ran, Intel

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## Background

- Tx equalization maximum step size specification was $5 \%$ in 50 G electrical PMDs (clauses 136, 137, also annex 120D)
- c(-2) was specified as 2.5\%.
- In 802.3ck:
- Following hidaka 3ck adhoc 01120518 and sun 3ck adhoc 01a 120518 all analysis assumed a $2 \%$ step size for $c(-3)$ through $c(0)$, and this value was included in the baseline proposal heck 3ck 03b 0319.
- $5 \%$ for $\mathrm{c}(+1)$
- The 2\% step size can create an additional burden on DAC-based transmitters. Power impact estimated as $\sim 0.5 \mathrm{pJ} / \mathrm{bit}$.
- In ran 3ck adhoc 01021920 we have shown that step size has small and very irregular effect on COM results.
- Comments \#62, \#63, \#74, \#10249 against D1.1 address Tx equalization step sizes.


## Goals of this presentation

- In ran 3ck adhoc 01021920 it was stated that "Moving from 2.5\% to 2\% requires an additional DAC bit, otherwise some steps will have no measurable effect."
- Feedback received suggested that the additional bit may be required only in digital calculations, and not necessarily in the DAC, by rounding the calculated FFE output to 7 bits.
- The claim about "no measurable effect" was indeed incorrect.
- Rounding will be discussed below (spoiler: possible, but with increased Tx noise).
- Other comments suggest that having a 5\% step size for c(+1) alone does not benefit Tx design and can create unexpected complexity for optimization algorithms.
- This will be explained.


## Possible designs choices

To meet a $2.5 \%$ step size specification

- 7-bit integer 2-tap FFE calculation can work as follows:
- Input is $\{-3,-1,+1,+3\}$
- Multipliers are 0:0.5:21 (42 values) for $c(0)$, and -5:0.5:0 (11 values) for c(-1)
- Normalized step size is $1 / 42=\mathbf{2 . 3 8 \%}$
- Output range is $21 * 3-21 *(-3)=126$
- Output is shifted to an unsigned range of 0 to 126 (so the value 63 corresponds to zero differential output)

To meet a $2 \%$ step size specification

- 8-bit integer FFE calculation is required:
- Input is $\{-3,-1,+1,+3\}$
- Multipliers are 0:0.5:42.5 (85 values) for $\mathrm{c}(0)$, and -10:0.5:0 (21 values) for c(-1)
- Normalized step size is $1 / 85=1.18 \%$
- Output range is $42.5^{*} 3-42.5^{*}(-3)=255$
- Output is shifted to an unsigned range of 0 to 255 (so the value 127.5 corresponds to zero differential output)


## Results of 7-bit design

- Outputs for different coefficient combinations:

| $c(-1)$ | $c(0)$ | NRZ outputs | PAM4 outputs |
| :--- | :--- | :--- | :--- |
| 0 | 21 | $0 ; 126$ | $0 ; 42 ; 84 ; 126$ |
| -0.5 | 20.5 | 0,$3 ; 123,126$ | $0,1,2,3 ; 41,42,43,44 ; 82,83,84,85 ; 123,124,125,126$ |
| -2.5 | 18.5 | 0,$15 ; 111,126$ | $0,5,10,15 ; 37,42,47,52 ; 74,79,84,89 ; 111,116,121,126$ |

(Unnormalized values)


## Results of 8-bit design

- Outputs for different coefficient combinations:

| $c(-1)$ | $c(0)$ | NRZ outputs | PAM4 outputs |
| :--- | :--- | :--- | :--- |
| 0 | 42.5 | $0 ; 255$ | $0 ; 85 ; 170 ; 255$ |
| -0.5 | 42 | 0,$3 ; 252,255$ | $0,1,2,3 ; 84,85,86,87 ; 168,169,170,171 ; 252,253,254,255$ |
| -5 | 37.5 | 0,$30 ; 225,255$ | $0,10,20,30 ; 75,85,95,105 ; 150,160,170,180 ; 225,235,245,255$ |

(Unnormalized values)



## What if output DAC is 7 bits?

## With 7-bit calculation

- FFE calculation is fed directly to DAC
- Pure linear system, no additive noise
- Equalization control is more coarse than with 8 bits
- But, as we have shown, with the Rx adaptive equalization the result may actually be better


## With 8-bit calculation

- Outputs have to be divided by 2
- Problem: some outputs are even, some are odd
- Truncation error is either 0 or 1 LSB depending on input sequence $\rightarrow$ additive quantization noise
- With RMS $=\frac{1}{\sqrt{2}}$ LSB, effect on SNDR is small - but this quantization noise can't be mitigated by the Rx
- More refined equalization control is not necessarily beneficial
- More expensive digital calculations


## What about $\mathrm{c}(+1)$ ?

- If the max step size is $>2 x$ larger than the rest, implementations may actually apply double steps
- This creates complications for receivers trying to optimize Tx equalization.
- Suppose the receiver wants to sweep possible values of $c(+1)$ starting from preset 1:
- Prior to decrementing $\mathrm{c}(1), \mathrm{c}(0)$ must be decremented, to prevent getting an "equalization limit" response
- In the Tx (unlike COM calculation) c(0) is not automatically determined from other coefficients
- If step sizes are the same, one decrement of $c(+1)$ requires one decrement of $c(0)$
- If $c(1)$ has $2 x$ step size, one decrement of $c(+1)$ requires two decrements of $c(0)$.


## The problem with unequal step sizes

- The Rx has no way to tell what step sizes the Tx has.
- Uncertainty exists regardless of the "search" algorithm chosen.
- Step sizes can vary even more...
- Planning for all possible combinations is difficult; validation is a nightmare.
- There is no real design benefit for having unequal step sizes.
- This ambiguity also exists with uniform step size limits...
- It was not realized until we had different limits in 802.3cd.
- We should add a recommendation to have nominally equal step sizes for all coefficients, to enable simple "step counting" logic.


## Summary

- Current max step size spec of 2\% is too aggressive
- For a digital implementation, requires at least 8-bit calculations, if not 8-bit DAC
- Changing to max $2.5 \%$ would enable full 7 -bit design with negligible impact (if any) on Rx
- Finer steps have no real benefit, and cost power
- COM grid is not necessarily related, but run time can be reduced by changing $2 \%$ to $2.5 \%$
- Allowing $c(+1)$ to have larger steps creates unexpected complexity in Rx optimization - COM grid is not related; can stay with a larger step to reduce run time
- Recommended changes in D1.1 $\rightarrow$ D1.2:
- In transmitter characteristics
- Use uniform step size specs for all taps, with absolute step min: 0.005 and max: 0.025. Apply in subclause text and Tx summary table.
- Add a recommendation in the subclause text: "The step sizes of all coefficients should be nominally equal, to enable efficient scanning of the coefficient space."
- With editorial license
- In COM
- Change search step to $2.5 \%$ for all precursor taps.
- Apply the above for clause 162, clause 163, and annex 120G.

BACKUP

Results



In both cases, COM vs. step size trend is very small in all channels
Effect of $2 \%$ to $2.5 \%$ is between $\sim 0.05 \mathrm{~dB}$ (for low COM channels) and 0.13 dB (for the high COM channel)
Results are very "noisy" and inconclusive even at relatively large steps ( $\mathrm{R}^{2}$ maximum value was only $\sim 0.75$; most were much worse)

## What was the $2 \%$ recommendation based on?

## TX Resolution Impact

DFE vs [CM]DFE


Source: sun 3ck adhoc 01a 120518 Slide 8

FFE vs [CM]FFE

$>2.5 \%$ (CDFE and CFFE) are often much worse than 1.5\% (DFE and FFE)
$>2.0 \%$ (MDFE and MFFE) are close to $1.5 \%$ (DFE and FFE)
CrędO゙

## Digging into the data

Full data set provided in hidaka 3ck adhoc 02120518 to enable further analysis

Same, excluding the "AZ1" and "AZ2" Data CDFE0.85
 zambell_100GEL_01a_0318.pdf zambell_3ck_01_1118.pdf

Coarse DFE (0.25\%) vs. medium DFE (0.2\%)


| 16 | AZ1 | Orthogonal Backplane Channel |
| :---: | :---: | :---: |
| $\mid 89-115$ | AZ2 | Measured Orthogonal Backplane with Varied Impedances |

Source: sun 3ck adhoc 01a 120518 slide 4

## Eventually we chose a subset of channels for analysis

## The Highlighted Channels

| Contribution | Channel |  |
| :---: | :---: | :---: |
| heck 3ck 011118 | 28dB Cabled Backplane/Cable_BKP_28dB_Op575m_more_isi | "AZ" channels not in the list |
|  | 16dB Cabled Backplane/Cable_BKP_16dB_Op575m_more_isi |  |
| mellitz 3ck adhoc 02081518 | 24,28,30dB including BGA Via/CaBP_BGAVia_Opt2_28dB |  |
| tracy 3ck 010119 | Traditional Backplane Channels/Std_BP_12inch_Meg7 |  |
|  | Orthogonal Backplane Channels/DPO_IL_12dB |  |
| kareti 3ck 01a 1118 | Measured Orthogonal Backplane Channels/OAch4 |  |
|  | Measured Orthogonal Backplane Channels/Och4 |  |
|  | Measured Cabled Backplane Channels/CAch3_b2 |  |
|  | Measured Traditional Backplane Channels/Bch2_a7p5_7 |  |

Source: kochuparambil 3ck 01c 0119 slide 5

## Tap Values By Channel

## From ran 3ck adhoc 01021920






## Tap Values By Channel



31/29mm Tx/Rx Package


