TDECQ, slowness, badness and overshoot

Piers Dawe Mellanox Technologies

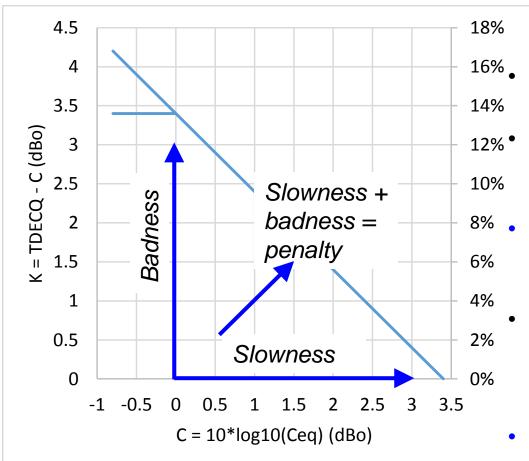
Supporters

- Marco Mazzini Cisco
- Johan Jacob Mohr Mellanox

Introduction

- In the TDECQ method, there are several specs on a signal
 - Four limits in the in-force 802.3cd and 802.3cn
- We have four limits to protect the different parts of the receiver, and the link quality, against different threats
- Another limit, against overshoot, is in IEEE P802.3cu draft 2.0, but one of the four has been partly removed
- This presentation addresses their different characteristics and uses, showing that all four, and probably overshoot too, are separate and needed
- Also presents a way of measuring overshoot

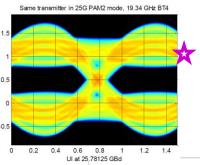
Cost of correcting + what's not corrected = penalty Slowness + badness = penalty



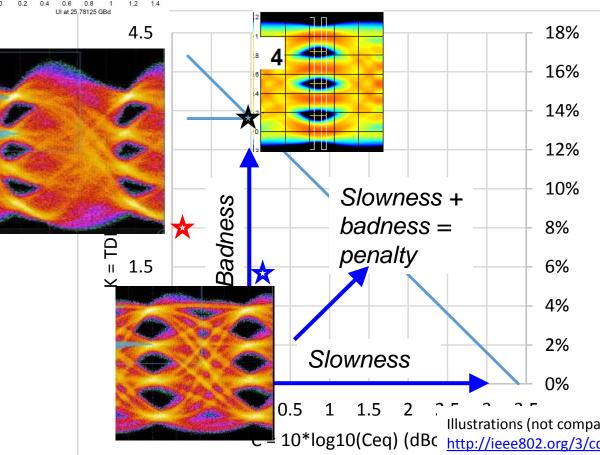
- in dawe 3cd 01a 0318 (nearly two years ago)
- We have used different names over time, e.g.
- Slowness penalty, noise enhancement [penalty] or **C** for "slowness":
- C is the cost of correcting what the reference equalizer can correct
- roughness and signal's noise penalties, residual ISI and noise penalties, residual eye penalty, or K for "badness";
- K is what isn't corrected

See:

Transmitter and Dispersion Eye Closure Quaternary (TDECQ) and Its Sensitivity to Impairments in PAM4 Waveforms Santiago Echeverri-Chacón, Johan Jacob Mohr, Juan José Vegas Olmos, Piers Dawe, Bjarke Vad Pedersen, Thorkild Franck, and Steen Bak Christensen, Journal of Lightwave Technology, **37**(3), pp 852-860, February 1, 2019



Examples (as before equalizer)



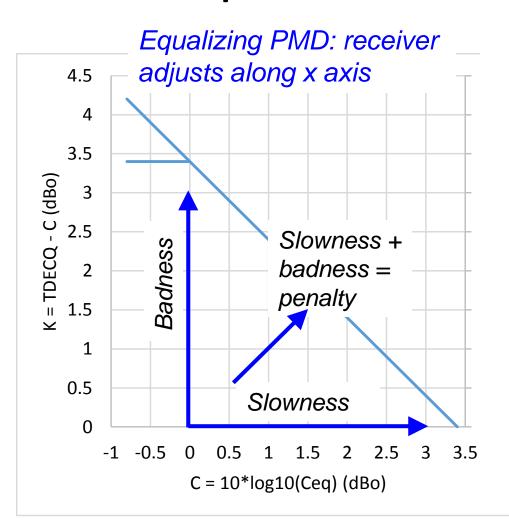
Alternative
—— limits, same
measurement

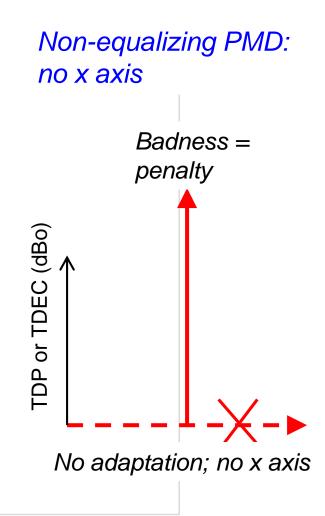
- The "four limits" on slide 2 are left, top, diagonal-right and right
- This presentation concentrates on two of them, as shown in blue

Illustrations (not comparable with each other) from

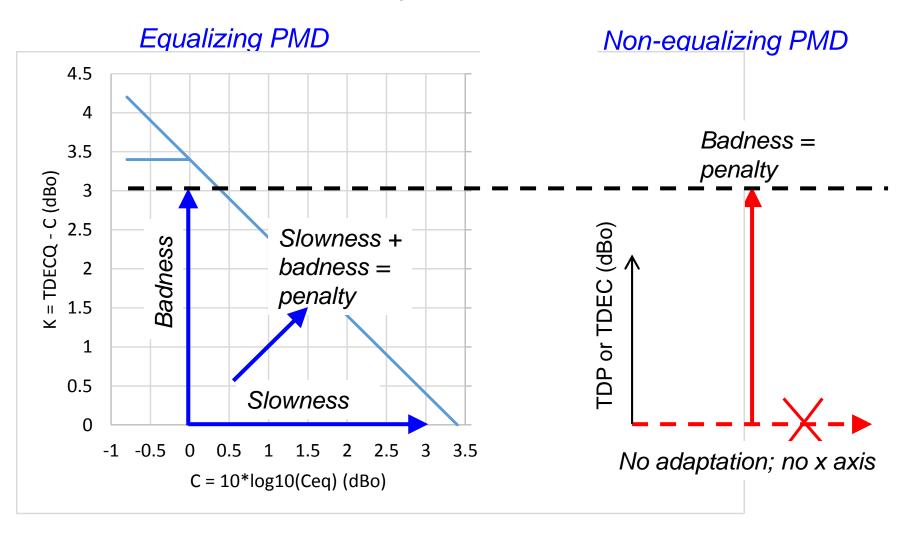
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Compare a non-equalized PMD



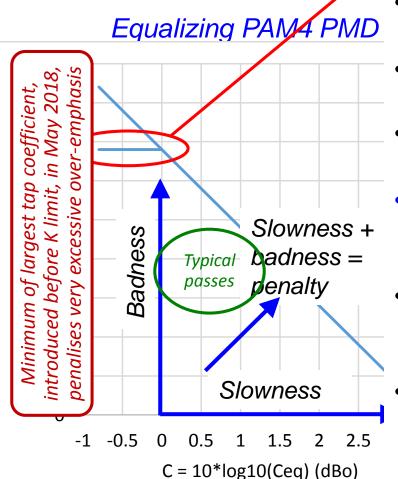


We always limit badness



All in-force optical PAM4 is done

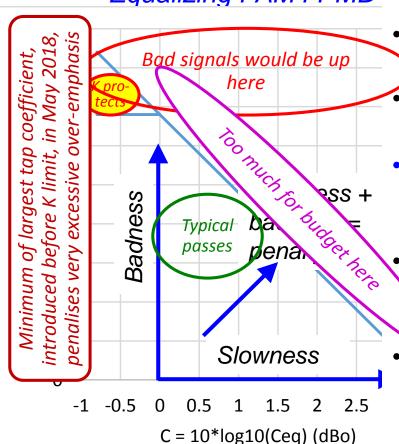
this way



- Transmitter can have +ve or –ve C
 - ve is usual
- Budget is calculated correctly for either sign of C
- K is limited to the same limit as TDECQ
 - which depends on PMD type
- K protects receiver and helps avoid bad error floor
 - See slide 11
- Protects against non-equalizable elements of signal
 - including non-equalizable overshoot
 - K is not intended to protect against overemphasis
 - Another spec protects against equalizable over-emphasis: see <u>dawe 3cd 01b 0518</u>
- TDECQ, C and K from one measurement:
 no measurement cost impact

All in-force optical PAM4 is done this way

Equalizing PAM4 PMD



Transmitter can have +ve or –ve C

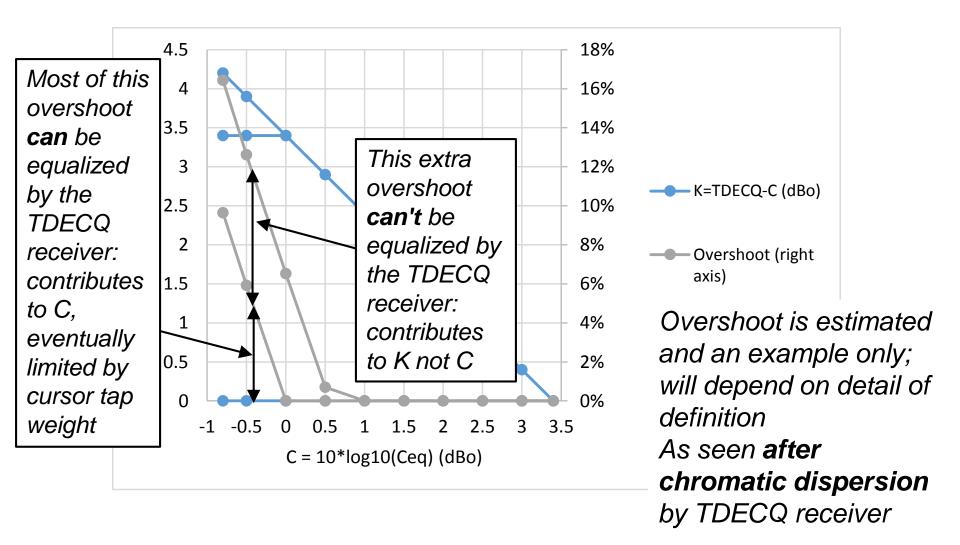
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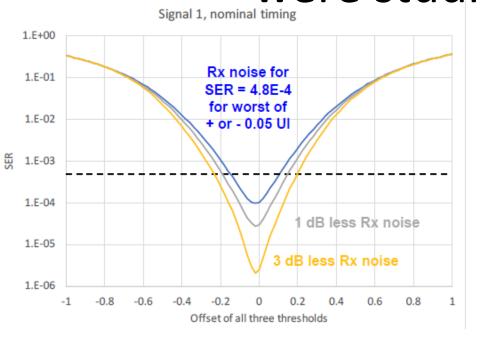
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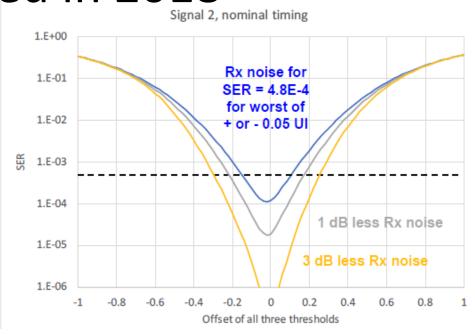
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At least two sorts of overshoot



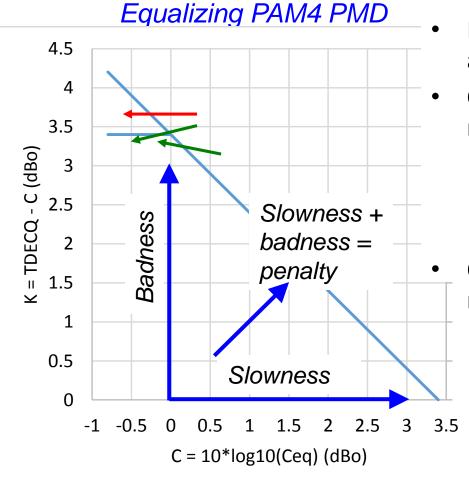
Bad signals with equalizable overshoot were studied in 2018





- Both signals have 3 dB TDECQ
- Signal 1 is over-emphasised, signal 2 is near neutral
- This is from http://ieee802.org/3/cd/public/adhoc/archive/anslow-062718-3cd-adhoc.pdf
- Signal 1 has worse error floor; real-world impairments will exacerbate the difference
- So would a higher TDECQ limit
- http://www.ieee802.org/3/cd/public/July18/dawe_3cd_01b_0718.pdf slides 9 and 10 show the same effect

Examples of setting up marginal or poor transmitters

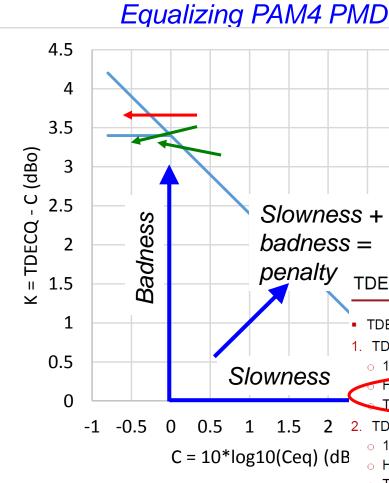


- From the start point of a green or red arrow;
- Changing the transmitter setups can make the two green transmitters pass
 - if they don't fail any overshoot spec
 - If transmitter power or swing is limited, the lower green arrow hasn't improved the link
 - OMA down as emphasis up

Changing the transmitter setup didn't make this red transmitter pass

It shouldn't; it is too noisy or distorted

Examples of setting up marginal or poor transmitters



IEEE P802.3cu 5 Mar 2020

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TDECQ - 10log₁₀(C_{EQ}) Explanation

- TDECQ 10log₁₀(C_{FO}) ≤ TDECQ_{MAX} Table 140-6, 151-7
- 1. TDECQ ≈ TDECQ_{MAX}
 - $0.10\log_{10}(C_{EO}) > 0 \leftrightarrow C_{EO} > 1$
 - H_{EQ}(f) can not adopt to Low Pass
 - TX pre-emphasis qualitatively constrained
- 2. TDECQ < TDECQ_{MAX}
 - o $10log_{10}(C_{EQ})$ unconstrained $\leftrightarrow C_{EQ}$ unconstrained
 - H_{FO}(f) unconstrained
 - TX pre-emphasis unconstrained

TDECQ - $10\log_{10}(C_{EQ})$ is just a lousy pre-emphasis spec

From -Yes, it's that simple! http://ieee802.org/3/cu/public/Jan20/cole 3cu 01b 0120.pdf#pas

TDECC20, January 2020 vvii CSS, Dadiness and 70 versione

No, you can do as you please (green arrows) as long as there isn't too much noise and distortion (red arrow)

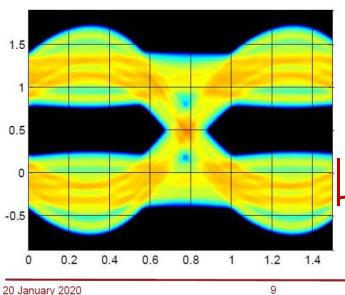
No. K is not a preemphasis spec, it's a badness spec. It's that simple.

Equalizable overshoot again

TDECQ - 10log₁₀(C_{EQ}) Explanation Confirmation•

Piers Dawe analysis

http://www.ieee802.org/3/cd/public/July18/dawe 3cd 01b 0718.pdf#page=11



"very bad" TX eye that can be stopped by TDECQ - $10log_{10}(C_{EQ})$ i.e. "bad" or "very bad"

.e. pau or very pau

= pre-emphasis

This signal would be allowed by the tap weight rules, and excluded by the K = TDECQ – 10log₁₀(Ceq) rule

"bad = pre-emphasis" is not correct

In dawe_3cd_01b_0718 and in TDECQ theory, overshoot was/is not bad in itself

But the thick bands are really bad for PAM4!

- Remember this is a PAM2 eye
- K = TDECQ 10log₁₀(Ceq) screens out this bad signal
- because it's noisy or distorted

But for real receivers with real impairments, too much overshoot must be bad

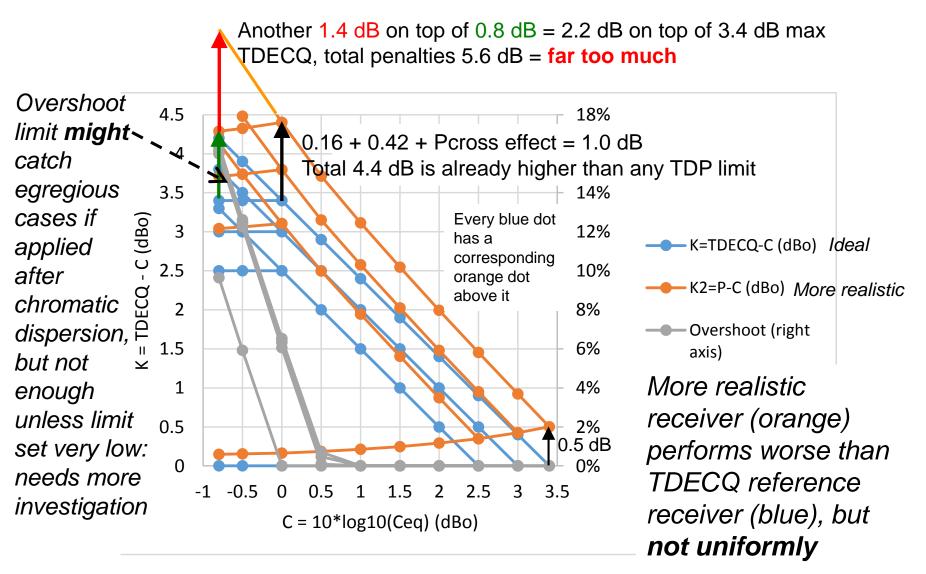
From http://ieee802.org/3/cu/public/Jan20/cole 3cu 01b 0120.pdf#page=9

 The minimum cursor tap weight in the TDECQ method limits equalizable overshoot at the receiver

 Need this to protect the equalizer, but probably need something more to protect the receiver front end

II-VI

Including more realistic receiver impairments



Different impairments affect the right and top areas

Summary of relation between TDECQ and overshoot

- Two sorts of overshoot, equalizable (emphasis) and not
- Equalizable overshoot is constrained by minimum of largest tap coefficient (0.8)
 - Towards the left on the TDECQ map
- Unequalizable overshoot is in the same category as all unequalizable ISI and signal noise, measured by K
 - Towards the top of the TDECQ map
- K is like TDP, TDEC and EVM: a necessary screen for bad signals. All modern optical specs have one
 - K was never intended to screen for too much emphasis;
 largest tap coefficient (from May 2018) already does that

Non-equalizable overshoot, questions

- See http://ieee802.org/3/cu/public/Jan20/cole-3cu-01b-0120.pdf slides
 12 to 15
- Should know what the threat mechanism is before we really know how to screen it out
- E.g. the overshoot could cause bad clipping in the receiver
 - Where in the receiver?
 - Is it overshoot relative to signal or peak-to-peak swing that matters?
 - If the former, the overshoot should be measured after the chromatic dispersion (i.e. at TP3), as TDECQ is
 - If the latter, the allowed swing at TP3 could be larger than at TP2 by the minimum loss of the fibre
- Next slides address measuring total overshoot
 - We have a spec in place for equalizable over-emphasis, probably too loose
- How do we determine how much an overshoot limit should be?

Thoughts on measuring overshoot

- Limiting overshoot at TP2 may not be enough if chromatic dispersion can make the overshoot higher at TP3, as seems very likely
- A measurement on a square wave measures the worst of preemphasis and post-emphasis, but a real signal's overshoot can be determined by the sum of these
- Square wave is a bad choice of pattern anyway because PMAs may fail to lock on it and forward the signal correctly to the PMD transmitter output (as 120.5.11.2.4 says)
- Traditional peak measurements are distorted by scope noise, particularly for optical scopes at such high bandwidths

How to measure overshoot 1/2

- Apply the spec to the same cases as TECQ and TDECQ: TP2, TP3 with most positive chromatic dispersion, and TP3 with most positive chromatic dispersion
- Use the same pattern and observation bandwidth as for T(D)ECQ so that determining the overshoot is another free byproduct of measuring for T(D)ECQ, with a much simpler, noniterative, calculation
- Find the scope noise
- Create a vertical histogram from the measured waveform (not the equalized one)
 - Width of histogram is 1 UI (or e.g. 2, as SSPRQ's length is an odd number of UI
- Convolve the histogram with the noise that could be added to it at maximum T(D)ECQ, e.g. OMAouter*10^{-3.4/10}/(6*3.414), RSS-reduced by the scope noise

How to measure overshoot 2/2

- Find the points on the two outer tails where the CDFs come to a number such as 5e-5
- Either find the distance from the "three" level to the upper point, and from the lower point to the "zero" (these are the overshoot and undershoot before normalisation)
- or find the distance from the average level to the upper point, and from the lower point to the average (these are the peak excursions)
 - Which one is more relevant to the receiver?
- Normalise by either OMA or standard deviation of the waveform, or don't normalise but scale for minimum TP2-TP3 loss
 - The first is more familiar, the second avoids some error from the pattern dependency of the OMA definition
- Limit upper and lower separately because excursions on just one side could overload a receiver
- To do: choose the limits based on what information?

Conclusions

- Receivers that cope with unnecessary corner cases cost more power, and typically, something else is compromised
- Receiver protection should not be removed retrospectively
 - But adding an overshoot limit is attractive. Can avoid test cost by reusing TDECQ measurement, but much simpler, quicker calculation
- Limiting C, C+K, and K are all worth doing
 - K just as we included a TDP limit in previous standards
 - Protect against transmitter evading the intention of the spec
 - Not clear yet if real transmitters would be found in the high-K corner, but free to screen
 - I offered a more gentle limit that slopes up to the left in 2018, but it got no traction
- Limiting K is as practical and convenient as TDECQ, a by-product of TDECQ, and what we have been used to since July 2018 (P802.3cd/D3.4)
 - No Tx measurement cost
- Stay with K limit, add spec for overshoot at TP3 as well as TP2