

Channel Code Considerations for 10GbT Signaling

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Jose Tellado, Dariush Dabiri, Teranetics

Ofir Shalvi, Ariel Yagil, TI

Supporter: Scott Powell, Broadcom

Motivation

- Two different signaling architectures have been discussed so far
 - TCM: 10PAM 4D-8State code at 833Ms/s – Solarflare
 - LDPC: 8PAM (1723,2048) code at 1000Ms/s – Intel
- These proposals differ in other aspects too
 - Packet format, overhead bits etc.
 - Equalization approach
- This presentation compares the coding schemes while normalizing other factors out
- We also include some other well known schemes

Ideal Performance Bounds

- A “good” transceiver design would convert the ISI+Xtalk+noise channel into an (approx) AWGN channel
- Shannon capacity for and ideal AWGN channel
 - $C = \frac{1}{2} \log_2(1 + \text{SNR})$ bits per 1D symbol (bps)
- For example, a capacity approaching code (“infinite” delay) can operate at 2.5bps with “zero” BER at SNR of 15 dB

Practical Performance Bounds

- For uncoded MPAM, M even
 - $\text{BER} \sim Q(\sqrt{3/(M^2-1)} \cdot \text{SNR})$
- Solving the equation above we have
 - $\text{Rate} = \frac{1}{2} \log_2(1 + \text{SNR}/G(\text{BER}))$ per 1D symbol
 - $G(\text{BER})$ is the Gap or Loss relative to capacity and depends on the target BER.
- For uncoded PAM, $G(10^{-12}) \sim 12\text{dB}$
- For coded systems, d increases and $G(\text{BER})$ is reduced
 - The reduction of $G(\text{BER})$ is called coding gain

Multi-channel SNR in code design

- Note that the IL and residual X-talk (NEXT, FEXT, ANEXT) level may vary from pair to pair due to the physical channel parameters (cable length, separation, connectors, ...)
- SNR variations should be considered in the code design

Possible Channel Codes

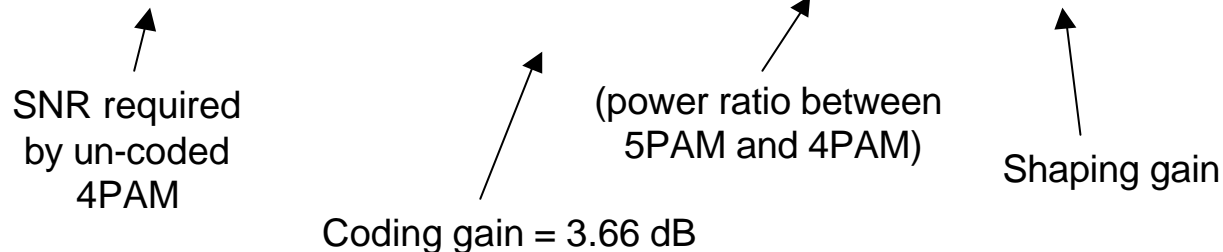
Scheme	Loss from Capacity @ BER=1e-12	Delay [μsec]	Maturity	Receive Complexity	Equalizer
TCM + shaping	5-8dB (Solarflare TCM=8dB)	< 0.1	Mature	Low – moderate	Pre-coding or receiver equalization
TCM+RS +Shaping	3-6dB	0.75-4.5	Mature	Low – moderate	Pre-coding recommended
LDPC + shaping	1-4dB (Intel LDPC=3.8dB)	0.5-2	New	High	Pre-coding

- Lower loss from capacity translates to better link quality
 - For each scheme, lower loss typically requires a more complex receiver and more delay (latency)

The 1GbT 5PAM-4D-TCM code

Achieves BER=1e-12 for 2 bps in an ideal (no ISI) AWGN channel at Es/N0 (dB) of

$$23.9 - (5.7 - 10 \log_{10}(8/5)) - 0.4 = 19.9 \text{ dB}$$

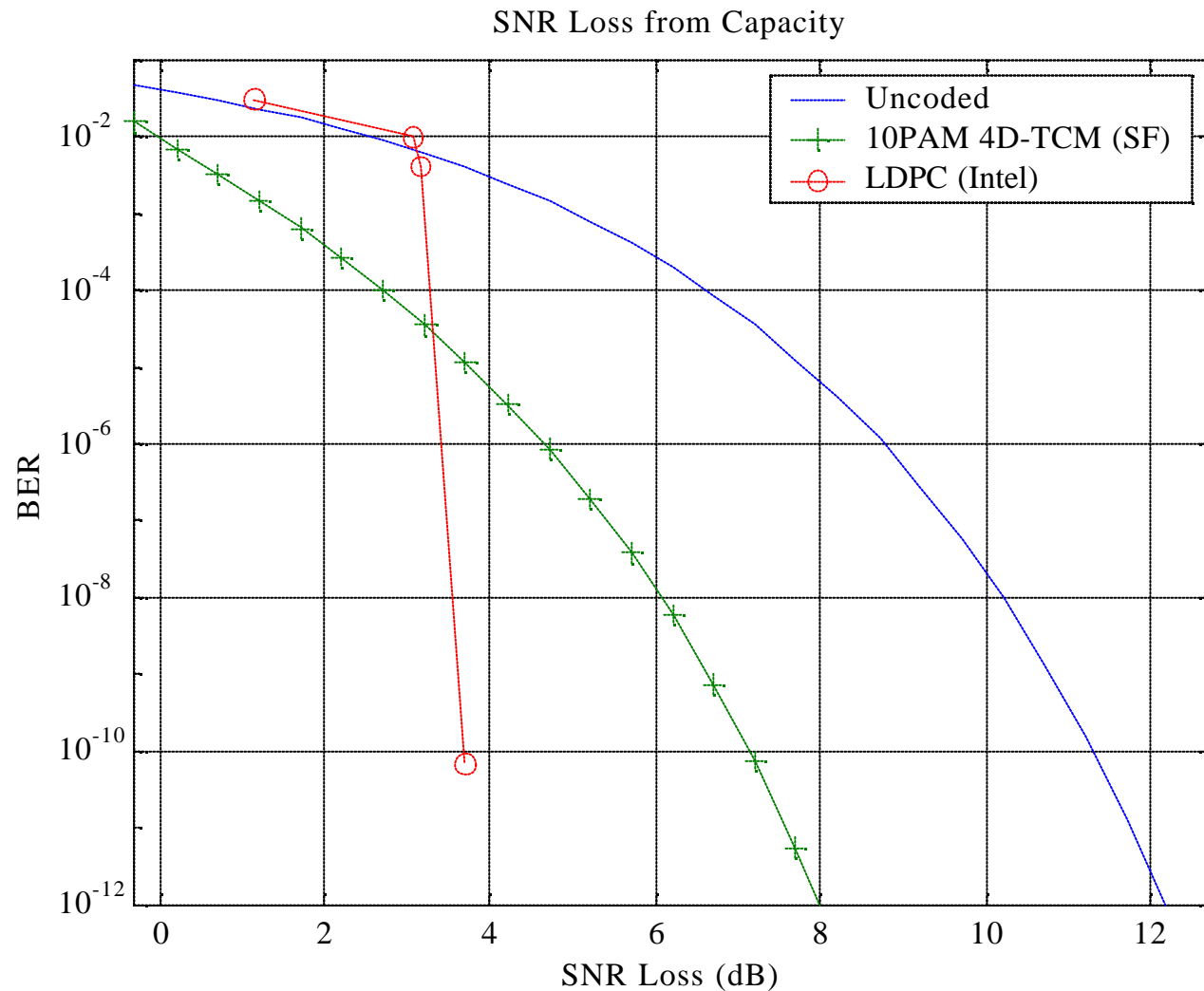


Gap from capacity = 8dB

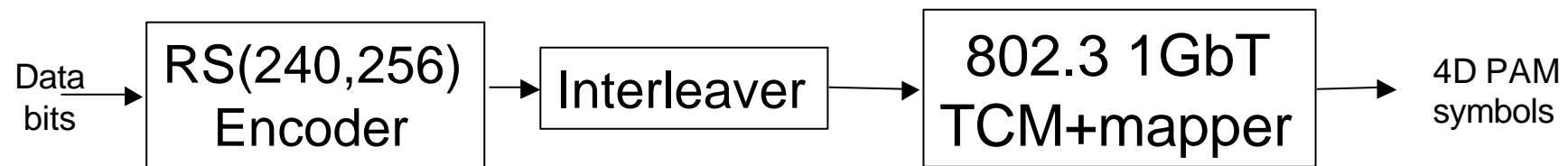
LDPC/Turbo Codes

- A large body of work (most starting mid 90s) has shown that LDPC/Turbo codes can approach the Shannon bound
 - Most of the published literature has focused on the low SNR
- Intel's LDPC 8PAM (1753,2048) proposal:
 - Achieves $\text{BER}=10^{-12}$ for $2774/1024=2.68$ bps and $\text{SNR}=19.8\text{dB}$
 - The Shannon bound at 2.68 bps is $\text{SNR}=(2^{2 \cdot 2.68}-1) \sim 16\text{dB}$
 - The loss from capacity is $19.8-16 = \mathbf{3.8dB}$
 - The intrinsic decoding delay (i.e. with infinite HW) is 0.256 micro-seconds, but practical decoders will have additional delay
 - The SNR margin could be improved by using shaping algorithms, different 4D mappings, larger block sizes, etc. typically at the expense of more complexity and/or latency

Performance of current Proposals

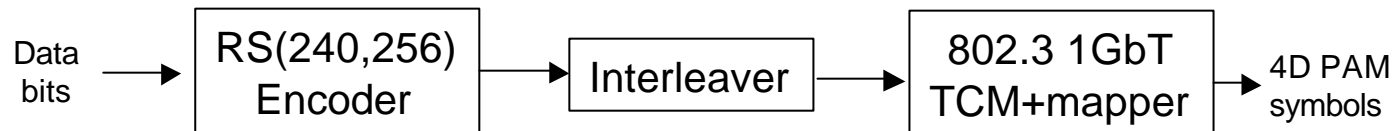


A concatenated RS+TCM based on the 1GbT code

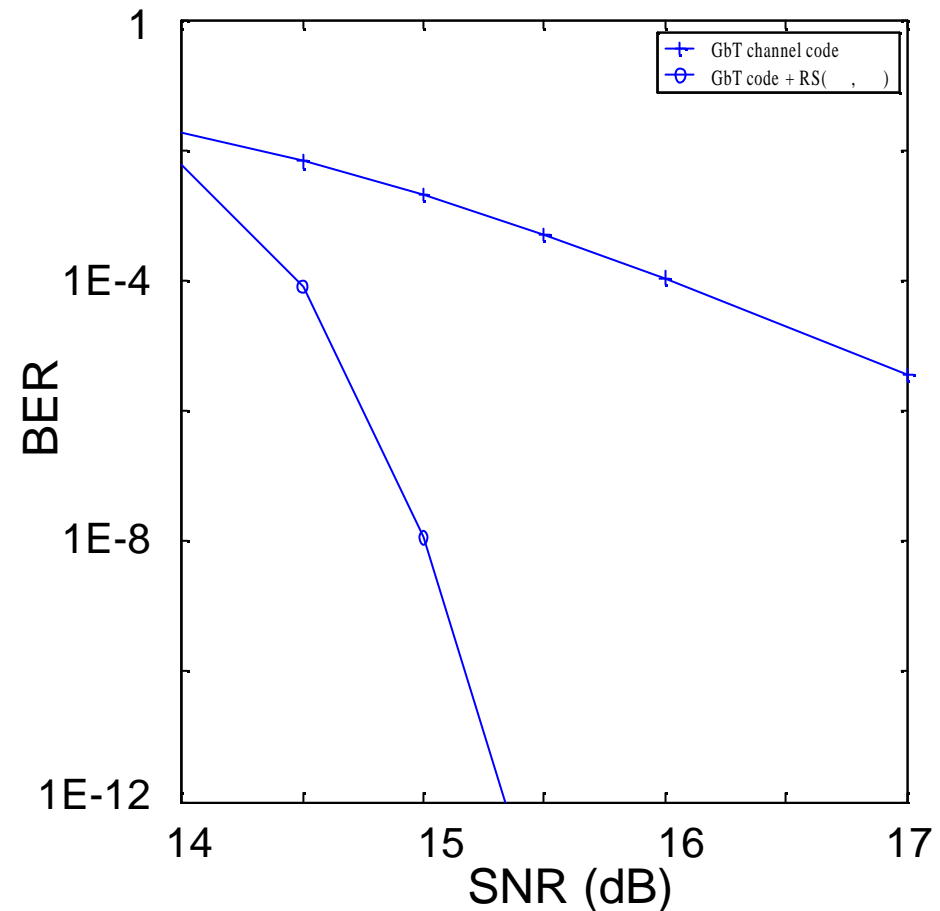


- Data rate of 1.875 bits per dimension
- Can easily be generalized to 10PAM (2.8125 bits per 1D-PAM symbol)
- Low complexity, mature decoding algorithms

Concatenated RS+TCM – Performance Analysis

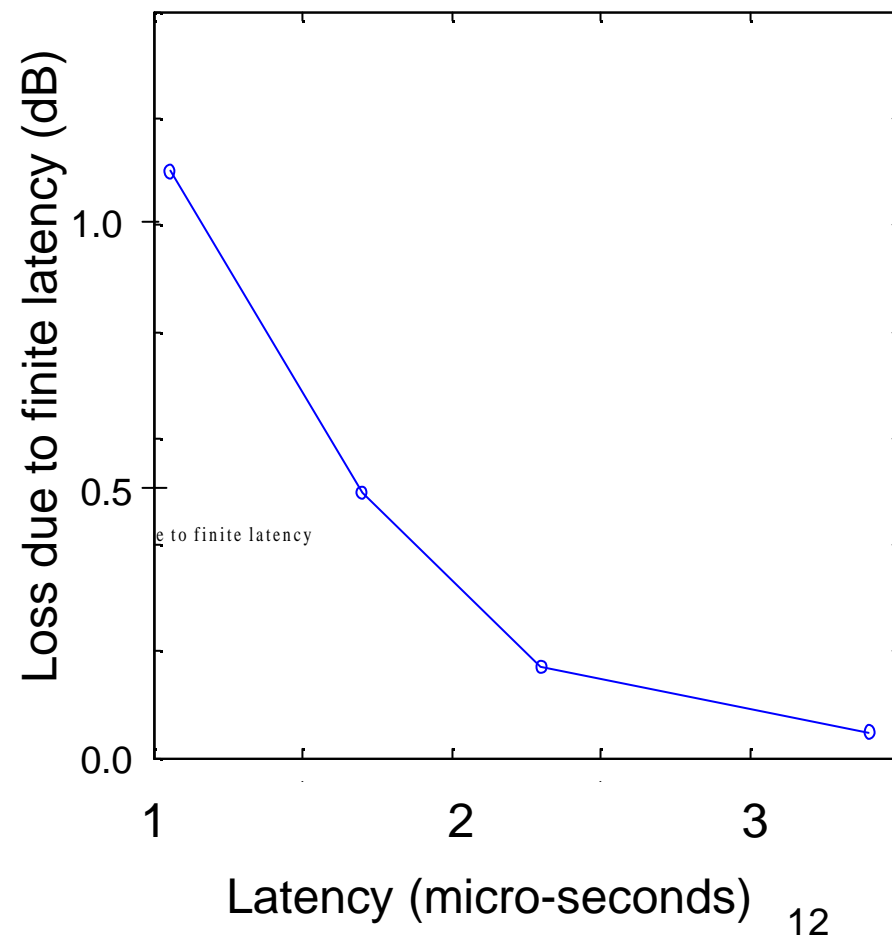


- BER=1e-12 @ SNR of 15.4 dB – **only 4.4 dB short of Shannon Capacity**
- Using a standard hard decoding algorithm
- Similar gain to that of the LDPC proposed in the November meeting
- Analysis assumes ideal interleaver



Concatenated RS+TCM – Performance/Latency tradeoff

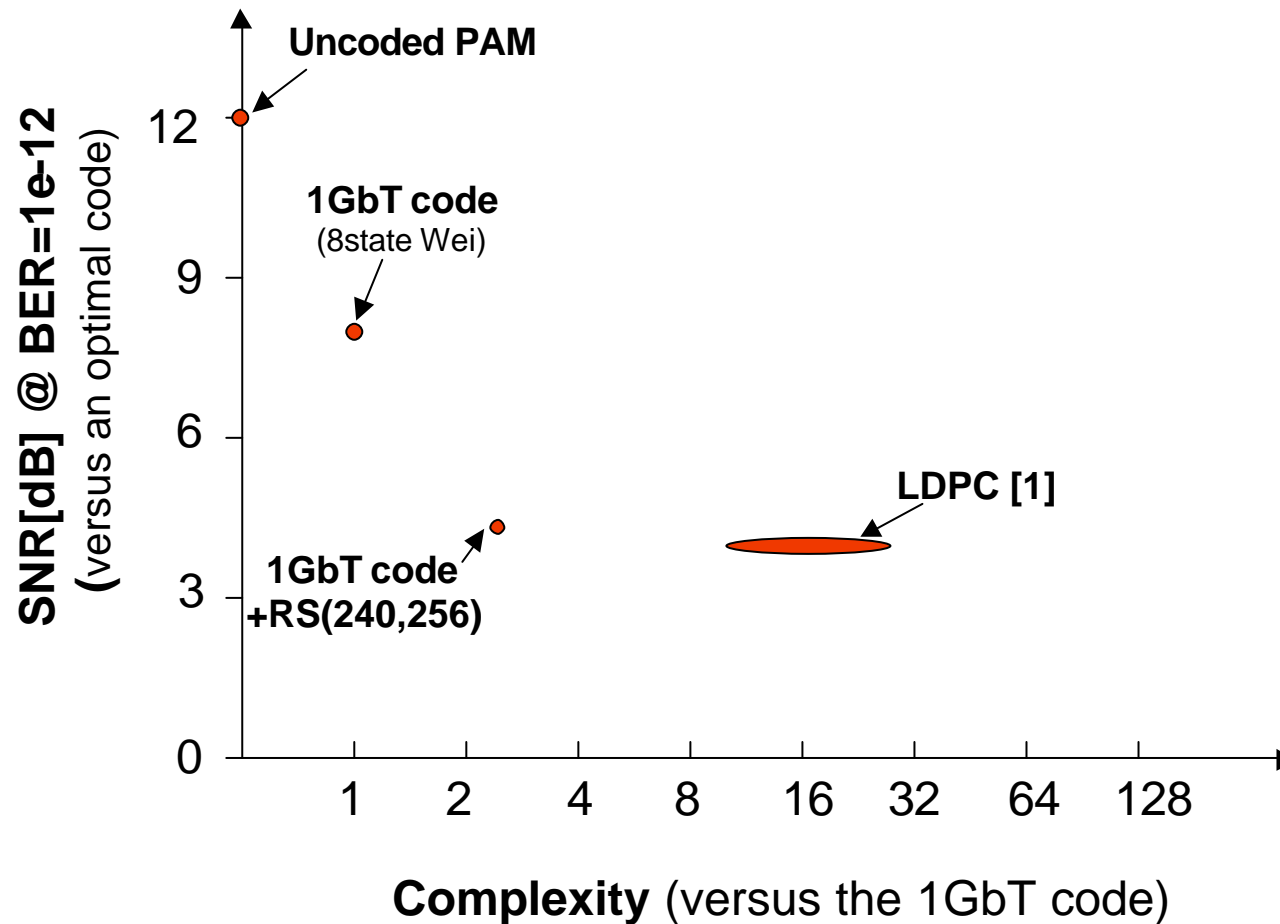
- The analysis of the code with a short interleaver is more complicated.
- We assess that the loss due to an interleaver latency of 2.5 micro-seconds is a small fraction of a dB.
- We assess that the loss due to a latency of 1 micro-second is about 1-1.5 dB.



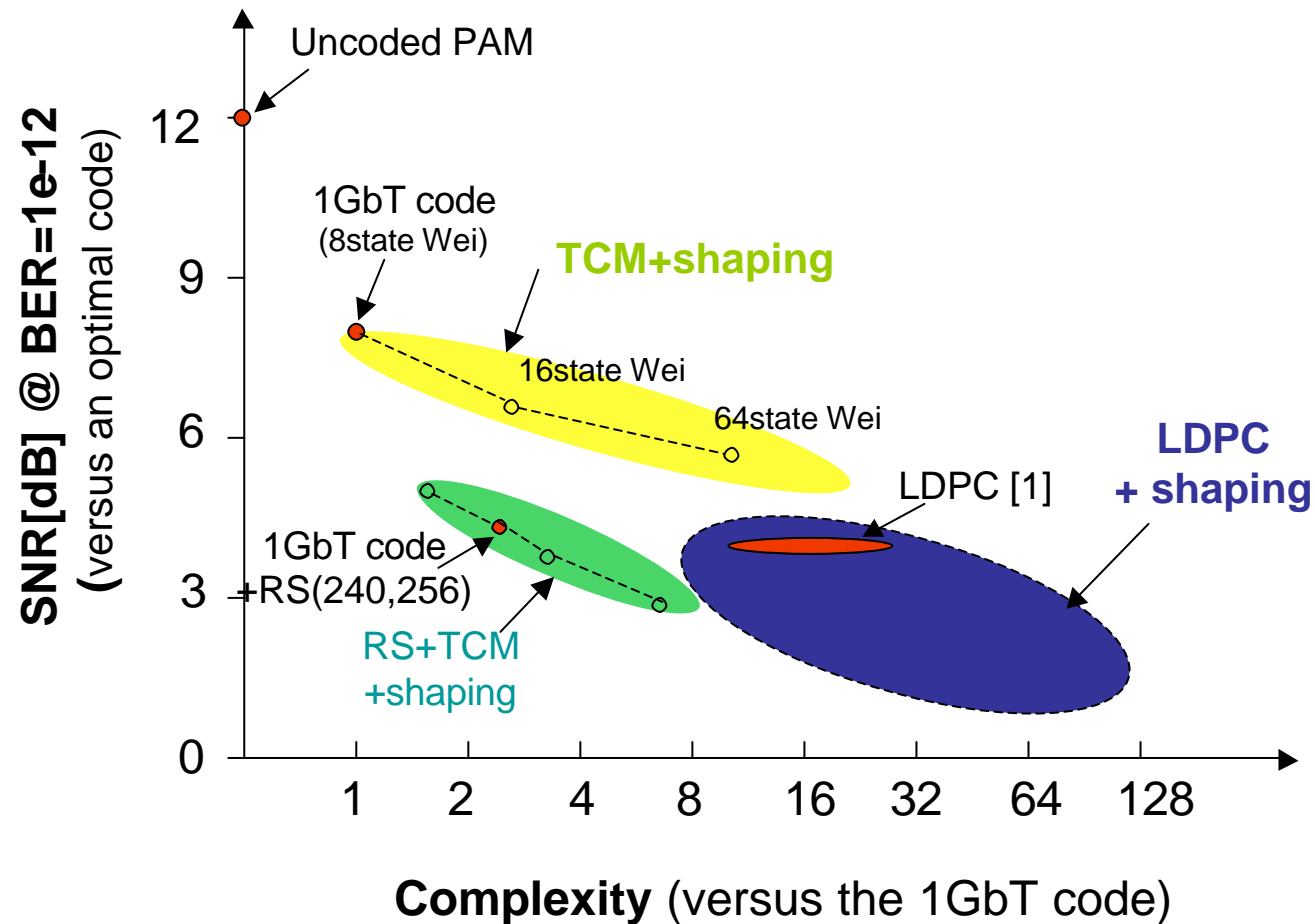
Improved concatenated codes

- It is possible to further gain 0.5-0.7 dB by employing constellation-shaping algorithms.
- Lower latency or additional coding gain can be achieved by employing other concatenated coding schemes.

Performance-Complexity Summary



Performance-Complexity Summary



Conclusion

- We seek input from task force participants on:
 - Latency budgets
 - Performance/complexity tradeoffs
- Based on these inputs, specific codes can be optimized for the 10GBASE-T application
 - Concatenated Codes
 - Optimize for a tolerable latency range
 - LDPC
 - Optimization and more detailed evaluation of performance and complexity.
 - Optimize symbol rate and packet overhead
 - Evaluate addition of constellation-shaping gain to codes