#### **Tx Equalizer Coefficient issues**

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#### Comment #64

- The off-axis requirements in table 72-11 do not match the governing equations of the transmit equalizer. All 3 measurement points are dependent on all 3 coefficients.
  - If Vpk is kept constant, a step on any coefficient will affect at least two of the 3 measurement points.
  - If Vpk is not kept constant, a step on any coefficient will affect all 3 measurement points.
- Recalculate the off-axis entries based on the governing equations of the transmit



#### Comment #65

- Draft 2.0 required that C<sub>q</sub> shall be adjusted to maintain Vpk/A over all transmitter states (k). This requirement has been removed in Draft 2.1, and the transmitter output waveform requirements have been changed to render constant Vpk implementations non-compliant.
- Implementing Tx equalization on SERDES using assignable CML output fingers is an area-efficient alternative to DAC style structures. Forty fingers of 2.5% meet the performance requirements adopted in May Motion #10, whilst automatically providing constant Vpk. However the coefficient step trading (to/from C<sub>0</sub>) required to maintain constant Vpk mean that the measured step changes in Table 72-11 are doubled.
- We are concerned that the changes in Draft 2.1, preclude the use of natively constant-Vpk transmit structures for no demonstrable benefit.
  - Of course it is possible to make a fingered approach work with non-constant Vpk, by doubling native resolution, or by turning fingers off, but this increases transmitter complexity and area for the dubious benefit of reduced output swing.
- Re-instate the constant Vpk requirement, and reflect this requirement in Table 72-11 values.
  - Or allow constant Vpk by providing an additional or modified Table 72-11.

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### Observations

- There are 2 schools of thought on TX FIR implementations
  - 1) Constant V<sub>pk</sub>
    - Does not require Cursor control
    - Maximizes signal strength
    - Usually augmented by an overall gain control
      - Allows signal strength to be adjusted independent of equalization
      - Allows link power reduction
    - 3 tap FIR only has 2 degrees of freedom (128 states)
      - Leads to simpler, more efficient, implementations
  - 2) Independent coefficients (Variable  $V_{pk}$ )
    - Analogous to Rx DFE coefficients
      - Training algorithms resemble Rx algorithms
    - 3 tap FIR has 3 degrees of freedom (>1000 states ?)
    - Difficult to adjust signal strength just via coefficients
      - Still need gain control
- All contributed Tx FIR analysis has been for constant V<sub>pk</sub>



# Changes

- The changes recommended in healey\_01\_0905 to clean up the transmitter waveform tests removed the constant V<sub>pk</sub> requirement.
  - Rather than just removing a requirement on non-constant Vpk transmitters it made constant Vpk transmitters non-compliant
  - The bulk of this presentation was on improving the test methodology
  - In this context, I don't believe the practical impact of removing  $V_{\text{pk}}$  was realized.
    - It wasn't discussed
  - I have voted NO on the ballot based on this change to ensure that we discuss it now
- There may be practical reasons to limit ourselves to one style of equalizer
  - Is there an equalizer training algorithm that will work with both?
    - Maintaining constant Vpk in the training commands will work



## Conclusion

- I am concerned that in order to reach consensus we have ended up with a lowest common denominator approach.
  - Efficient constant Vpk implementations are now excluded
  - We still have issues with (comments on) the Tx test methodology
- I'd like to see a return to a constant Vpk based test methodology.
  - Maintain Vpk by changing  $C_0$  in concert with pre or post cursor in the training packets.
    - This is compatible with either implementation
- Are we wise to not define an algorithm ?



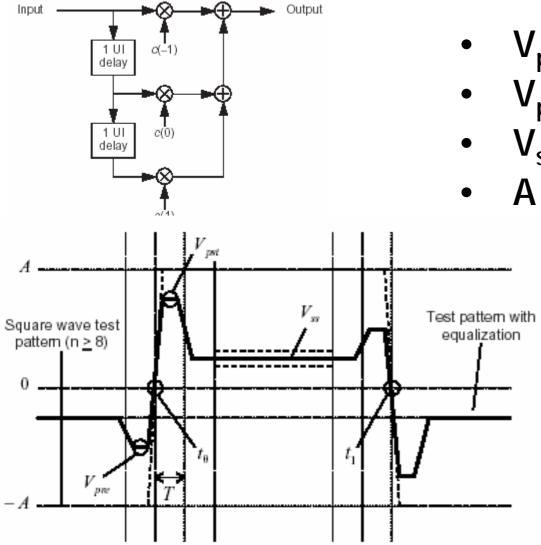
### **Backup slides**

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#### Measurements are not independent



• 
$$V_{pre} = +c(-1) -c(0) -c(+1)$$
  
•  $V_{pst} = +c(-1) +c(0) -c(+1)$   
•  $V_{ss} = +c(-1) +c(0) +c(+1)$ 

$$A = V_{pst} - V_{pre} - V_{ss} = -c(-1) + c(0) - c(+1)$$

- Note c(-1) & c(+1) are always negative !
- So A = |C(-1)| + |C(0)| + |C(+1)|

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## Interdepenency (A not constant)

- V<sub>pre</sub>, V<sub>pst</sub>, V<sub>ss</sub> & A are all sums of the 3 coefficients
   They differ only in the sign of the coefficients
- If A is not kept constant, a step in any coefficient will cause an equal changes in  $V_{pre}$ ,  $V_{pst}$  &  $V_{ss}$

- The effect will differ only in sign

- c(-1)++ causes  $V_{pre}$ ++ ,  $V_{pst}$ ++ &  $V_{ss}$ ++
- c(0)++ causes V<sub>pre</sub>-- , V<sub>pst</sub>++ & V<sub>ss</sub>++
- c(+1)++ causes  $V_{pre}$ -- ,  $V_{pst}$ -- &  $V_{ss}$ ++



#### Interdepenency (with constant Max amplitude)

- To keep A constant (-c(-1) +c(0) -c(+1)) must be kept constant.
  - A change in one coefficient must be offset by an equivalent total change in the other 2 coefficients.
- Practical implementations cannot arbitrarily scale output.
  - The high speed DSP needed to scale the output to keep A constant is not practical at 10Gbps
  - Must make simple changes to the other coefficients Eg
    - Change other two coefficients by 1/2 step each
    - Or change one of the other coefficients by one step



### Effect of <sup>1</sup>/<sub>2</sub> step trade

• Consider a change in one coefficient offset by ½ step changes in the other 2 coefficients.

- If 
$$C(-1)$$
++,  $C(0)$ +=  $\frac{1}{2}$ ,  $C(+1)$ -=  $\frac{1}{2}$   
•  $V_{pre}$  = + $C(-1)$  - $C(0)$  - $C(+1)$  =  $V_{pre}$  +=

• 
$$V_{pst} = +C(-1) + C(0) - C(+1) = V_{pst} + =2$$

• 
$$V_{ss} = +c(-1) + c(0) + c(+1) = V_{ss}$$

• One measurement point changes by one step, another by two steps, the other stays the same.



### Effect of 1 step trade

- Consider offsetting a change in c(-1) or c(+1) by a 1 step change in c(0).
  - If c(-1)++, C(0)++, no change in C(+1) for constant A

• 
$$V_{pre} = +c(-1) - c(0) - c(+1) = V_{pre}$$
  
•  $V_{pst} = +c(-1) + c(0) - c(+1) = V_{pst} + =2$   
•  $V_{ss} = +c(-1) + c(0) + c(+1) = V_{ss} + =2$   
If  $c(+1)++$ ,  $C(0)++$ , no change in  $C(-1)$  for constant A  
•  $V_{pre} = +c(-1) - c(0) - c(+1) = V_{pre} - =2$   
•  $V_{pst} = +c(-1) + c(0) - c(+1) = V_{pst}$   
•  $V_{ss} = +c(-1) + c(0) + c(+1) = V_{ss} + =2$ 

- Changes in c(0) must be offset against c(-1) or c(+1)
  - How to decide which ?

