



# Simulation Results for 10 Gb/s Duobinary Signaling

*Populating the Signaling Ad Hoc Spreadsheet*

***IEEE 802.3ap Task Force  
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# Talk Outline

- ❑ Overview
- ❑ Description of the Modeled Architecture
- ❑ Simulation Assumptions
- ❑ Simulation Results
- ❑ Conclusion



# Overview

- ☐ Introduction
- ☐ Simulation model will be explained
  - Block Diagram
  - Parameters
  - Assumptions
- ☐ Simulation results for duobinary signaling will be presented for 16 channels
- ☐ Conclusion

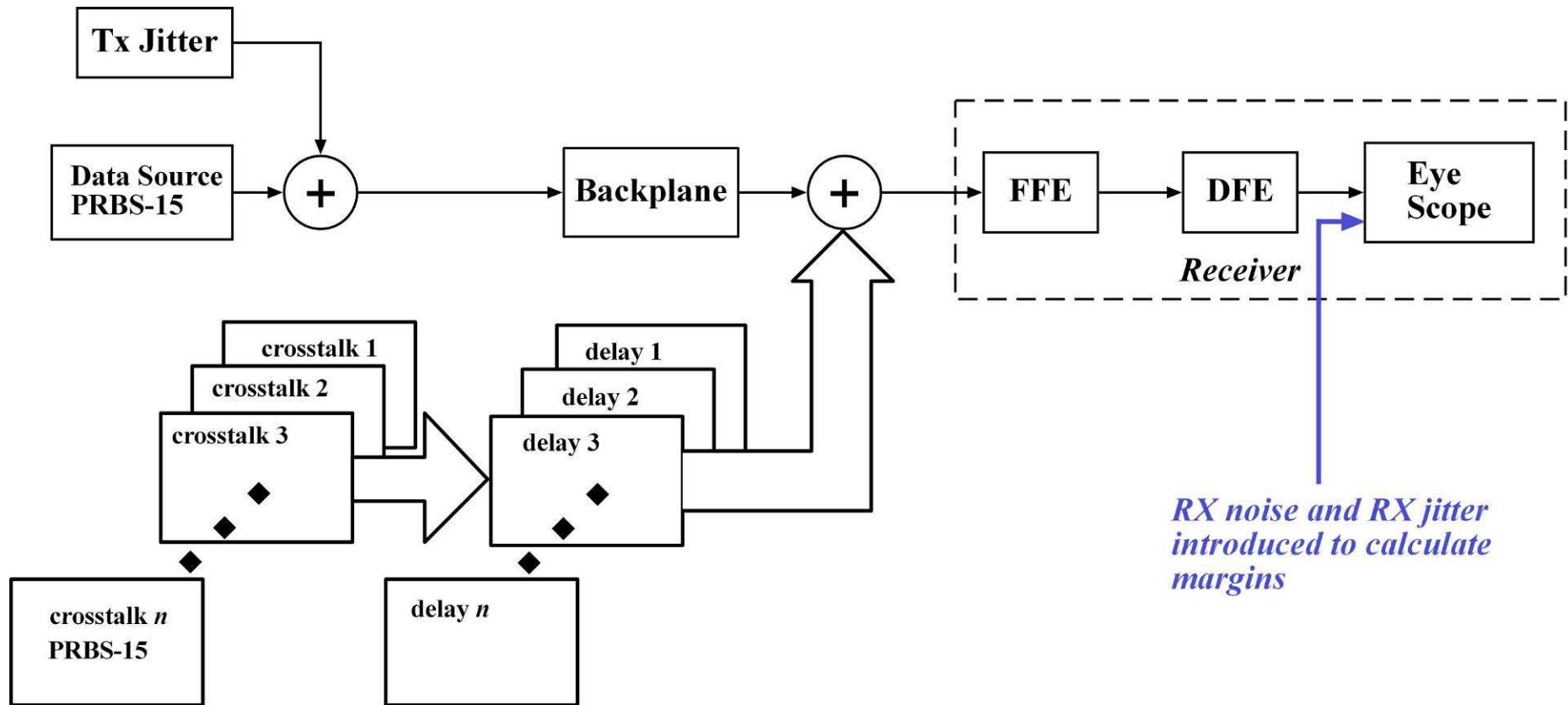


# Duobinary Signaling for 802.3ap

- ❑ Has been suggested as one of the possible formats for use in this standards body
- ❑ Contingent on the decision of the group...
  - Duobinary **may** be selected as a stand alone solution
  - Duobinary **may** be selected as part of the Unified Signaling Proposal for the 10G PHY
- ❑ *Either way, a complete understanding of the performance of duobinary signaling is necessary.*



# System Topology used in Simulations



# Simulation System Parameters

<i>Basic Signaling Properties</i>		
Parameter	Value	Units
Line code	Duobinary	
Bit Rate	10	Gb/s
Symbol Rate	10	Gb/s
Number of Levels	3	
Bits/symbol	1	

<i>Source Data Parameters</i>		
Parameter	Value	Units
Data Pattern	PRBS15	
Launch Amplitude	0.8	volts (p-p) diff.
TX jitter DJ (p-p)	10	Gb/s
Tx Jitter - DCD (p-p)	0.15	UI
Tx Jitter - Rj (rms)	0.0107	UI
AC/DC coupling	DC	

<i>Receiver Parameters</i>		
Parameter	Value	Units
Random noise	1.46	mv <sub>rms</sub>
Cross-talk type	Random	Random/deterministic
Crosstalk scaling	No	NEXT Scaled to the mask? Yes or no
Rx input Dj	0.15	UI(p-p)
Rx input Rj	0.15	UI(p-p)

<i>Simulation Parameters</i>		
Parameter	Value	
Number of bits simulated	15,000	
Samples/symbol	120	
Simulation tool	Matlab	



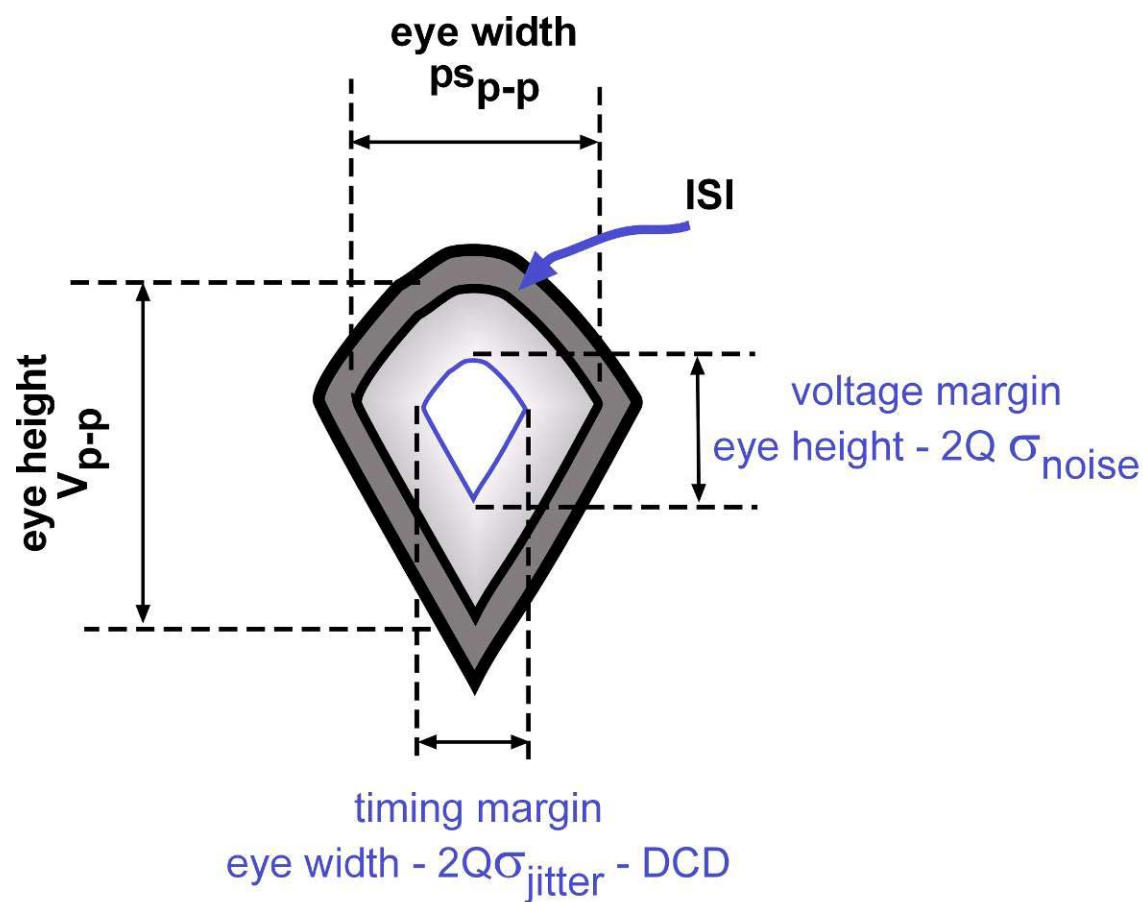


# Simulation Assumptions

- ❑ The number of crosstalk aggressors depends on available data from vendor
- ❑ Measured 4-port S-Parameters are used to represent transmission and cross-talk characteristics
- ❑ Crosstalk data streams are delayed in such a way as to provide a worst case scenario and are identical to the thru channel data stream
- ❑ The FFE gain is normalized so that it is PASSIVE
- ❑ The RX Noise and RX jitter are modeled by reducing the eye opening by the amount ...
  - $V_{p-p} - 2 Q \sigma_{\text{noise}}$
  - $ps(p-p) - 2 Q \sigma_{\text{jitter}} \cdot DCD$
  - where  $Q = 7.03, 7.94,$  and  $8.75$  for BERs of  $10^{-12}, 10^{-15},$  and  $10^{-18}$  respectively



# Explanation of Eye Margins



Q	BER
7.03	$10^{-12}$
7.94	$10^{-15}$
8.75	$10^{-18}$



# Simulation Results

## *Tyco ATCA Backplanes*

Channel ID	FFE Taps	DFE Taps	Simulated eye opening for data stream		Margin for BER=10 <sup>-12</sup>		Margin for BER=10 <sup>-15</sup>		Margin for BER=10 <sup>-18</sup>	
			Voltage mVp-p(diff.)	Timing ps(p-p)	Voltage mVp-p(diff.)	Timing ps(p-p)	Voltage mVp-p(diff.)	Timing ps(p-p)	Voltage mVp-p(diff.)	Timing ps(p-p)
Tyco Case1	3	3	50	37	31.68	6.80	29.02	4.86	26.64	3.12
Tyco Case2	5	3	76	37	77.36	7.3	74.7	5.36	32.34	3.62
Tyco case3	5	3	122	41	104.08	11.29	101.42	9.35	99.06	7.62
Tyco Case4	3	5	76	45	57.54	15.44	54.9	13.49	52.52	11.76
Tyco Case5	3	5	112	51	92.86	21.27	90.22	19.32	87.84	17.59
Tyco Case6	5	3	74	38	55.36	8.57	52.7	6.63	50.34	4.89
Tyco Case7	3	3	76	37	56.7	6.8	54.04	4.8	51.68	3.13

Note: 15,000 bits used to obtain these results



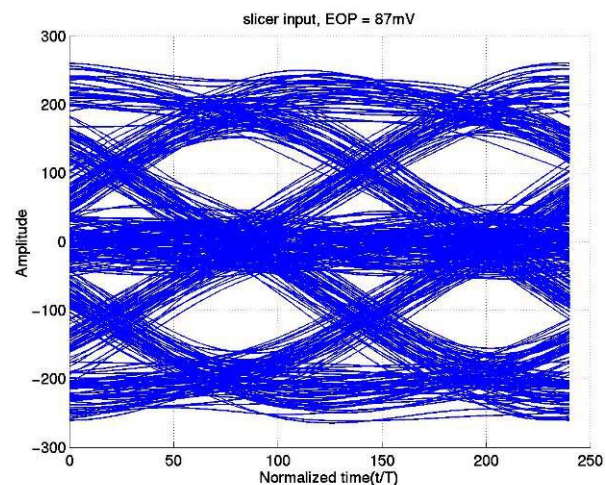
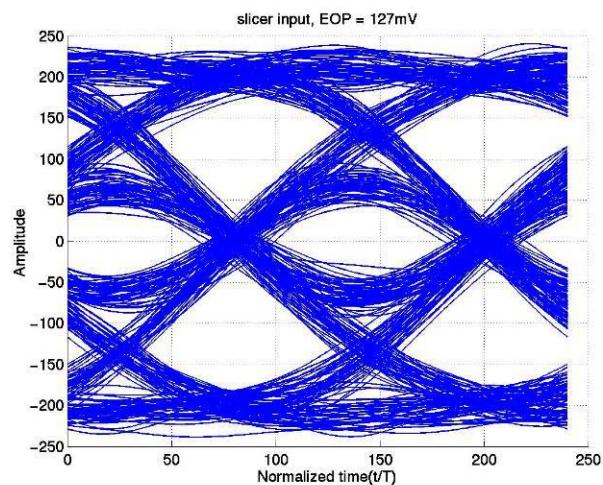
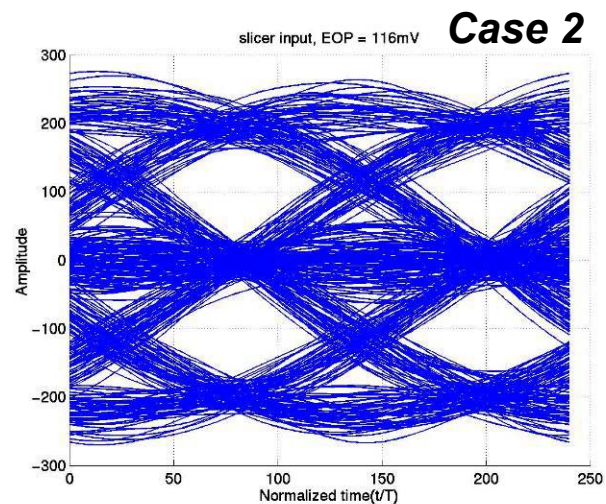
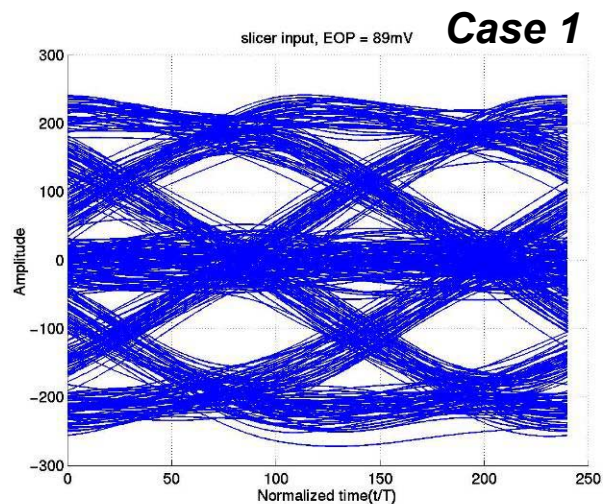
# Simulation Results

## *Intel ATCA Backplanes*

Channel ID	FFE Taps	DFE Taps	Simulated eye opening for data stream		Margin for BER=10 <sup>-12</sup>		Margin for BER=10 <sup>-15</sup>		Margin for BER=10 <sup>-18</sup>	
			Voltage mVp-p(diff.)	Timing ps(p-p)	Voltage mVp-p(diff.)	Timing ps(p-p)	Voltage mVp-p(diff.)	Timing ps(p-p)	Voltage mVp-p(diff.)	Timing ps(p-p)
Intel B01	3	3	88	49	68.72	19.10	66.06	17.15	63.68	15.41
Intel B12	3	3	84	52	65.76	21.69	63.1	19.75	60.74	18.02
Intel B20	3	3	82	45	62	15.22	59.34	13.28	56.98	11.55
Intel M1	3	3	56	48	37.32	18.36	34.64	16.42	32.28	14.68
Intel M20	3	3	72	39	52.62	9.15	49.96	7.21	47.6	5.47
Intel T1	9	5	54	35	55.24	4.58	52.58	2.63	50.2	0.90
Intel T12	7	3	72	35	52.2	4.97	49.54	3.02	47.18	1.29
Intel T20	7	3	72	37	52.58	7.40	49.92	5.46	47.54	3.73
Intel T32	7	3	66	39	47.94	9.09	45.28	7.14	42.92	5.41



# Tyco Cases 1-4



## Notes:

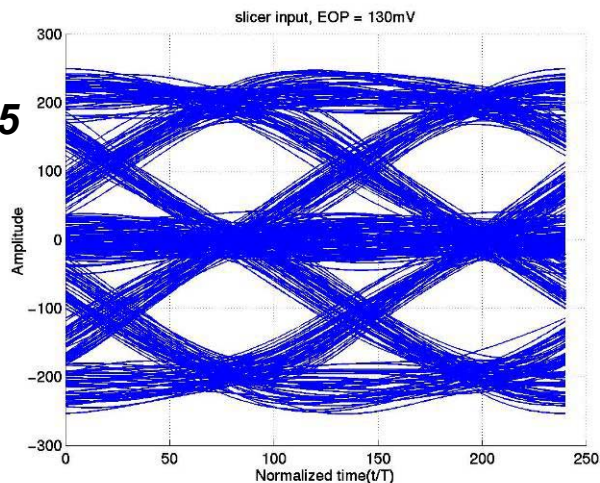
- **Vertical Axis**  
Units = mV
- **Horizontal Axis**  
120 units = 100ps
- **EOP** = eye opening
- 1200 bits are plotted



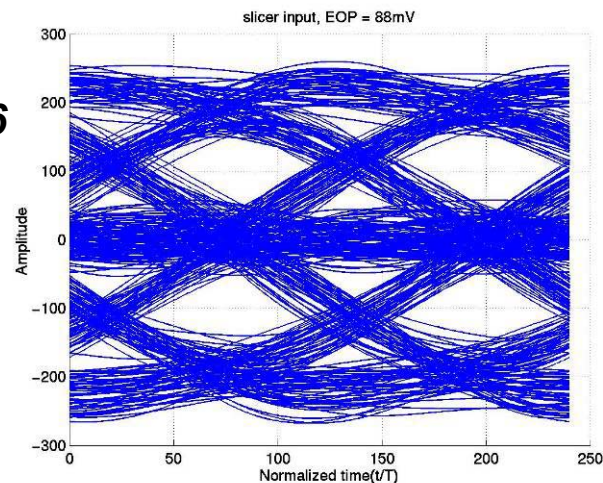


# Tyco Cases 5-7

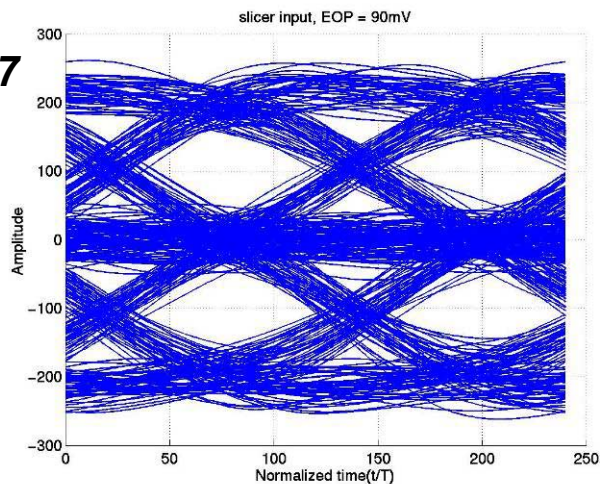
**Case 5**



**Case 6**



**Case 7**

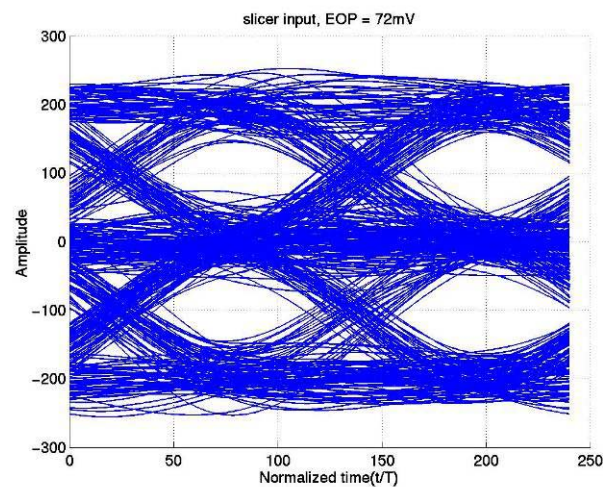
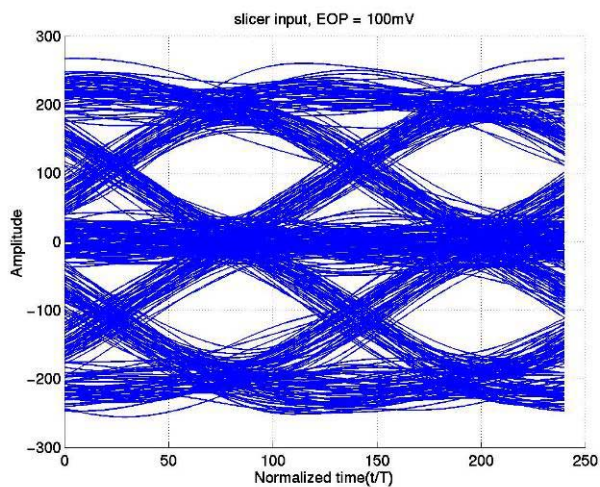
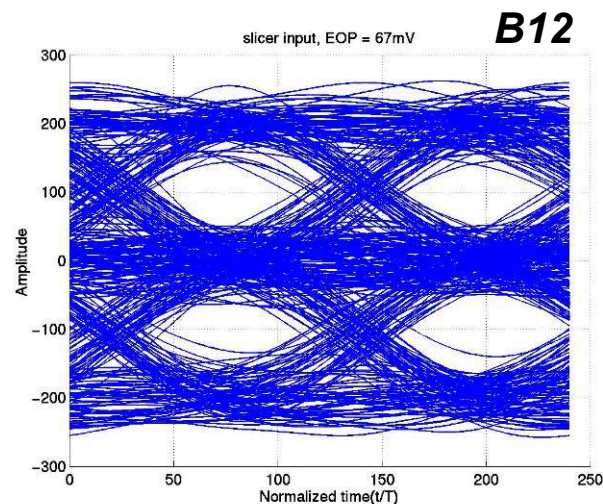
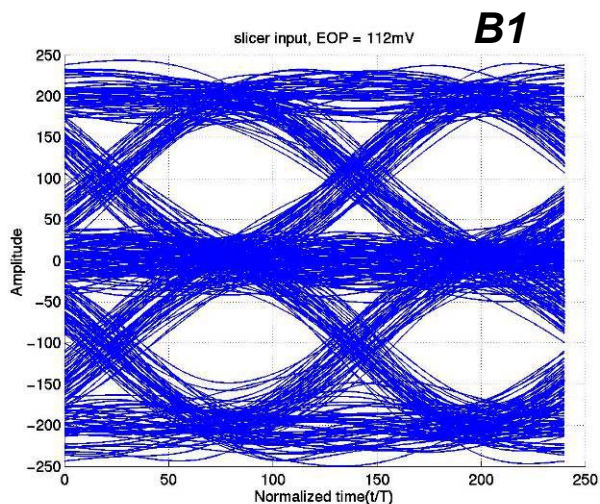


## Notes:

- **Vertical Axis**  
*Units = mV*
- **Horizontal Axis**  
*120 units = 100ps*
- **EOP= eye opening**
- **1200 bits are plotted**



# Intel B1, B12, B20, M1



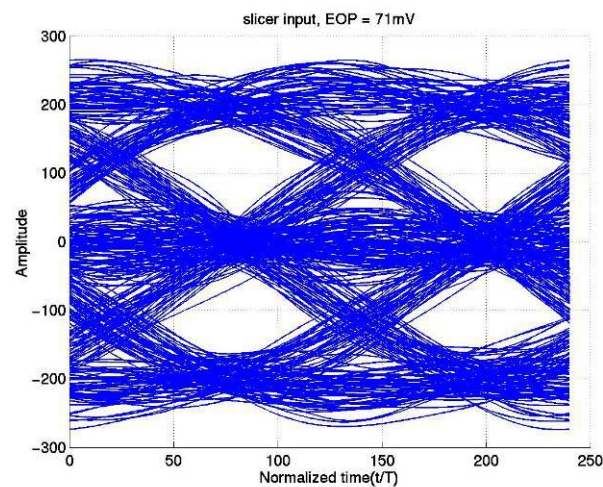
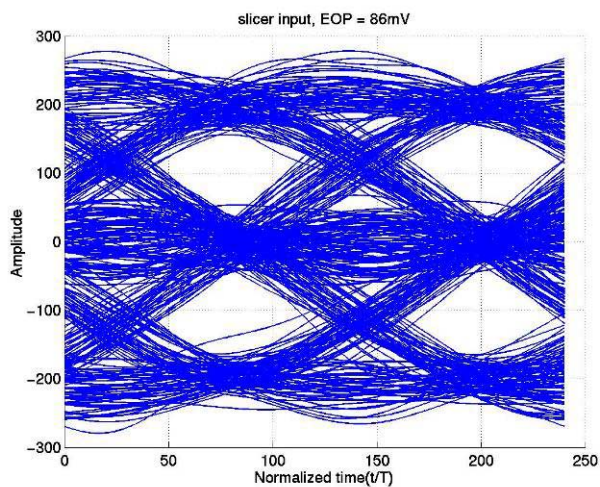
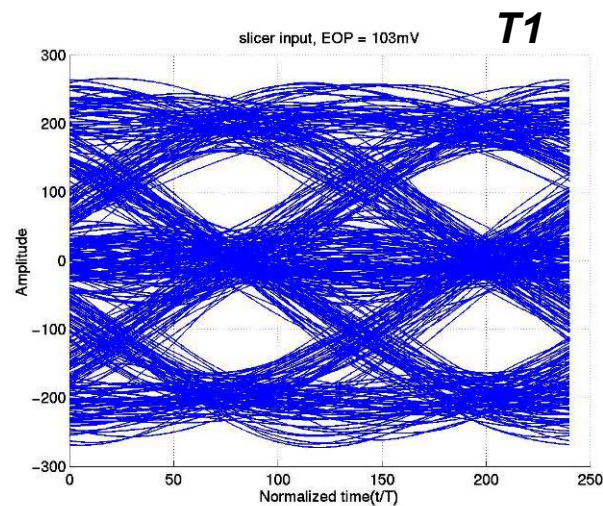
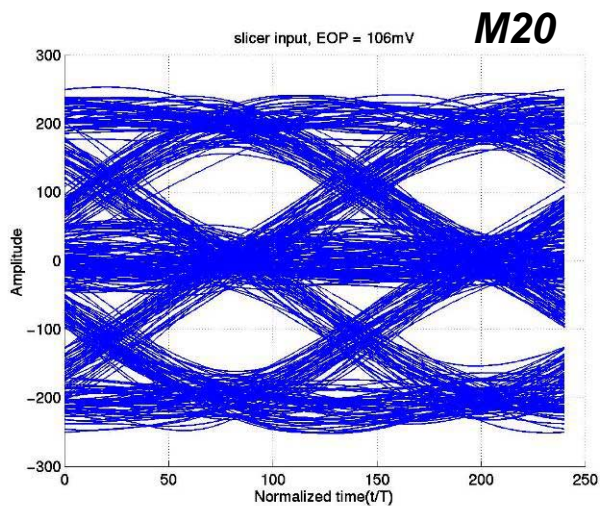
## Notes:

- **Vertical Axis**  
Units = mV
- **Horizontal Axis**  
120 units = 100ps
- **EOP= eye opening**
- **1200 bits are plotted**





# Intel M20 T1 T12 T20



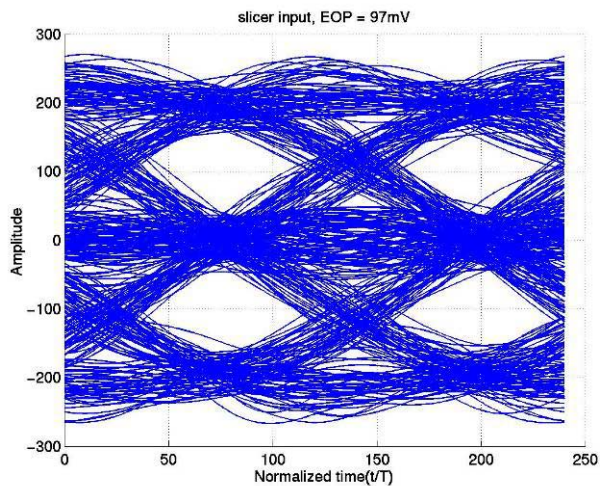
## Notes:

- **Vertical Axis**  
Units = mV
- **Horizontal Axis**  
120 units = 100ps
- **EOP= eye opening**
- **1200 bits are plotted**





# Intel T32



## Notes:

- **Vertical Axis**

**Units = mV**

- **Horizontal Axis**

**120 units = 100ps**

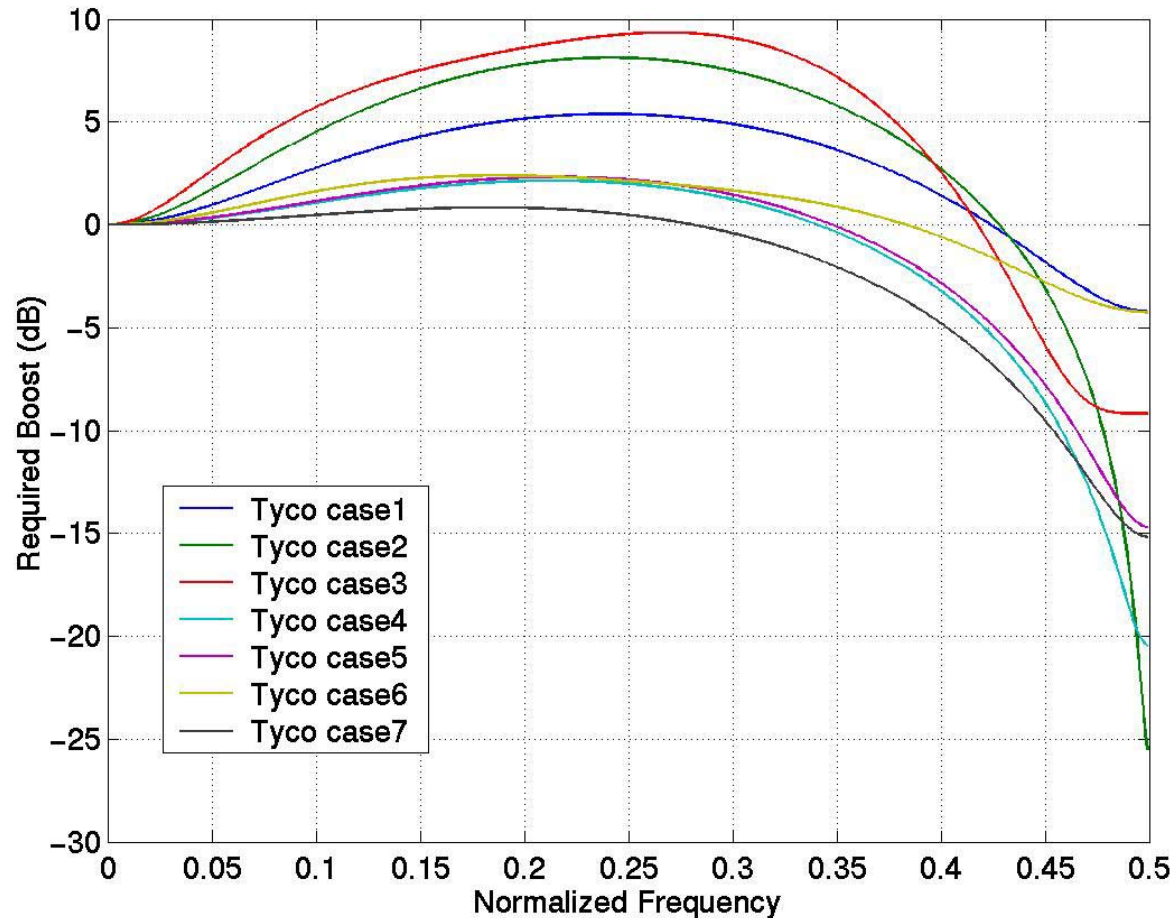
- **EOP= eye opening**

- **1200 bits are plotted**



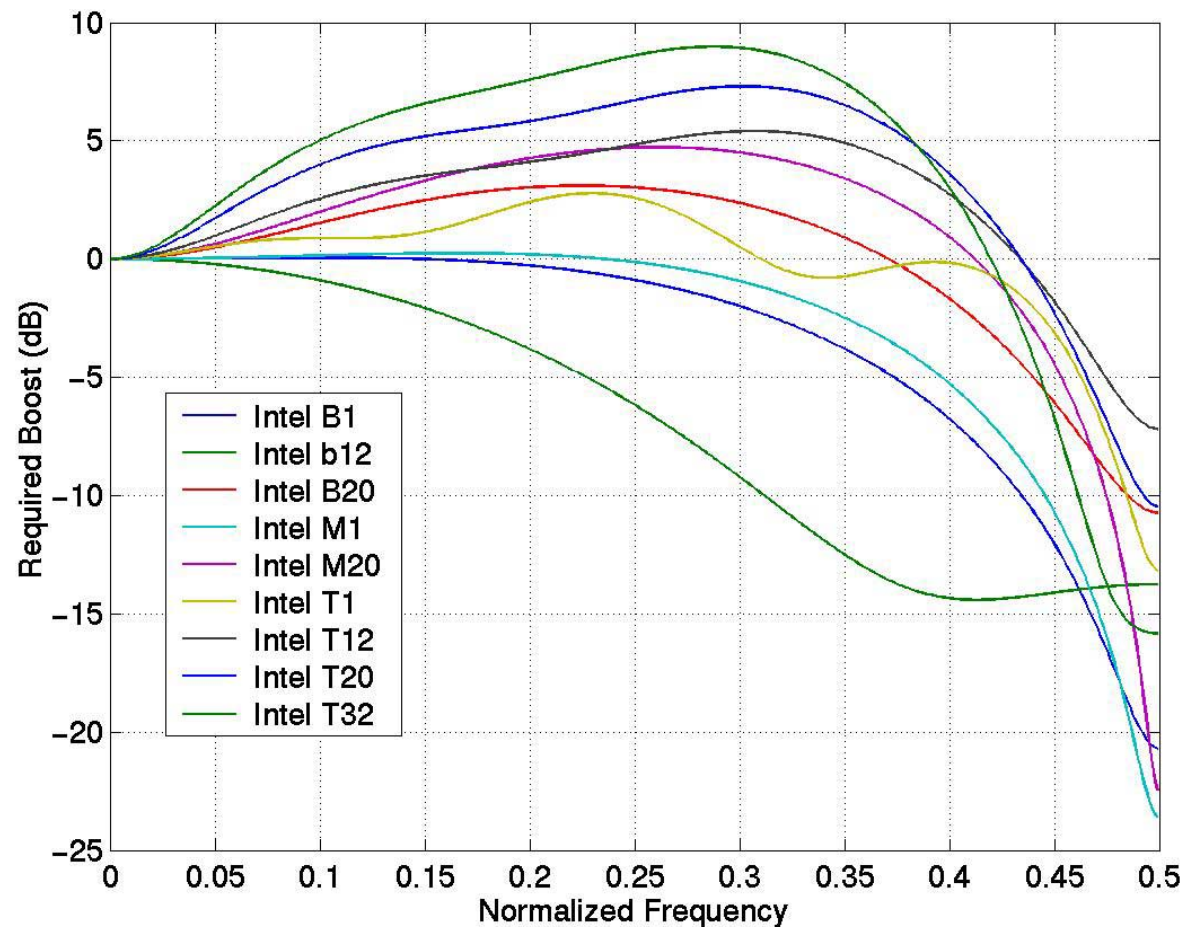
# FFE Boost required for each channel

## *Tyco Backplanes*



# FFE Boost required for each channel

## *Intel Backplanes*



# Discussion of Findings

- ❑ We were able to achieve open eyes on ***all*** of the channels analyzed.
- ❑ Less than 10 dB of boost is required in the FFE for ***all*** channels analyzed
  - Important for crosstalk sensitivity
  - Reduces required power consumption
  - Reduces complexity of FFE design
- ❑ Tyco channels required from 3-5 FFE taps and 3-5 DFE taps
- ❑ Intel channels were more challenging – required from 3-9 FFE taps and 3-5 DFE taps.



# Conclusion

- ❑ A comprehensive simulation of duobinary signaling has been carried out.
- ❑ Using Duobinary, we were able to achieve margin, even at a  $\text{BER}=10^{-18}$ , on ALL CHANNELS.
- ❑ FFE boost was  $< 10$  dB which will help us to achieve a signaling solution that has relatively low complexity and power requirements.
- ❑ **Clearly** duobinary signaling provides an attractive solution for the 10G PHY.

