Considerations for CDR Bandwidth Proposal

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Supporters

- Ali Ghiasi Ghiasi Quantum LLC
- Venugopal Balasubramonian Marvell
- Vipul Bhatt Inphi
- Upen Reddy Kareti Cisco
- Jane Lim Cisco

Comments addressed

- Comment 93 CRU BW for 400Gbase-DR4
- Comment 94 stress receiver sensitivity for 400Gbase-DR4
- Comment 95 CRU BW for CDAUI-8
- Comment 96 stress receiver sensitivity for CDAUI-8
- Comment 103 CRU BW for CDAUI-8
- Comment 104, 105 for clause 120D
- Comment 106 module stress receiver sensitivity for CDAUI-8
- Comment 109-115 for clause 120E

CRU Bandwidth in existing Standards

Let's group the latest standards by speed and modulation:

- NRZ 10Gb/s is fb/2578 \rightarrow 4 MHz
- NRZ 25Gb/s is fb/2578 \rightarrow 10 MHz
- PAM4 25Gb/s is 1.6 MHz
 - also includes a loop delay of 28.6ns

OIF proposals for 56G:

• CEI 56G LR and VSR: fb/8496 \rightarrow 3.13 MHz

In Atlanta, there was a proposal by Ali Ghiasi to lower the requirements on CDR/CRU bandwidth to ease receiver implementations

• This presentation will provide some more data in support

- Let's consider a number of TX PLL phase noise plots as available from published material
 - As published at ISSCC (2013, 2014, 2015 and 2016)
- The rms jitter is computed in several ways:
 - Use a first order high pass filter, golden PLL with no latency
 - Use a realistic CDR with implementation latency
 - Second order loop, 10ns latency, pole separation 5x or complex with chi=0.707
- Support the proposal to set a CDR bandwidth somewhere between 2 and 4 MHz
 - Not only golden PLL improvements are negligible, but realistic CDR implementation actually may deliver worse performance with larger bandwidths

RX Jitter and CDR Loop

- CDR loop tracks TX phase
 - Jitter is rejected with a high pass transfer function
- Higher bandwidth improves performance
 - Exploit corner frequency to lower requirements on TX
- Latency limits bandwidth related improvements



RX Jitter and Second order CDR Loop

- CDR with complex poles are more sensitive to latency
 - Higher peaking
 - Larger noise emphasis after corner frequency
 - Digital implementation more complex to limit latency
- We should account for practical limitations in CDR high frequency rejection capabilities Jitter TF with corner at 8 MHz vs. latency, complex poles



How to read the following plots

Phase noise data are filtered with a high pass filter with corner frequency as specified in the x axis
 Phase Noise Plot and golden-pil @ 2.0MHz



The resulting spectral density is integrated to compute rms and plotted on the y
axis
 Phase Noise Plot and jitter integral @ 2.0MHz



• For any point in the plot, the x axis specifies the corner frequency and the y axis the total filtered jitter

How to read the following plots

- Different filter shapes are applied, CDR latency is modeled
 - Golden PLL, single pole TF, no latency (green)
 - CDR #1, second order loop, pole spacing 5x, 10 ns loop delay (blue)
 - CDR #2, second order loop, complex poles (chi=0.707), 10 ns loop delay (red)
- Corner frequency for CDR#1 and #2 is only swept up to 10 MHz as second order loops start to peak significantly



Source: ISSCC 2013

Paper D2.3

- A Sub-2W 39.8-to-44.6Gb/s Transmitter and Receiver Chipset with SFI-5.2 Interface in 40nm CMOS
 - PLL operating range: 19.9-22.3 GHz
 - RMS jitter: 1kHz-320MHz 0.12 ps
- Negligible improvement with higher corner frequency



Source: ISSCC 2014

Paper 2.2

- A 780mW 4 × 28Gb/s Transceiver for 100GbE Gearbox PHY in 40nm CMOS
 - PLL operating range: 10-14 GHz
 - RMS jitter: 10kHz-100MHz 0.16 ps
- Marginal improvement increasing corner
 - Low absolute levels (<150 fs)



Source: ISSCC 2015

Paper 10.9

- A 13.1-to-28GHz Fractional-N PLL in 32nm SOI CMOS with a ΔΣ Noise-Cancellation Scheme
 - PLL operating range: 13.1-28 GHz
- Negligible improvement with higher corner frequency
 - Low bandwidth PLL makes large corner frequency useless



Source: ISSCC 2016

Paper 3.4

- A 40/50/100Gb/s PAM-4 Ethernet Transceiver in 28nm CMOS
 - PLL operating range: 9.9-15.5 GHz
 - RMS jitter: 1kHz-100MHz 0.181 ps
- Marginal improvement increasing corner
 - Low absolute levels (<150 fs)



Conclusions

From state of the art published PLLs, we can observe:

- Jitter as measured at TX through high pass filter show marginal improvements when corner frequency gets above 2-3 MHz
- Jitter as expected to be experienced by a real CDR show negligible improvements above 2 MHz
 - Second order CDR loop cannot be fully exploited in presence of latency

Propose to set CRU bandwidth to 2 or 3 MHz.

 Corner frequencies above 2-3 MHz appear to be questionable with respect to the increased implementation complexity and power consumption that is implied in the receiver.

Thanks.

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