

# Dual Port Type MDI

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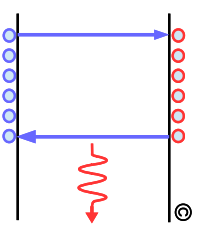
IEEE 802.3ck Meeting

San Diego

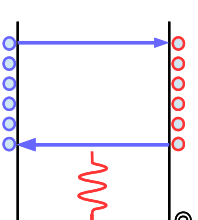
July 11, 2018

# List of Supporters

- ☐ Rob Stone – Broadcom
- ☐ Jane Lim – Cisco
- ☐ Pirooz Tooyserkani - Cisco
- ☐ David Piehler - Dell
- ☐ Xinyuan Wang – Huawei
- ☐ Hong Feng - Huawei



## Background



### □ During May interim meeting ghiasi\_3ck\_01a\_050918.pdf investigated 3 options for C2M and Cu MDI

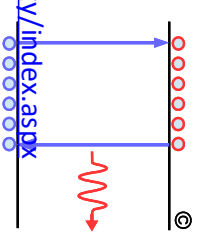
- Symmetric – single port type with 10.4 dB loss will double host retimers to support C2M applications
- Asymmetric – single port type switch-NIC ports with 14 dB loss for switch and 7 dB for NIC but NIC loss is restrictive
- Symmetric – dual port type 15 dB for C2M/AOC and 10.4 dB for port type supporting Cu

### □ After further investigations and consensus building 802.3ck group should consider symmetric dual port but compatible with 16 dB for C2M applications and ~11.5 dB for Cu MDI

- Symmetric single port type in order to support 2 m Cu restrictive on C2M applications and require adding retimers on the host card
- Asymmetric single port type supporting 2 m Cu cable challenging and require impractical loss on the NIC and does not support switch-switch applications

### □ This contribution also investigates how to improve measurement methodology and how to deal with mated boards for MDIs from one lane to 8 lanes.

# C2M Channel Reach



## PCB loss estimate assumptions and tools for calculation

- Rogers Corp impedance calculator (free download but require registration) <https://www.rogerscorp.com/acm/technology/index.aspx>
- The IEEE tool if updated could be another option to estimate channel reach [http://www.ieee802.org/3/bj/public/tools/Reference\\_DkDf\\_AlegbraicModel\\_v2.04.pdf](http://www.ieee802.org/3/bj/public/tools/Reference_DkDf_AlegbraicModel_v2.04.pdf)
- Stripline ~ 50  $\Omega$ , trace width is 5.5 mils, and with ½ oz Cu HVL
- Isola 408HR DK=3.65, DF=0.0095, RO=2.5  $\mu$ m, Meg-6 DK=3.4, DF=0.005, RO 1.2  $\mu$ m, Tachyon100 DK=3.02, DF=0.0021, RO=1.2  $\mu$ m
- To support equivalent PCB traces for C2M need at least 15 end-end channel loss consistent with tracy\_100GEL\_01a\_0118
- Assumed loss for two vias is 1 [dB@26.55 GHz](#).

Host Trace Length (in)	Total Loss (dB)	Via Loss (dB)	Host PCB Loss(dB) - 2 via loss	Isola 408HR	Megtron 6	Tachyon100
Nominal PCB Loss/in at 5.15 GHz	N/A	0.05	N/A	0.65	0.52	0.46
Nominal PCB Loss/in at 13 GHz	N/A	0.15	N/A	1.27	0.98	0.83
Nominal PCB Loss/in at 27 GHz	N/A	0.5	N/A	2.18	1.60	1.28
10GSEF+ with one connector & HCB	6.5	N/A	4.9	7.5	9.4	10.7
28G-VSR + stack connector *	10.3	N/A	6.31	5.0	6.4	7.6
100G Cu MDI SMT Connector **	11.5	N/A	7.5	3.4	4.7	5.9
100G Cu MDI Stacked Connector ***	11.5	N/A	6.5	3.0	4.1	5.1
C2M with SMT connector **	16	N/A	11.5	5.3	7.2	9.0
C2M with Stacked connector ***	16	N/A	10.5	4.8	6.6	8.2

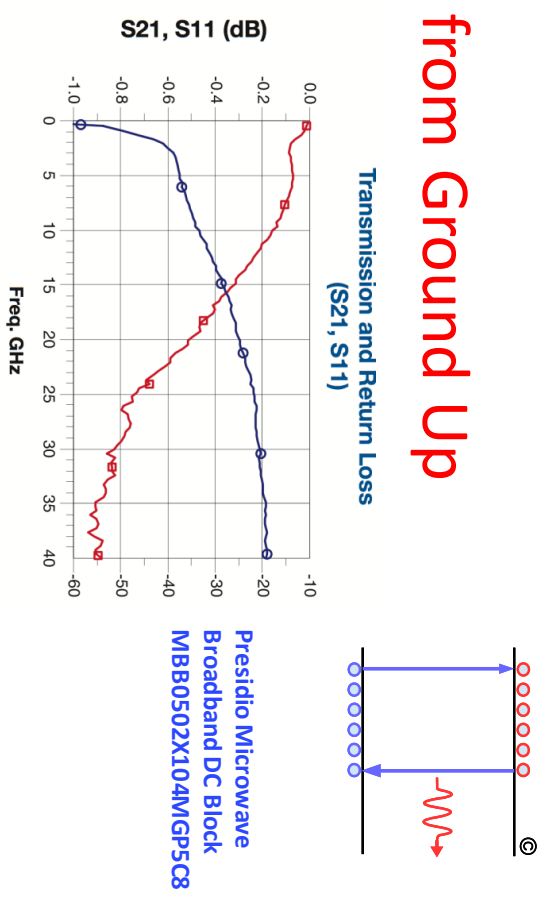
\* Assumes connector loss is 1.69 dB and HCB loss is 2.0 dB at 12.89 GHz

\*\* Assumes SMT connectors with 1 dB loss, 2.5 dB for HCB, and 1 dB for 2 vias at 26.55 GHz.

\*\*\* Assumes SMT connectors with 2 dB loss, 2.5 dB for HCB loss at 26.55 GHz, and 1 dB for 2 vias at 26.55 GHz.  
A. Ghiasi  
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# Building Cu Cable Assembly Loss from Ground Up

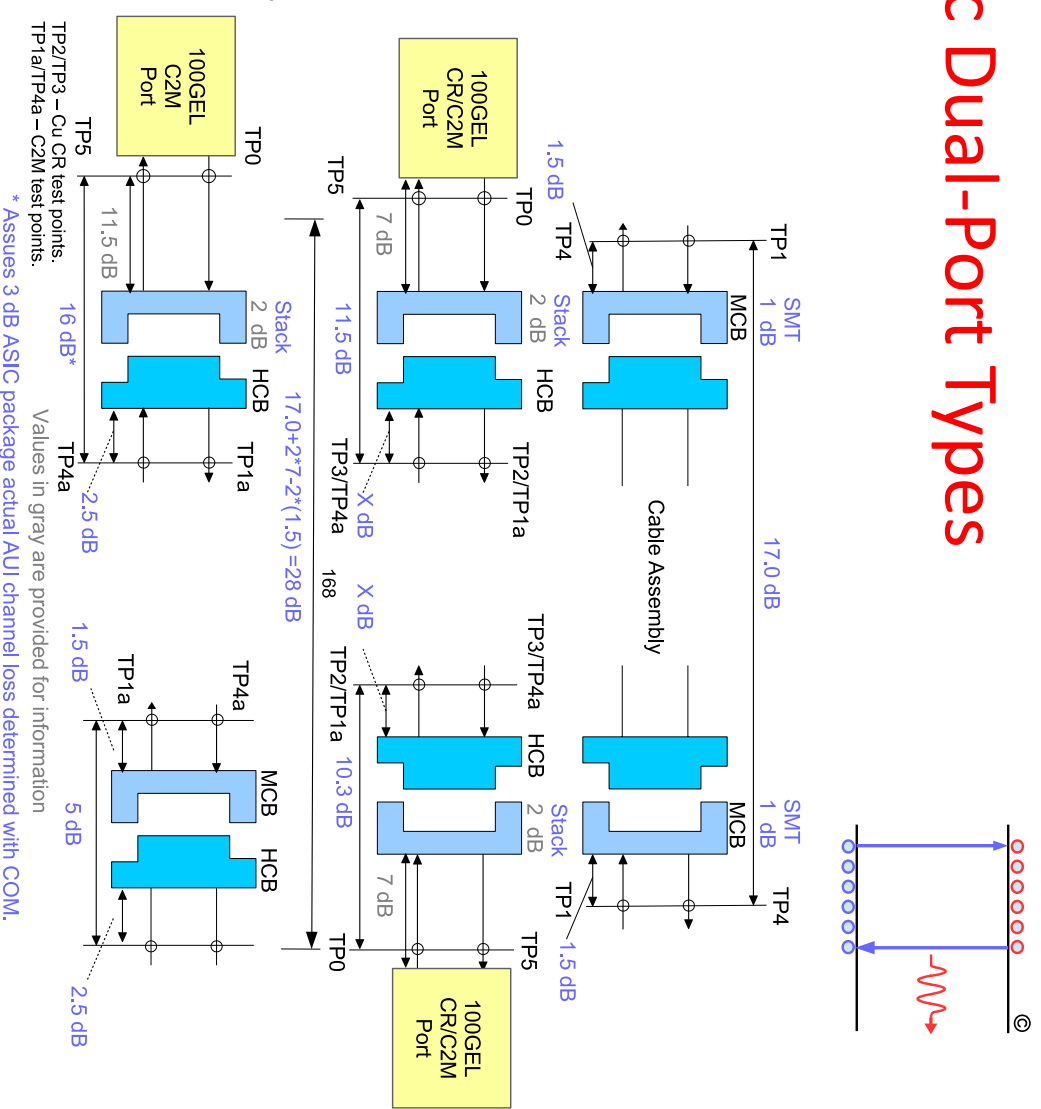
- ❑ Assuming 2 m objective can only be met with 26 AWG
  - Also supporting ~1.3 m on 28 AWG
- ❑ QSFP Cu cable loss estimate 4.85 dB/m for 26 AWG and 7.4 dB for 28 AWG, slightly better than reported in [http://www.ieee802.org/3/100GEL/public/18\\_03/palkert\\_100GEL\\_01a\\_0318.pdf](http://www.ieee802.org/3/100GEL/public/18_03/palkert_100GEL_01a_0318.pdf)
- ❑ Key assumed cable assembly losses:
  - DC block 0.8 dB one of
  - Plug PCB loss 0.75 dB 2 of
  - MCB connector 1 dB 2 of
  - Nominal MCB PCB board loss 1 dB 2 of
  - Analysis does not include any via loss associated with QSFP-dd rear contacts
- ❑ Reducing Cu cable assemblies loss <17.0 dB may result not meeting our 2 m reach objective!



Cable assembly element loss @ 26.55 GHz		2 m 26 AWG	1.3 m 28 AWG
Cable loss dB/m		4.85	7.40
Cable loss (dB)		9.7	9.6
Nominal MCB PCB loss 2 of (dB)		3.0	3.0
MCB connector loss 2 of (dB)		2.0	2.0
Cable plug PCB loss 2 of (dB)		1.5	1.5
DC block (dB)		0.8	0.8
Cable assembly end-end loss (dB)		17.0	16.9
Host PCB + Host Connector Loss (dB) (28 - 17.0 + 2*MCB Loss )/2		7.00	7.04

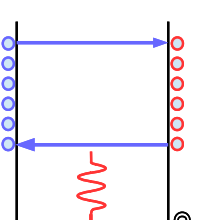
# Proposed Symmetric Dual-Port Types

- ❑ Symmetric dual-port type allow building a superset port supporting passive Cu cable and optical port/AOC or build an optical/AOC/Active Cu ports if passive Cu cable support not required
- ❑ As the figure illustrate the normative compliance points TP2/TP3 and TP1a/TP4a can support multiple MDIs and each of the MDIs may have distinct MCB/HCBs
- ❑ QSFP-dd/OSFP 8 lanes HCB may require construction of HCB1/2 or de-embedding
- ❑ Symmetric dual-port type with 11.5 dB host budget supports both C2M and Cu MDI
  - Proposed symmetric dual-port type budget assumes 28 dB ball-ball to support 2 m of passive Cu cable with loss of 17.0 dB
- ❑ C2M ports with 15 dB supports optical/AOC
  - C2M host channel loss based on 11.5 dB after assuming stack connector with 2 dB loss but an SMT host may allocate 11.5 dB
    - 2 Vias will reduce above budget by ~ 1 dB.



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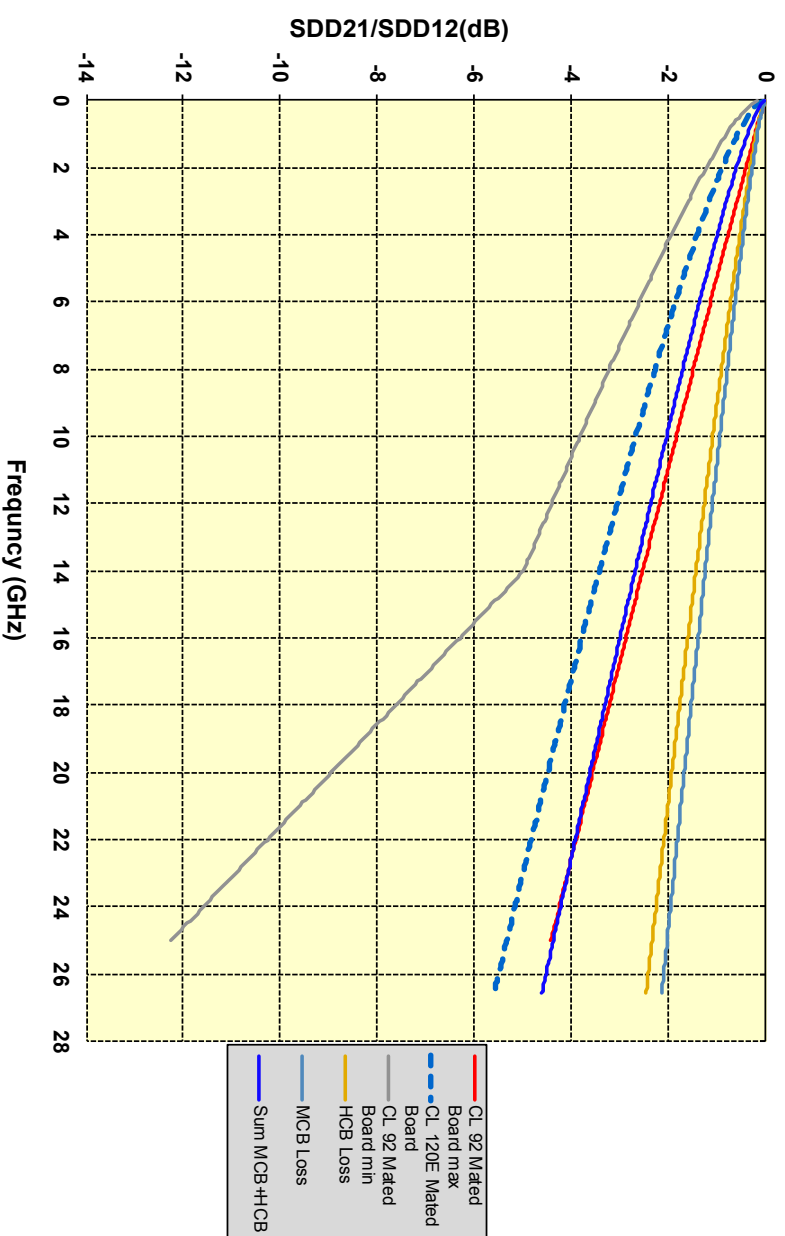
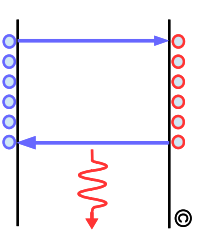
# Why Use COM for C2M



- ❑ **The normative TP1a/TP4a EW/EH historically measured with reference EQ on the scope**
  - Clause 120E defines C2M loss up to 10.2 dB
  - In practice any host/SerDes that deliver the TP1a EW/EH is compliant then why use COM?
  - We have a wide range of MDIs SFP-112, SFP-dd, QSFP112, QSFP-dd, and OSFP from 1-8 lanes with different crosstalk, ILD, and reflections
    - Crosstalk will be very different between 1 vs 8 lanes
    - Some of these connector may perform better than others
    - Stack connector vs SMT connector
    - More complex PCB routing to 8 lanes modules
  - COM can be used as channel design guide to improve compliance given the diverse set of ports supported
- ❑ **Propose starting point for COM/channel analysis (ADS, etc):**
  - 4 tap TX FFE (2 pre)
  - 5 tap RX FFE with 4 post or equivalent
    - To measure TP1a/TP4a signals the reference EQ needs to be implementable on sampling scopes
    - The 5 tap FFE T-spaced already used for TDECQ
  - Host ASIC package having 4 dB loss and CDR package with 1 dB loss @26.55 GHz
  - Operation up to 15 dB of well constructed channels with 4 dB package or 16 dB channel with 3 dB package
    - [http://www.ieee802.org/3/100GEL/public/18\\_01/tracy\\_100GEL\\_01a\\_0118.pdf](http://www.ieee802.org/3/100GEL/public/18_01/tracy_100GEL_01a_0118.pdf)
    - [http://www.ieee802.org/3/100GEL/public/18\\_03/lim\\_100GEL\\_01b\\_0318.pdf](http://www.ieee802.org/3/100GEL/public/18_03/lim_100GEL_01b_0318.pdf)

# Overview of Current Mated Boards

- IEEE 802.3 CL92.11 mated boards are the bases for CL120E.4.1 and CL 136B
  - CL120E/136B nominal mated board loss to CL92 mated boards
  - CL92 mated board bases were Ghiasi\_3bj\_01a\_0912.pdf
  - CL92 boards were constructed from Rogers 4350B with  $DF=0.037@10$  GHz where today Megtron-7N has  $DF=0.02@10$  GHz so improved board can be built now even with standard PCB material
- CL 92 states deviation from reference mated board loss should be accounted in the measurement
  - No specific method how to account for any deviation
  - Unless the mated board are improved drastically then it further raises the need to better account for board variations at 53 Gb/s!

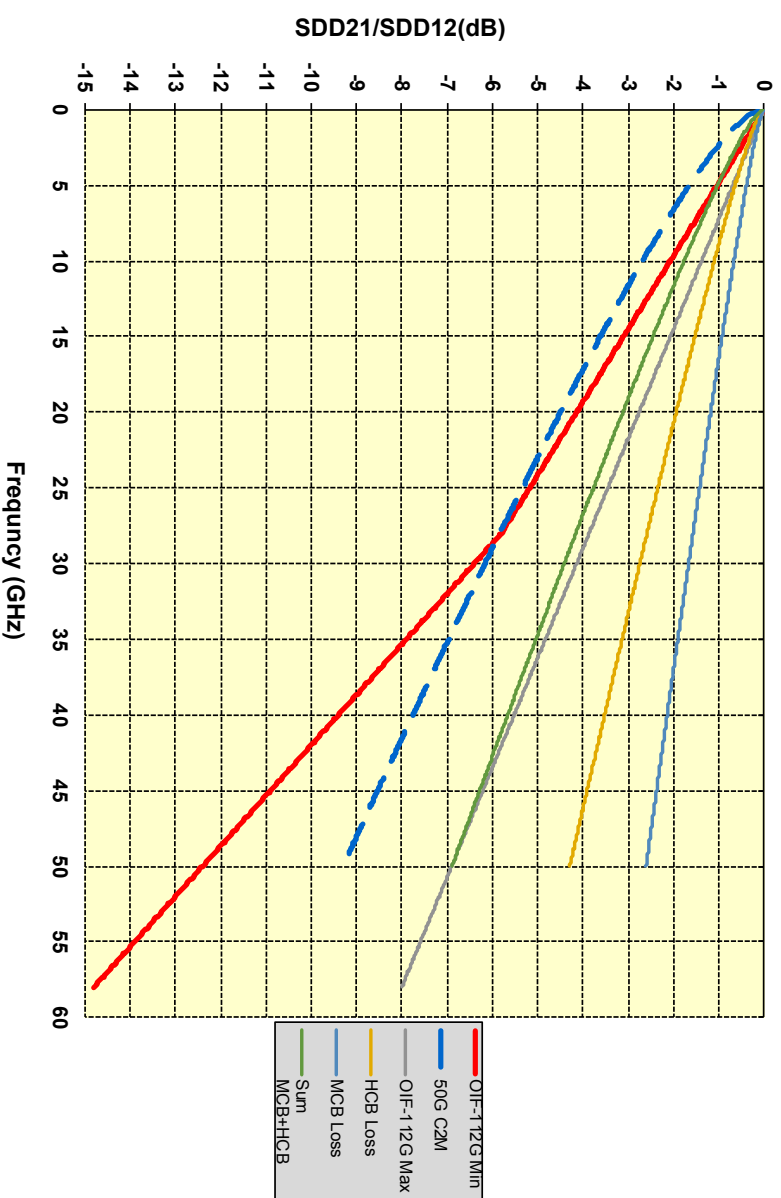
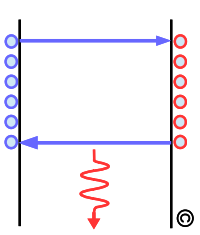




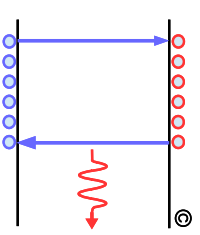
# Starting Point for 53 Gb/s Mated Boards

## Graph on the right extend CL92 and CL120E/CL136B from 25 GHz to 50 GHz

- In addition the graph includes OIF 112G-VSR mated boards
- At 28 GHz OIF-112G-VSR max loss overlaps with CL120E/CL136B mated board having a loss of 6 dB
- CL92 mated board loss can be improved by ~1/3 if one uses higher grade material instead of RO4350B that was used in Ghiasi\_3bj\_01a\_0912.pdf
- **We have more MDI with different attribute than ever before**
  - SFP112 – single lane
  - QSFP112 – 4 lanes
  - QSFP-dd – 8 lanes but 4 of the lanes have 2 via
  - OSFP – 8 lanes
- **Unless we account for MDI differences, will end up throwing away precious host PCB trace reach and increase product passing/failing hysteresis!**

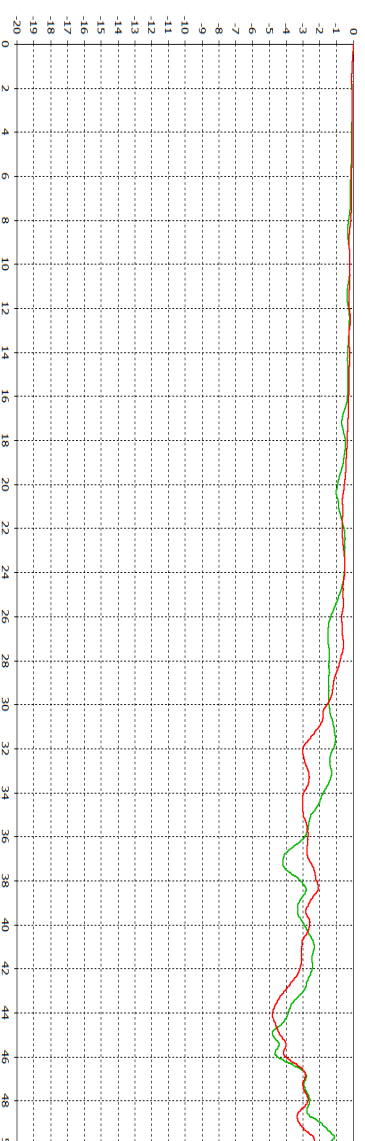


# Example of SMT Connector Suitable for 100GEL

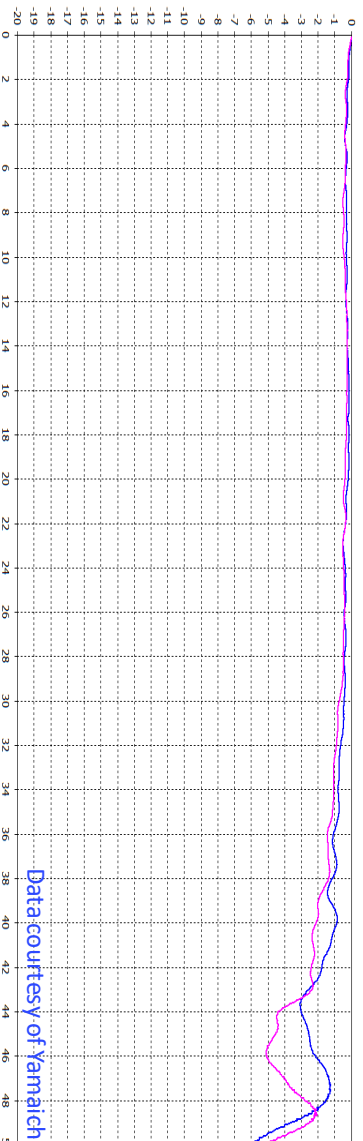


- ❑ Yamaichi QSFP-56 and OSFP SMT connector suitable for MCB construction has loss <1 dB
  - But the stack connector loss estimated to be in 2-2.5 dB @26.55 GHz!

QSFP-56



OSFP



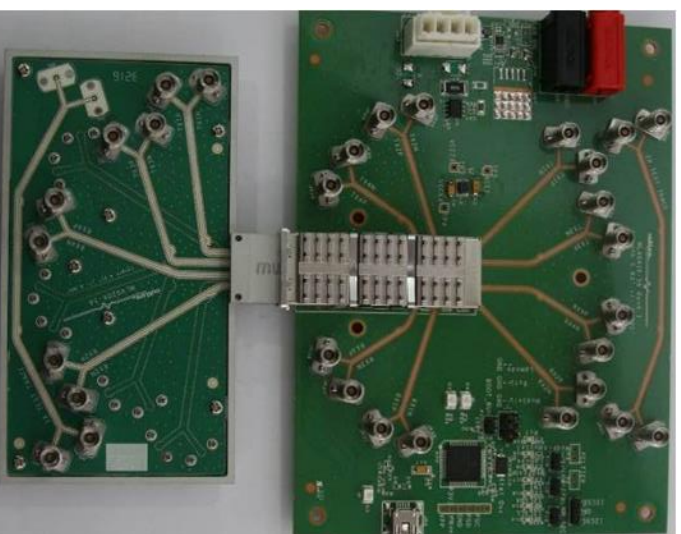
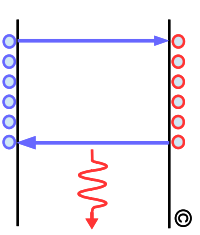
Data courtesy of Yamaichi

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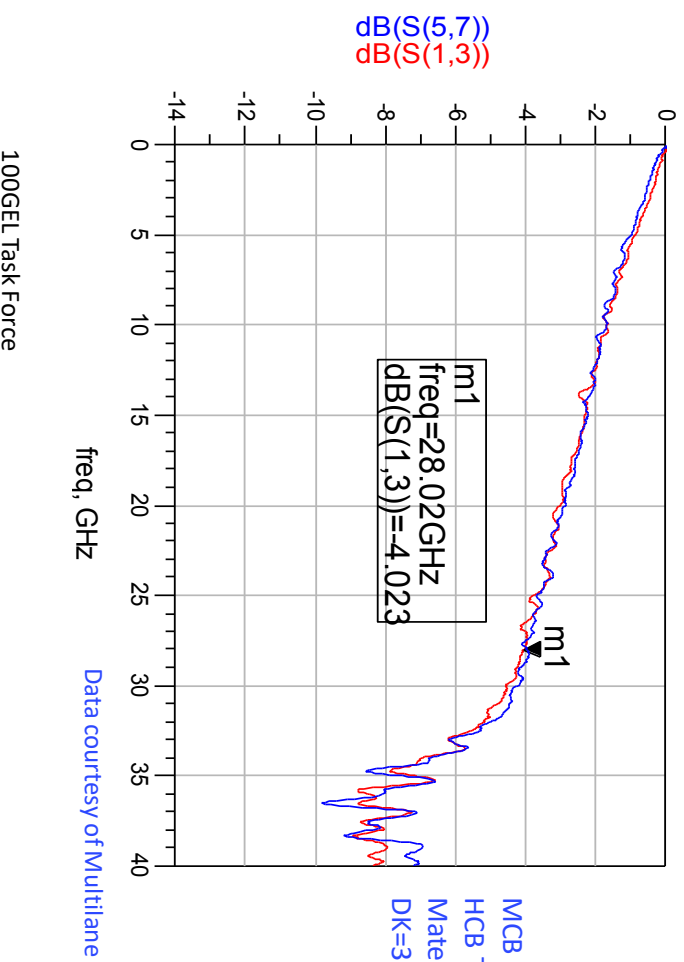
## Example Mated Board Suitable for 100GEL

### ❑ Multilane mated board using Yamaichi QSFP-56 connector

- A QSFP56 mated board below has about the same loss at 26.55 GHz as current 50G test boards at 13.27 GHz!
- An SFP112 board could be constructed to have even lower loss
- OSFP/QSFP-dd HCB with octal I/O would require at least 6" long HCB with narrower traces which may push the HCB loss to 5 and the mated board to 8 dB
- Should we instead consider cabled HCB or custom high density RF connectors for OSFP/QSFP-dd to reduce the loss to sub 5 dB?

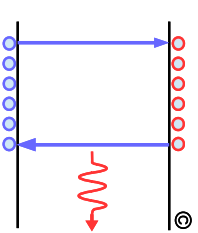


A. Ghiasi



MCB Trace length ~1.7"  
HCB Trace Length ~4.2"  
Material Rogers 3003  
DK=3.0, DF=0.001@ 10 GHz

# Distinct Port/Connectors Will Use Distinctly Different Mated Boards

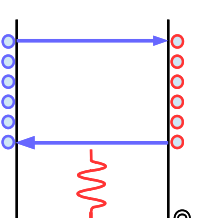


- ❑ Is it really outrageous to have different mated board specifications/loss for SFP112 vs QSFP-dd?
- Clause 136C defines each of these distinct SFP28, QSFP28,  $\mu$ QSFP, QSFP-dd, and OSFP connectors
- We shouldn't saddle the SFP112/QSFP112 with additional loss in order to have one mated board specifications
- We could define 5 mated boards but high loss 7.5-8 dB mated board will not provide representative results for QSFP-dd/OSFP!

**Table 136C-1—Number of PMDs supportable for each connector type**

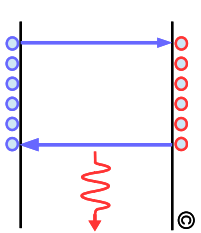
MDI types	50GBASE-CR	100GBASE-CR2	200GBASE-CR4	Mated Board Loss 26.55 GHz (dB) if all 8 signal pairs are broken
SFP28	1	—	—	~3.5 dB
QSFP28	1, 2, 4	1, 2	1	~4.0 dB
microQSFP	1, 2, 4	1, 2	1	~4.5 dB
QSFP-DD	1, 2, 4, 8	1, 2, 4	1, 2	~8 dB
OSFP	1, 2, 4, 8	1, 2, 4	1, 2	~7.5 dB

# How to Deal Perplexity of Mated Boards



- ❑ **Lets step back some 10 years ago when we created SFP+ mated board as part of SFF-8431**
  - The HCB RF connector location was representative of the module PMA/PMD input/output chip balls
  - The MCB RF connector was representative of min host loss
- ❑ **For sake of simplicity we can go with 1.5 dB MCB loss for all MDIs**
- ❑ **Technically we can define 4 different mate board specifications if needed with identical C2M specifications at TP1a/TP4a and Cu CR specifications**
- ❑ **We have two options how to deal with perplexity of HCB in support of OSFP/QSFP-dd**
  - Go with 2.5 dB HCB loss and mated board loss of 5 dB for all MDIs as shown on page 6 figure
    - QSFP-dd/OSFP HCB would need to be constructed with cable, use HCB1 and HCB2 to reduce trace length ~4", or use de-embedding
    - Define a set of optimized MCB/HCB for each of the representative MDIs SFP112, QSFP112, QSFP-dd, and OSFP and use higher loss for QSFP-dd/OSFP HCB but may not produce representative and accurate results
- ❑ **What ever scheme we choose shouldn't penalize SFP112/QSFP112 and should produce representative/accurate results for QSFP-dd/OSFP!**

# Summary



- ❑ **Dual port symmetric offers practical solution to support C2M and CRx applications**
  - The proposed C2M budget can support up to 16 dB with 3 dB package consistent with lim\_3ck\_01\_0718.pdf
    - To keep module PMA power reasonable should keep channel+ASIC package loss  $\leq 19$  dB
  - The proposed CRx TP0-TP2 is 11.5 dB
- ❑ **At 112G no longer we have luxury of extra margin and overall test methodology needs to improve**
  - Testing cables using an MCB with SMT connectors with lower loss/crosstalk but with deployed system using stacked connector may result in 2-3 dB increase in end-end loss and possibly as much as 6 dB increase in PSXT
    - COM could potentially provide two additional knobs to adjust for connector loss and PSXT
    - Another alternative is to test cable with MCB having stack connectors
  - Given the diverse set of MDIs with varying degree of crosstalk/loss need to consider using COM for C2M
- ❑ **802.3ck is defining PMDs for a diverse set of MDIs: SFP112, QSFP112,  $\mu$ QSFP, QSFP-dd, and OSFP**
  - Defining 8 dB mated board loss needlessly penalizes SFP112, QSFP112,  $\mu$ QSFP and wouldn't produce representative/accurate results equivalent to PMA/PMD BGA balls instead need to consider using HCB1/2, cabled HCB, or de-embedding
  - SFP112, QSFP112 and  $\mu$ QSFP can be constructed using premium PCB material with mated loss of  $\leq 5$  dB
    - We should not saddle SFP112/QSFP112 mated boards with extra loss and loose precious link budget
  - Overall the best option is to use 5 dB mated board loss for all MDIs and use HCB1/HCB2 or de-embedding for QSFP-dd/OSFP ports.