Constructing a BCH/Hamming Inner Code for 200 Gb/s per lane PMDs

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Background: Previous FEC Related Discussions

- In <u>welch_3df_01b_220602</u>, discussions were focused on feasibility and availability of FEC capability to enable 200 Gb/s per lane with IM-DD PAM4.
 - $\sim 2 2.4E-3$ is the possible pre-FEC BER range based on data in this group.
- <u>wang 3df 01a 220609</u> and <u>zhang 3df 01 220609</u> proposed 800 Gb/s single carrier coherent based on DP-QAM modulation for 800GBASE-LR1.
 - Analysis was based on ~4 4.5E-3 pre-FEC BER range.
- For 200 Gb/s per lane AUI, ~1E-5 pre-FEC BER is under discussion in the Task Force.
- Concatenated code with BCH/Hamming as inner code based on soft decision implementation is interested in the Task Force.



Background: Adopted Logic Layer Baseline

• Concatenated code is one of the FEC schemes adopted in logic architecture baseline in May.



• Motivation: Introduce general methods on how to construct a suitable inner code to support 200 Gb/s per lane PMDs of 800G/1.6TbE, and new 200/400GbE.



General Approach for Concatenated Codes

- Basic concepts:
 - Outer code: the first encoded FEC code. It appears on the "outer" side of the whole link.
 - Inner code: the later encoded FEC code. It appears on the "inner" side of the whole link.



- For 802.3dj, we assume RS(544,514) as the outer code.
 - To construct an inner code working with RS(544,514) code and can be adapted to the PCS/PMA architecture.
 - BCH code with t=1 and Hamming code are under discussion as moderate ASIC implementation cost and reasonable FEC performance as shown in <u>he_3df_01a_220308</u>, <u>bliss_3df_01a_220517</u>, <u>patra_3df_01a_2207</u>, <u>ghiasi_3df_02a_2207</u>, etc.



Brief Introduction of BCH Code

- Bose–Chaudhuri–Hocquenghem (BCH) codes are a class of cyclic error-correcting codes that are constructed using polynomials over a finite field (Galois Field).
- As the inner code of the concatenated architecture in this project, we only consider binary BCH codes to work together with the non-binary BCH outer code, such as RS(544,514).
- For any positive integers $m \ge 2$ and $t < 2^{m-1}$, there exists a binary BCH code with the following parameters:
 - Codeword length: $n = 2^m 1$
 - Number of message bits: k
 - Number of corrected bits: t
 - Galois field index, $GF(2^m)$: m
 - Number of parity bits: $n k \le mt$
 - Minimum hamming distance: $d \ge 2t + 1$



- BCH(n,k) code can be *extended* to eBCH(n+1,k) by adding an additional parity bit.
- BCH(n,k) code can be *shortened* to BCH(n-c,k-c) by removing the c bits of padded 0s.



Derive Overhead of Inner Code from Bit Rate in Ethernet

Choosing the inner code (n, k) with the right n/k ratio could enable simpler 200 Gb/s per lane based PLL design, similar as 66/64(or 33/32) and 34/32(or 17/16) overhead ratio we chose before.



• BCH(144,136), BCH(126,119) will both give a 225 Gb/s per lane rate.



Step-by-Step Construction of a BCH Code

- The following steps describe how to construct a narrow-sense binary primitive BCH code.
- 1. Determine the desired codeword length and error correction capability.
 - Finding the suitable $n(=2^{m}-1)$ and t (or d equivalent, d = 2t + 1).
- 2. Find a degree *m* primitive polynomial p(x) to construct the Galois Field $GF(2^m)$.
 - A primitive element α is the root of p(x).
 - Every non-zero element in $GF(2^m)$ can be expressed by α^j with integer *j*.
- 3. Find the minimal polynomials f(x) for each primitive element α^i (i = 1, 2, 3, ..., 2t) in GF(2^m).
 - The minimal polynomial $f_i(x) = (x c_1)(x c_2) \cdots (x c_l), l \le m, \{c_l\}$: conjugate root of $f_i(x)$ including α^i .
- 4. Get the generating polynomial $g(x) = LCM[f_1(x), f_2(x), ..., f_{2t}(x)]$
 - The generating polynomial is similar as the one we use to define the encode process in previous standard specification, like RS(544,514) code in CL119.2.4.4.
 - g(x) can be expressed as: $\sum_{i=0}^{p} g_i x^i = g_p x^p + g_{p-1} x^{p-1} + \dots + g_1 x + g_0$, which can be implemented using linear feedback shift register form and be illustrated as such. The number of parity bits is *p*. The coefficient g_i takes value 0 or 1.
- 5. Encode the k message bits with the generating polynomial g(x) to get n bits codeword.



Step-by-Step Construction of a Binary BCH Code (t = 1)

- The following steps describe how to construct a narrow-sense binary primitive BCH code with t = 1.
- 1. Determine the desired codeword length and error correction capability.
 - For t = 1, we can simply use $m = \lceil \log_2(n+1) \rceil$.
- 2. Find the degree m primitive polynomial to construct the Galois Field GF(2^m).
 - Primitive polynomials have been well studied in academic research.
 - Many polynomials can be found online such as: <u>https://www.partow.net/programming/polynomials/index.html#deg08</u>
- 3. The generating polynomial is the same as the primitive polynomial for t=1 primitive BCH.



Why and How We Choose BCH(144,136) (and other codes)

- Various codes with different lengths and t values are tried in a concatenated scheme.
 - Codeword length between 100 ~ 1000 bits were studied (m = 7:10). Some examples were shown in he_b400g_01_210426 and he_3df_01a_220308.
 - Shorter codeword length had better performance when using soft-decision and limited LRPs.
 - t = 1 codes have lower latency and lower complexity.
 - Overall shorter BCH codes with t = 1 works better with the RS(544,514) outer code.
- Shorten the m = 8 primitive BCH(255,247), by prefixing to the message bits a sequence of 0s.
 - E.g., we can use primitive polynomial $x^8 + x^4 + x^3 + x^2 + 1$ ("implicit + 1" notation 0x8E) to construct the code.
 - There are many other primitive polynomials with degree of 8: 0x95, 0xAF, 0xB1, 0xB2, 0xB4, 0xE1, 0xF3, ...
 - The zero prefix sequence is not transmitted and is only used to calculate the parity of the primitive code.
- Apply the overhead ratio n/k = 36/34 (18/17), we can have BCH(144,136).
- eBCH(76,68) used in <u>he_3df_01a_220308</u> can be constructed in a similar way, with m = 7 and 1-bit extension, and its overhead ratio n/k = 38/34 (19/17).
 - Example polynomial: $x^7 + x^3 + 1$
 - The same code length and overhead can also be a non-extended version of BCH(76,68) with m = 8.



BCH(144,136) Encoder Functional Model

- For $g(x) = x^8 + x^4 + x^3 + x^2 + 1$ of BCH(144,136), the 9 coefficient are 100011101, MSB on the left.
- The parity calculation shall produce the same result as the shift register implementation below in the similar form as in 119.2.4.4 for RS encoder and 115.2.3.3, 115.2.4.3.2 for BCH encoder.
- The outputs of the delay elements are initialized to zero prior to the computation of the parity for a given message.





Relationship between BCH and Hamming Code

- BCH code is a generalization of the Hamming code for multiple-error correction.
- The single-error-correcting binary BCH code of length 2^{r-1} is a Hamming code.
 - Codeword length: $n = 2^r 1, r \ge 2;$
 - Message length: $k = 2^r r 1$
 - Hamming Distance: d = 3;
- Hamming code can be constructed with parity-check matrix:

Bit position		1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	
Encoded data bits (Message bit:d1,d2,d3,,,)		p1	p2	d1	р4	d2	d3	d4	p8	d5	d6	d7	d8	d9	d10	d11	p16	d12	d13	d14	d15	
Parity bit (p1,p2,p3,,,) Coverage	p1	<		>		>		>		>		>		>		<		<		>		
	p2		<	1			<	<			<	>			<	<			<	<		•••
	p4				<	<	<	<					<	<	<	<					~	
	p8								<	~	<	>	~	>	<	~						
	p16																1	~	\checkmark	>	~	

Refer to: https://en.wikipedia.org/wiki/Hamming_code



Example: BCH(144,136) and Hamming(144,136)

- Construct the BCH code with a polynomial.
- Once the generator polynomial is selected and its LFSR is given, this code is uniquely defined.
 - Calling it BCH or Hamming does not matter.
 - Functional test and verification can be conducted easily with generator polynomial and LFSR.
 - This is the method IEEE 802.3bj/bs/bv used to define the RS and BCH FEC.

Generator polynomial
$$g(x) = x^8 + x^4 + x^3 + x^2 + 1$$
 \swarrow \bigcirc Generator matrix $G = [I_{136} | P^T]$ \bigcirc $H = [P | I_8]$

 I_n is an $n \times n$ Identity matrix.



Example: Double-Extended Hamming(128,119)

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• Definition used in ITU-T G.709.3 Annex D, and IEEE P802.3cw.

• Parity-check column vectors
$$\boldsymbol{g}(\boldsymbol{i}) = \begin{bmatrix} s_{1,i} \\ s_{1,i} \\ \vdots \\ s_{6,i} \\ s_{7,i} \\ 1 \end{bmatrix}$$
, where $\boldsymbol{i} = 64s_{6,i} + 32s_{5,i} + \dots + 2s_{1,i} + s_{0,i}$
 $s_{7,i} = (s_{0,i} \wedge s_{2,i}) \vee (\overline{s_{0,i}} \wedge \overline{s_{1,i}} \wedge \overline{s_{2,i}}) \vee (s_{0,i} \wedge s_{1,i} \wedge \overline{s_{2,i}})$

• Parity-check matrix

 $\bigcup_{i=1}^{n}$

 $\begin{array}{l} \boldsymbol{H} = & [g(0): g(62), g(64): g(94), g(96): g(110), g(112): g(118), g(120), g(122), g(124), \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ \boldsymbol{H} = \begin{bmatrix} \boldsymbol{P} \mid \boldsymbol{I_9} \end{bmatrix} \end{array}$

• Generator matrix

 $\boldsymbol{G} = [\boldsymbol{I_{119}} \mid \boldsymbol{P^T}]$ $\boldsymbol{I_n}$ is an $n \times n$ Identity matrix.





- Suggest to develop BCH inner code for concatenated FEC using the polynomial methodology in this contribution.
- BCH(144,136) can be a candidate inner code that well matches with RS(544,514) outer code and Ethernet rate.



Thank you.

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