Considerations for the choice of η_0

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Introduction

- The Channel Operating Margin (COM) calculation includes two terms that allocate margin for practical receiver implementations
 - 1. Input-referred noise spectral density η_0
 - 2. Minimum COM value required for channel compliance
- There has been a lack of clarity about what η_0 is intended to represent and this has led to challenges in choosing an appropriate value
- There has been no assessment of the minimum COM value to determine if it is sufficient to account for receiver impairments not addressed by η_0
- This presentation is intended to start a *conversation* about these terms

What should be considered in the value of η_0 ?

- Known sources of noise at the receiver input e.g., ...
- Thermal noise
- Receiver package near-end crosstalk (NEXT)
- Receiver package far-end crosstalk (FEXT)
- It could also include input-referred equivalents to likely sources of noise internal to the receiver
- More about this later...

Receiver package near-end crosstalk

Transmitter

Device

Near-end aggressor

Channel NEXT is included in the channel measurements input to COM.

Receiver package NEXT should be accounted for in the COM reference package model.

 R_d

Package PSNEXT model (example) -30 $PSNEXT = 30 \log_{10}(f / 53.1 \text{ GHz}) - 50$ -35 -40 -45 Pulse amplitude A_{ne} dB Unit interval 1 / f_b -50 Magnitude, -50 dB at 53.1 GHz Transition -55 -60 -65 Input chosen randomly from L-level alphabet -70 -75



time filter

A side note about transmitter and aggressor rise times

The device model has an s21 characteristic and hence an intrinsic rise time.

The rise time at the "bump" is the combination of this intrinsic rise time and the impact of the transition time filter.



Interim summary of the components of input-referred noise



Component	Spectral density, V ² /GHz
Thermal	2e-9
Package NEXT	3.6e-9
Package FEXT	f.f.s.
Other, margin	f.f.s.
Total	> 5.6e-9

An η_0 value of 4e–9 or 5e–9 V²/GHz is not sufficient to represent all of the input-referred noise sources that practical receivers will need to tolerate.

(a) "PSNEXT equivalent" is the flat input spectral density that produces the same PSNEXT RMS voltage at the receiver noise filter output.

(b) The thermal noise spectral density is $4k_BTR$ where k_B is the Boltzmann constant, T is (125C + 273.15) K, and R is $2R_d = 90 \Omega$ (matching the COM configuration spreadsheet).

What impairments can the minimum COM allocation cover?



- Likely sources of noise internal to the receiver e.g., ...
- Amplifier (gain, continuous-time equalization) output-referred noise
- Amplifier distortion
- Analog-to-digital converter (ADC) effective number of bits (ENOB)
- Sampling time jitter
- Certain noise sources may be better modeled by an additional receiver noise term, or referred to the receiver input, due to scaling with loss

Some test cases for a receiver impairment study

#	Die-die IL, dB	Source
1	23.5	mellitz_3dj_03_elec_230504, C2C_withXtalk_Mezz_1_PCB-25mm_25mm_*
2	32.7	mellitz_3dj_02_elec_230504, KRCA_wXTALK_MX_4_PCB-25-25_mm_FO-200-200_mm_CA-200_mm_*
3	40.5	shanbhag_3dj_01_2305, CR_1mOSFPDAC_TP0TP5_25p9dB_PCBHost_4p9dB_*

Normalize all of the test cases to COM = 3 dB by adding noise at the receiver FFE output.

#	Rx	Added noise for COM = 3 dB, mV RMS			
1	DFE	2.7			
2	DFE	1			
3	MLSE	0.4			



COM implementation penalty due to ADC ENOB



#	Rx	Minimum ENOB for COM > 0 dB				
1	DFE	5.2				
2	DFE	5.8				
3	MLSE	5.6				

#	Rx	COM penalty for ENOB = 6, dB				
1	DFE	1.3				
2	DFE	2.5				
3	MLSE	1.9				

This one receiver impairment consumes the bulk of the 3 dB implementation allowance.

What about the other impairments?

NOTE – ENOB is modeled as a flat spectral density at the input to the receiver FFE. It can be equated to low-frequency ENOB since it does not include degradation at higher frequency due to jitter. This also means that the jitter impairment is <u>not included</u> in this estimate of the COM penalty. COM penalty calculation assumes 90% loading of the ADC input.

Other consideration for the minimum COM allocation



- Differences between reference equalizer and practical implementation e.g., limited number of taps, finite precision arithmetic
- Imprecise performance predictions due to use of approximations
- Optimistic performance predictions due to idealized models

Summary and next steps

- Commonly-used values for η_0 do not appear to be large enough to account for the most likely sources for receiver input-referred noise
- Certain receiver impairments can be expected to scale with channel loss
- A fixed allocation for these impairments e.g., a minimum COM, may be too generous for low-loss channels and insufficient for high-loss channels
- More explicit accounting of these impairments using new or existing noise terms is needed
- This presentation is intended to start a *conversation*
- The desired outcome of that conversation is consensus on what the noise terms represent and values for those noise terms that enable reasonable implementations to comply with the standard
- The minimum COM allocation may then need to be revisited based on the consensus noise definitions

Back-up materials

COM configuration spreadsheet, 1 of 2

	Table 93A-1 parameters			I/O control			Table 93A–3 parameters			
Parameter	Setting	Units	Information	DIAGNOSTICS	1	logical	Parameter	Setting	Units	Information
f_b	106.25	GBd		DISPLAY_WINDOW	1	logical	package_tl_gamma0_a1_a2	2 [0 0.0008455 0.000340225]		
f_min	0.05	GHz		CSV_REPORT	0	logical	package_tl_tau	0.00644805	ns/mm	
Delta_f	0.01	GHz		RESULT_DIR	esults\CACR_set1_{da	ite}\	package_Z_c	[92 92 ; 70 70; 80 80; 100 100]	Ohm	
C_d	[0.4e-4 0.9e-4 1.1e-4;0.4e-4 0.9e-4 1.1e-4]	nF	[TX RX]	SAVE_FIGURES	0	logical	z_p select	2		[test cases to run]
L_s	[0, 13 0, 15 0, 14; 0, 13 0, 15 0, 14]	nH	[TX RX]	Port Order	[1324]		z_p (TX)	[6 31; 1 1; 1 1; 0.5 0.5]	mm	[test cases]
С_Ь	[0.3e-4 0.3e-4]	nF	[TX RX]	RUNTAG	KR_set1_eval_		z_p (NEXT)	[829;11;11;0.50.5]	mm	[test cases]
R_0	50	Ohm		COM_CONTRIBUTION	1	logical	z_p (FEXT)	[6 31; 1 1; 1 1; 0.5 0.5]	mm	[test cases]
R_d	[4545]	Ohm	[TX RX]				z_p (RX)	[829;11;11;0.50.5]	mm	[test cases]
A_v	0.386	V	vp/vf=	TDR and E	RL options		C_p	[0.5e-4 0.5e-4]	nF	[TX RX]
A_fe	0.386	V	vp/vf=	TDR	1	logical				
A_ne	0.6	V		ERL	1	logical		Filter: Rx FF	FE	
L	4			ERL_ONLY	0	ns	ffe_pre_tap_len	6	U	
М	32			TR_TDR	0.01		ffe_post_tap_len	60	U	
	filter and Eq			N	4000	logical	ffe_tap_step_size	0		
f_r	0.58	*fb		TDR_Butterworth	1		ffe_main_cursor_min	0.7		
c(0)	0.55		min	beta_x	0		ffe_pre_tap1_max	0.7		
o(-1)	[-0.4:0.02:0]		[min:step:max]	rho_x	0.618		ffe_post_tap1_max	0.7		
o(-2)	[0:.02:0.1]		[min:step:max]	TDR_W_TXPKG	0	UI	ffe_tapn_max	0.7		
o(-3)	0		[min:step:max]	N_bx	20					
o(-4)	0		[min:step:max]	fixture delay time	[00]			Operational		
c(1)	[-0.2:0.05:0]		[min:step:max]	Tukey_Window	1		ERL Pass threshold	10	dB	
N_b	1	UI		Noise, jitt	er	UI	COM Pass threshold	3	db	
b_max(1)	0.75		As/dffe1	sigma_RJ	0.01	UI	DER_0	1.00E-04		
b_max(2N_b)	0.3		As/dfe2N_b	A_DD	0.02	V^2/GHz	T_r	0.00400	ns	
b_min(1)	0		As/dffe1	eta_0	4.00E-09	dB	FORCE_TR	1	logical	
b_min(2N_b)	-0.15	S	As/dfe2N_b	SNR_TX	33		PMD_type	C2C		
g_DC	[-15:1:-3]	dB	[min:step:max]	R_LM	0.95		EW	1		
f_z	25.16	GHz					MLSE	0	logical	
f_p1	40.00	GHz					ts_anchor	1		
f_p2	56.00	GHz					sample_adjustment	[-32 32]		
g_DC_HP	[-5:1:0]		[min:step:max]				Local Search	2		
f_HP_PZ	1.328125	GHz								
Butterworth	1	logical	include in fr							

NOTE – This configuration was used exclusively for the purpose of producing the examples shown in this presentation. It is not a proposal for COM parameter values.

COM configuration spreadsheet, 2 of 2

SAVE_CONFIG2MAT	0			
	Receiver testing			
RX_CALIBRATION	0	logical		
Sigma BBN step	5.00E-03	V		
	ICN parameters	•		
f_v	0.278	Fb		
f_f	0.278	Fb		
f_n	0.278	Fb		
f_2	61.625	GHz		
A_ft	0.450	V		
A_nt	0.450	V		
Parameter	Setting			
pard_tl_gamma0_a1_a2	[0.6.44084e-4_3.6036e-05]	1.4 db/in @ 53.125G		
board_tl_tau	5.790E-03	ns/mm		
board_Z_c	100	Ohm		
z_bp (TX)	32	mm		
z_bp (NEXT)	32	mm		
z_bp (FEXT)	32	mm		
z_bp (RX)	32	mm		
z_bp (RX) C_0	32 [0.2e-4 0]	mm nF		
z_bp (RX) C_0 C_1	32 [0.2e-4 0] [0.2e-4 0]	mm nF nF		
z_bp (RX) C_0 C_1 Include PCB	32 [0.2e-4 0] [0.2e-4 0] 0	mm nF nF logical		
z_bp (RX) C_0 C_1 Include PCB Seletions (<u>32</u> [0.2e-4 0] [0.2e-4 0] 0 rectangle, gaussian,dual_ra	mm nF nF logical ayleigh,triangle		
z_bp (RX) C_0 C_1 Include PCB Seletions (stogram_Window_Weight	<u>32</u> [0.2e-4 0] [0.2e-4 0] 0 rectangle, gaussian,dual_r. gaussian	mm nF NF logical ayleigh,triangle selection		

N_bg	N_bg 0 012 or 3 groups			
N_bf	N_bf 4 taps per group			
N_f	80	UI span for floating taps		benartsi_3df_01a_2211
bmaxg	0.2	max DFE value for floating taps		mli_3df_02_220316
B_float_RSS_MAX	0.1	rss tail tap limit		
N_tail_start 25 (UI) start of t		(UI) start of tail taps lim	it	

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