### Burst Mode Technology

A Tutorial

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**Quantum Bridge** 

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#### Overview

▲ Burst Mode Transmitters ▲ *Rise and fall times* ▲ Automatic power control ▲ Burst Mode Receivers ▲ Several approaches to level recovery ▲ *Fast clock recovery* ▲ *Delimiters and error tolerance* 

# Rise and Fall times

- Conventional circuits are designed to maintain a constant bias current
- ▲ No attention was paid to being able to change the bias current quickly
- ▲ It is no surprise that some conventional circuits have very poor performance
- However, some conventional circuits have pretty good performance



#### Positive solution to problem

Simple argument: if you can build a circuit that can modulate the diode at the bit rate, then you must also be able to modulate the bias current at the same speed

*Existence proof is following circuit* 



#### Basic Take-away Message

 Laser driver circuits can be designed that have short Ton and Toff performance
 155 Mb/s systems have Ton=Toff=6.4ns
 Such devices carry little to no complexity premium over 'standard' drivers

### Automatic power control

Conventional circuits often use
 Slow monitor photodiode
 Analog filters to average signal
 Analog control loop to maintain desired operating point.
 Burst mode prevents the use of simple

analog control loops

# Digital Burst APC

Monitor diode output is sampled at appropriate points in burst waveform
 Samples drive digital control loop
 Drive outputs are stored in memory for 'instant on' capability
 Such a scheme can be built using a cheap

*micro-controller device* 

#### **Extinction Ratio Control**

- Ideally, APC should maintain ER as well as average power
- ▲ Given enough 'structure' in the physical layer overhead, this can be done
- Power control fields that are all zeros or all ones allow the slow monitor to accurately measure these levels
- ▲ Digital control takes care of the rest

# Example of Tx control

▲ At end of burst, when no more data is to be sent, a Tx control sequence can be sent  $\checkmark$  Tx control is a block of all-zeroes The length of the block determines the required speed of the monitor diode ▲ The OLT doesn't see this signal *Would require an extra 'signal' from the* MAC to the PMD to start Tx control

# Example of Burst with Transmitter Control Appendix



#### Overview

▲ Burst Mode Transmitters ▲*Rise and fall times* ▲ *Automatic power control* ▲ Burst Mode Receivers ▲ *Dynamic Sensitivity Recovery* ▲ Level Recovery ▲ Fast Clock Recovery ▲ Delimiters and error tolerance

# Dynamic Sensitivity Recovery

- A weak burst following a strong burst is hard to see
- The recovery process is limited by:
  Photodiode carrier transport effects
  Amplifier slew and charging rates
  Unintentional "AGC" effects

### Photodiode Effects

▲ A good PIN is actually quite linear

▲ Some PIN diodes are made so that light can stray into low field regions of the junction

▲ Carriers generated there are slow, and lead to a long tail in time response

▲ The solutions include

▲ Only use diodes that don't have the 'tail'

▲ Use an AC-coupled level recovery scheme

# Amplifier Slew Rate

 $\checkmark$  The analog chain must settle quickly ▲ *Limitations are mainly RC time constants* ▲ Solutions are: decrease R and C ▲ Integration helps to reduce C ▲ An analog 'reset' can momentarily reduce R, substantially accelerating recovery

### Unintentional AGC

▲ In general, a good burst mode preamplifier should have a simple transfer function, and no 'memory'

▲ Many pre-amps exhibit slow gain compression characteristics

This is great for continuous mode
 This is the <u>kiss-of-death</u> for burst mode
 Solution: choose your amp wisely

#### Receiver Level Recovery

▲ The first challenge is to restore the logic levels to the burst mode signal  $\checkmark$  DC coupled methods ▲ Feedback (automatic gain control) ▲ Feedforward (automatic threshold control) ▲*AC* coupled methods ▲ *Frequency domain (analog filters)* ▲ *Time domain (differential delay receiver)* 

### DC methods

- Concept is simple: when each burst comes in, measure its power levels, and adjust accordingly
- ▲ Implementation requirement: signal path must be linear and DC coupled up to decision circuit
- Theoretically, this approach has low burst mode penalty
- ▲ Can be limited by nonlinear decay elements (like poor amplifiers or slow photodiodes)

# Feedback Topology





#### AC methods

▲ Basic concept: Make the Rx channel so that it rejects the burst-to-burst level shifts while maintaining signal integrity ▲ A high-pass filter does the job ▲ Level shifts are relatively slow signals ▲ Data bits are relatively fast signals ▲ Theoretically, this approach carries a small sensitivity penalty (~1.5dB) ▲ Good rejection of all slow decaying signals







### Measured performance

▲ DC based schemes at gigabit rates have been reported as 8~40 ns for total Tlr + Tdsr

Frequency domain AC schemes at 622 rates have been measured at 8 ns for total Tlr + Tdsr

Time domain AC schemes can approach single bit duration recovery times

# Fast Clock Recovery

Classical clock recovery (PLL) does not work well

Clock recovery falls into general classes
 Oversampling in time
 Oversampling in space
 Instant locking

▲ Instant locking

# Oversampling in time

Works by sampling the signal at several times the bit rate

- ▲ Best sample is selected by comparing to known good pattern (preamble)
- ▲ Becomes impractical at high rates
  - ▲ *Gb/s bit rate would require* ~5 *Gsamp/s*

# Oversampling in space

Works by generating several copies of the clock, each delayed by a different phase
 Best re-timing phase is determined by comparing outputs to known good pattern
 Approach is scalable

▲ *Requires low-clock skew circuits* 

# Instant locking

▲ Works by triggering the local clock on each incoming data transition

- Local clock carries system through periods of no transitions
- ▲ *Approach is scalable*
- ▲ Has a susceptibility to transient pulse distortions

#### Burst Delimiter

Signal is used to find the logical start of burst
 Provides fast protocol synchronization
 Standard synch methods don't work

# Analysis method

- The delimiter problem is equivalent to finding the true delimiter symbol from the set of symbols arising from time shifted segments of the preamble-delimiter sequence
- ▲ The discrimination of code symbols in the presence of errors can be described by the Hamming distance

The error resistance of a delimiter symbol is equal to N errors if its minimum Hamming distance is 2N+1 from all other symbols

#### Robustness needs and limits

▲ How robust must a delimiter be? ▲ Assume the raw BER is 1e-4 ▲ Assume delimiter lengths of 8 to 20 bits  $\checkmark$  At least 3 bit errors must be tolerated so that *burst error rate is* <1*E*-12 ▲ Delimiter should have Hamming distance of 7 ▲ How robust could a delimiter be? ▲ Assuming a preamble that is 1010 repeating pattern ▲ A delimiter of 2N bits can have a minimum Hamming distance no greater than N from the preamble

### Results 1

- Maximal Minimum Hamming distances computed for a selection of delimiter sizes via exhaustive search of all Delimiters
- Number of "Good delimiters" was found
  Good Delimiter has maximal minimum distance

▲ Good Delimiter has equal number of 1's and 0's Delimiter Maximal Number of

Length (bits) Minimum Distance Good Delimiters

8	3	17
12	5	78
16	7	311
20	9	713

# Results 2

- The set of "Good delimiters" can be further reduced by finding those with a minimum number of low weight distances from other symbols
- ▲ These could be described as the "Best delimiters"
- ▲ For 8 bit delimiters, there are 7 such codes:
  ▲ 1B, 27, 2D, 8D, 93, D8, E4
- ▲ For 16 bit delimiters, there are 5 such codes:
  ▲ 85B3, 8C5B, B433, B670, E6D0
- ▲ For 20 bit delimiters, there is 1 such code:
  ▲ B5983

### Summary

Burst mode technology is not new
 Large volume of scientific literature
 Many systems have reduced it to practice
 Using these design principles, one can achieve good performance for no extra cost

Interested parties should work together on finding consensus values